#### Matrix Formulation of Visibility Statistics

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## The Mythical Thermal Noise

- "We want to integrate down to the thermal noise level"
- The ideal radiometer equation gives this level as

$$\sigma_{\rm P} \propto \frac{P}{\sqrt{K}}$$
 (Dicke, 1946),

where *P* is expected power and *K* is number of measurements (time-bandwidth product  $B\tau$  for continuous-time signals)

- This is actually the (*standard*) *RMS error of the sample variance estimator*  $\hat{P} = \frac{1}{K} \sum_{k=1}^{K} x_k^2$  size of statistical fluctuation of power measurement (Dicke)
- Purely a result of measuring on a finite sample

## Studies of Visibility Statistics

#### Kulkarni (1989)

- *Self-noise:* Strong sources show same statistical fluctuations as strong receiver noise it's the total received power that matters in radiometer equation
- For strong sources, visibilities are correlated (especially on baselines with shared antenna)
- Statistics of images and triple products
- Gwinn (2001, 2004, 2006)
  - PDF for product of correlated Gaussian variables (scalar)
  - Effect of quantisation on this PDF

#### It's All About Covariance Matrices

Describe instantaneous real samples from *N* antennas by *N*-dimensional zero-mean Gaussian random variable (RV)
 *x* ~ N<sub>N</sub>(**0**, *R*) with (*true*) covariance matrix *R* and PDF

$$p(\mathbf{x}) = \frac{1}{(2\pi)^{-N/2} |\mathbf{R}|^{1/2}} \exp\left[-\frac{1}{2}\mathbf{x}^{H}\mathbf{R}^{-1}\mathbf{x}\right]$$

• Correlator calculates sample covariance matrix

$$m{S} riangleq rac{1}{K} \sum_{k=1}^{K} m{x}_k m{x}_k^T$$

by multiplying and adding *K* of these vectors per dump

• Can also define unnormalised scatter matrix

$$oldsymbol{W} riangleq Koldsymbol{S} = \sum_{k=1}^{K} x_k x_k^T$$

## Sample Covariance Matrix

• Visibilities are not independent measurements, but elements of a matrix

	<i>s</i> <sub>11</sub>	<i>s</i> <sub>12</sub>	<i>s</i> <sub>13</sub>	• • •	$s_{1N}$
	s <sub>21</sub>	s <sub>22</sub>	s <sub>23</sub>	• • •	s <sub>2N</sub>
S =	$s_{31}$	s <sub>32</sub>	<i>s</i> <sub>33</sub>	•••	$s_{3N}$
	:	:	:	۰.	÷
	$s_{N1}$	$s_{N2}$	$s_{N3}$	•••	s <sub>NN</sub>

(measured visibilities in blue, autocorrelations in red)

- *S* has *structure* (symmetric, positive semi-definite)
- Instead of statistics of individual visibilities, rather consider statistics of *S* (or *W*) as a whole

#### The Real Wishart Distribution

- It's old news... (Wishart, 1928)
- *W* ~ *W*<sub>N</sub>(*K*, *R*) has *real Wishart distribution*, with *K* degrees of freedom, scale matrix *R* and PDF

$$p(\mathbf{W}) = \frac{|\mathbf{W}|^{(K-N-1)/2}}{2^{NK/2}\Gamma_N(K/2)|\mathbf{R}|^{K/2}} \exp\left[-\frac{1}{2}\operatorname{tr}(\mathbf{R}^{-1}\mathbf{W})\right],$$

where

$$\Gamma_N(K/2) = \pi^{N(N-1)/4} \prod_{i=1}^N \Gamma\left[(K-i+1)/2\right]$$

#### is multivariate gamma function

• Wishart is *sampling distribution* of covariance matrix, just like  $\chi^2$  is sampling distribution of scalar variance

#### When Does This Matter?

Two important limits for radio astronomy:

- The large-sample limit:  $K \gg 1$ Wishart becomes Gaussian, i.e.  $\operatorname{vec} S \sim \mathcal{N}_N(\operatorname{vec} R, R_R)$ with covariance between visibilities  $R_R \propto 1/K$
- **The weak-source limit:** *R* ≈ diagonal Visibilities become uncorrelated
- "Thermal noise" is usually uncorrelated and Gaussian... Wishart therefore more relevant for smaller *K* (fast dump rates, narrow-band channels) and stronger sources — low-frequency instruments?

