

Antenna Array Modeling in the Netherlands

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Outline

- infinite arrays: extension of classical approach
 - acceleration of MoM matrix fill
 - integration of FSS and phased array antennas
- finite arrays:
 - eigenvalue analysis
 - embedding approach (only applied to EBG structures, not in this presentation)

Infinite Arrays: Classical Approach

- at interfaces:
 - modal expansion (global expansion functions)
 - EFIE or MFIE (local expansion functions)
- Green's functions: spectral (Floquet) analysis \Rightarrow transmission-line equations
- handle layered medium in Green's function
- numerical solution: method of moments (MoM)
problem: matrix fill
- scattering operator for composite structures

Acceleration of MoM Matrix Fill

Bart Morsink, October 17, 2005
with Gertjan van Werkhoven,
sponsored by Thales Nederland

General form of MoM matrix elements:

$$A_{v,r} = \sum_{j,\alpha} G^{(\alpha)}(k_j) \gamma_{v,r}^{(\alpha)}(\vec{k}_j)$$

where

$$\vec{k}_j = \vec{k}_j^{inc} + \vec{R}_j \text{ with } \vec{R}_j \text{ a reciprocal lattice vector}$$

$$G^{(\alpha)}(k_t) = \text{solution of 1D transmission - line equation}$$

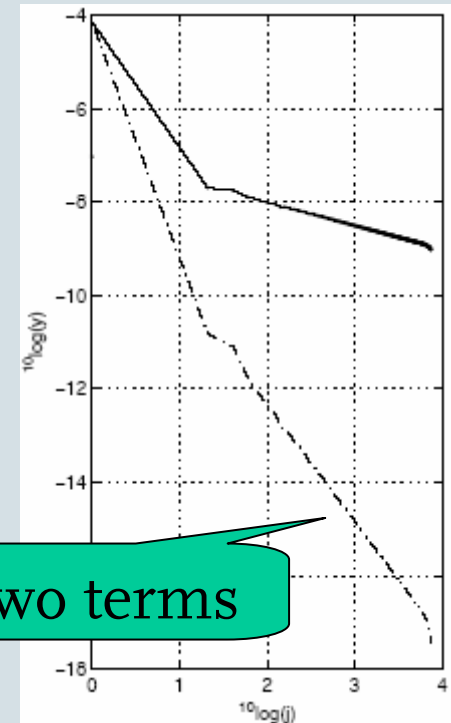
$$\gamma_{v,r}^{(\alpha)}(\vec{k}_t) = \begin{cases} [j\vec{k}_t \cdot \vec{g}_r^*(\vec{k}_t)][j\vec{k}_t \cdot \vec{g}_r(\vec{k}_t)] & \text{if } \alpha = 1 \\ \vec{g}_r^*(\vec{k}_t) \cdot \vec{g}_r(\vec{k}_t) & \text{if } \alpha = 2 \end{cases}$$

Asymptotic expansion:

$$\hat{G}^{(\alpha)}(k_t) = c_1^{(\alpha)} k_t^{-1} + c_2^{(\alpha)} k_t^{-3} + c_3^{(\alpha)} k_t^{-5} + \dots$$

Kummer transformation:

$$A_{v,r} = \sum_{j,\alpha} [G^{(\alpha)}(k_j) - \hat{G}^{(\alpha)}(k_j)] \gamma_{v,r}^{(\alpha)}(\vec{k}_j) + \sum_{j,\alpha} \hat{G}^{(\alpha)}(k_j) \gamma_{v,r}^{(\alpha)}(\vec{k}_j)$$



Asymptotic series: Ewald's transformation

$$\frac{1}{k_t^n} = \frac{2\lambda}{\Gamma(n/2)} \int_0^\infty \tau^{\lambda-1} \exp(-k_t^2 \tau^{2\lambda}) d\tau$$