## The Bartlett Decomposition

- More efficient way to generate random matrices from Wishart distribution than brute force via Gaussian RVs
- If  $W \sim \mathcal{W}_N(\boldsymbol{R}, \boldsymbol{K})$ , then  $W = \boldsymbol{L}\boldsymbol{T}\boldsymbol{T}^T\boldsymbol{L}^T$
- *L* is lower triangular matrix so that  $\mathbf{R} = \mathbf{L}\mathbf{L}^T$  (*Cholesky decomposition* of  $\mathbf{R}$ )
- *T* is random lower triangular matrix with elements

$$T = \begin{bmatrix} \sqrt{c_1} & 0 & 0 & \cdots & 0 \\ n_{21} & \sqrt{c_2} & 0 & \cdots & 0 \\ n_{31} & n_{32} & \sqrt{c_3} & \cdots & 0 \\ \vdots & \vdots & \vdots & \ddots & \vdots \\ n_{N1} & n_{N2} & n_{N3} & \cdots & \sqrt{c_N} \end{bmatrix}$$

where  $c_i \sim \chi^2_{K-i+1}$  and  $n_{ij} \sim \mathcal{N}(0, 1)$ 

## From Real to Complex

- Classical optical coherence theory represents real field by complex analytic signal to simplify the maths
- View *N*-dimensional complex RV  $\mathbf{x} = \operatorname{Re} \mathbf{x} + j \operatorname{Im} \mathbf{x}$ as 2*N*-dimensional real RV  $\mathbf{y} = [\operatorname{Re} \mathbf{x}, \operatorname{Im} \mathbf{x}]^T$
- Real covariance matrix E [yy<sup>H</sup>] has 4N<sup>2</sup> real elements, but complex covariance matrix R = E [xx<sup>H</sup>] only has 2N<sup>2</sup> real elements y is more general than x!
- The missing elements are in the *complementary (pseudo) covariance matrix* (also known as *relation matrix*)

$$\boldsymbol{C} \triangleq \mathrm{E}\left[\boldsymbol{x}\boldsymbol{x}^{T}\right]$$

• *C* is *N* × *N*, complex and *symmetric* (not Hermitian)

## Proper Complex Random Vectors

- Complex RV *x* is *proper* (or *circular*) if *C* = 0 all second-order information contained in *R*
- An equivalent definition is that real covariance matrix have the constrained form

$$\mathbf{E}\left[\boldsymbol{y}\boldsymbol{y}^{H}\right] = \left[\begin{array}{cc} \operatorname{Re}\boldsymbol{R} & -\operatorname{Im}\boldsymbol{R} \\ \operatorname{Im}\boldsymbol{R} & \operatorname{Re}\boldsymbol{R} \end{array}\right]$$

• Yet another way to state it:

$$E\left[\operatorname{Re} x \operatorname{Re} x^{T}\right] = E\left[\operatorname{Im} x \operatorname{Im} x^{T}\right]$$
$$E\left[\operatorname{Re} x \operatorname{Im} x^{T}\right] = -E\left[\operatorname{Re} x \operatorname{Im} x^{T}\right]^{T}$$

• For each element of *x*, real and imaginary parts have same variance and are uncorrelated

# Proper vs Improper

Let x(n) be a discrete-time real stationary bandpass signal.

The following complex signals are *proper*:

- Complex analytic representation of *x*(*n*) (vCZ is OK!)
- Complex envelope of *x*(*n*)
- Discrete-time Fourier transform of *x*(*n*)

The following complex signals are *improper*:

- x(n) itself (because  $C = R \neq 0$ )
- FFT of length-*N* vector *x* taken from x(n) (*asymptotically* proper as  $N \rightarrow \infty$  and FFT  $\rightarrow$  DTFT)

# FX Correlator Example

- F-step does FFT on blocks of 2*N* real voltage samples, producing *N* complex spectral samples per block
- Spectral channels 0 and *N* are always real and therefore improper another reason to discard these...
- The rest of the channels are asymptotically proper as  $M \rightarrow \infty$  check propriety for small *M* though
- Polyphase filterbank in front of FFT does not change this

## **Complex Gaussian Distribution**

- The *N*-dimensional RV *x* ~ N<sub>N</sub><sup>C</sup>(*m*, *R*) has a *complex Gaussian distribution* if the corresponding 2*N*-dimensional RV *y* is Gaussian and *x* is proper
- PDF is straightforward extension of real one:

$$p(\mathbf{x}) = \frac{1}{\pi^N |\mathbf{R}|} \exp\left[-(\mathbf{x} - \mathbf{m})^H \mathbf{R}^{-1} (\mathbf{x} - \mathbf{m})\right]$$

(only valid if *x* is proper!)

• Has maximum entropy among distributions with same *m* and *R* 

#### The Complex Wishart Distribution

- Let  $\mathbf{x} \sim \mathcal{N}_N^{\mathbb{C}}(\mathbf{0}, \mathbf{R})$  and form scatter matrix  $\mathbf{W} = \sum_{k=1}^K \mathbf{x}_k \mathbf{x}_k^H$
- $W \sim \mathcal{W}_N^{\mathbb{C}}(K, \mathbf{R})$  has complex Wishart distribution with PDF

$$p(\mathbf{W}) = \frac{|\mathbf{W}|^{K-N}}{\tilde{\Gamma}_N(K)|\mathbf{R}|^K} \exp\left[-\operatorname{tr}(\mathbf{R}^{-1}\mathbf{W})\right],$$

where

$$\tilde{\Gamma}_N(K) = \pi^{N(N-1)/2} \prod_{i=1}^N \Gamma(K-i+1)$$

is complex multivariate gamma function

- Similar properties to real Wishart, e.g. if *W* ~ *W*<sup>ℂ</sup><sub>N</sub>(*K*, *R*), then *AWA<sup>H</sup>* ~ *W*<sup>ℂ</sup><sub>N</sub>(*K*, *ARA<sup>H</sup>*) for any full-rank matrix *A*
- Even nicer: recursive formula to write down moments of any order (Letac & Massam, 2004)

#### Some Moments of Complex Wishart

- Let elements of  $S = \frac{1}{K}W$  be  $s_{ij}$  and those of R be  $r_{ij}$
- Mean visibility:  $E[s_{ij}] = r_{ij}$
- Correlation between arbitrary visibilities:

$$\mathbf{E}[s_{ij}s_{kl}^*] = r_{ij}r_{kl}^* + \frac{1}{K}r_{ik}r_{jl}^*$$

Covariance between arbitrary visibilities:

$$E[(s_{ij} - r_{ij})(s_{kl} - r_{kl})^*] = \frac{1}{K}r_{ik}r_{jl}^*$$

• Variance of visibility:

$$\mathrm{E}\big[|s_{ij}-r_{ij}|^2\big]=\frac{1}{K}r_{ii}r_{jj}$$

 $\rightarrow$  Thermal noise level!

#### **Closure Statistics**

- Closure phase is angle of *triple product* s<sub>ij</sub>s<sub>jk</sub>s<sub>ki</sub>
- Closure amplitude is  $|(s_{ij}s_{kl})/(s_{ik}s_{jl})|$
- Wishart moments are useful to characterise these
- E.g. Mean of triple product (autocorrelations in red):

$$E[s_{ij}s_{jk}s_{ki}] = r_{ij}r_{jk}r_{ki} + \frac{1}{K}\left(r_{ii}r_{jk}r_{kj} + r_{jj}r_{ik}r_{ki} + r_{kk}r_{ij}r_{ji}\right) + \frac{1}{K^2}\left(r_{ii}r_{jj}r_{kk} + r_{ji}r_{kj}r_{ik}\right)$$

• Variance of triple product straightforward but tedious to write down...

# Applications

- Good theoretical model
- Fancy thermal noise generator (full polarisation!)
- Detect deviation from visibility model via likelihood test
- MAP estimation: we have *p*(*S*|*R*), turn it around to form *p*(*R*|*S*) via *P*(*R*), for which convenient conjugate prior is available

## Appendix: Wishart Generator

```
import numpy as np
import scipy.stats
def complex_wishart(R, df, size=1):
    """Generate random matrices from complex Wishart distribution.
   R --- True covariance matrix, of shape (dim, dim)
    df --- Degrees of freedom
    size --- Number of random matrices to generate
    Returns sequence of random matrices as complex array S, of shape (size, dim, dim)
    ......
    # Obtain dimension of covariance matrix
   \dim = R.shape[0]
    # Pre-allocate the sequence of output matrices
   S = np.zeros((size, dim, dim), dtype=np.complex128)
    # Do Cholesky decomposition of covariance matrix
   L = np.linalg.cholesky(R)
    # Load diagonal elements with square root of chi—square variates (a.k.a. chi variates)
    for n in xrange(dim):
       S[:, n, n] = scipy.stats.chi.rvs(2 * (df - n), size=size)
    # Obtain all lower triangle indices
    lower = np.tri(dim, k=-1).ravel() > 0
   nlow = len(lower)
    for m in xrange(size):
        # Load lower triangle with standard complex Gaussian RVs
        S[m].ravel()[lower] = np.random.randn(nlow) + 1.0 j * np.random.randn(nlow)
        # Transform Bartlett decomposition factor by Cholesky factor
       LT = np.dot(L, S[m])
       # Finally form Hermitian matrix
       S[m] = 0.5 * np.dot(LT, LT.conj().transpose())
    return S
```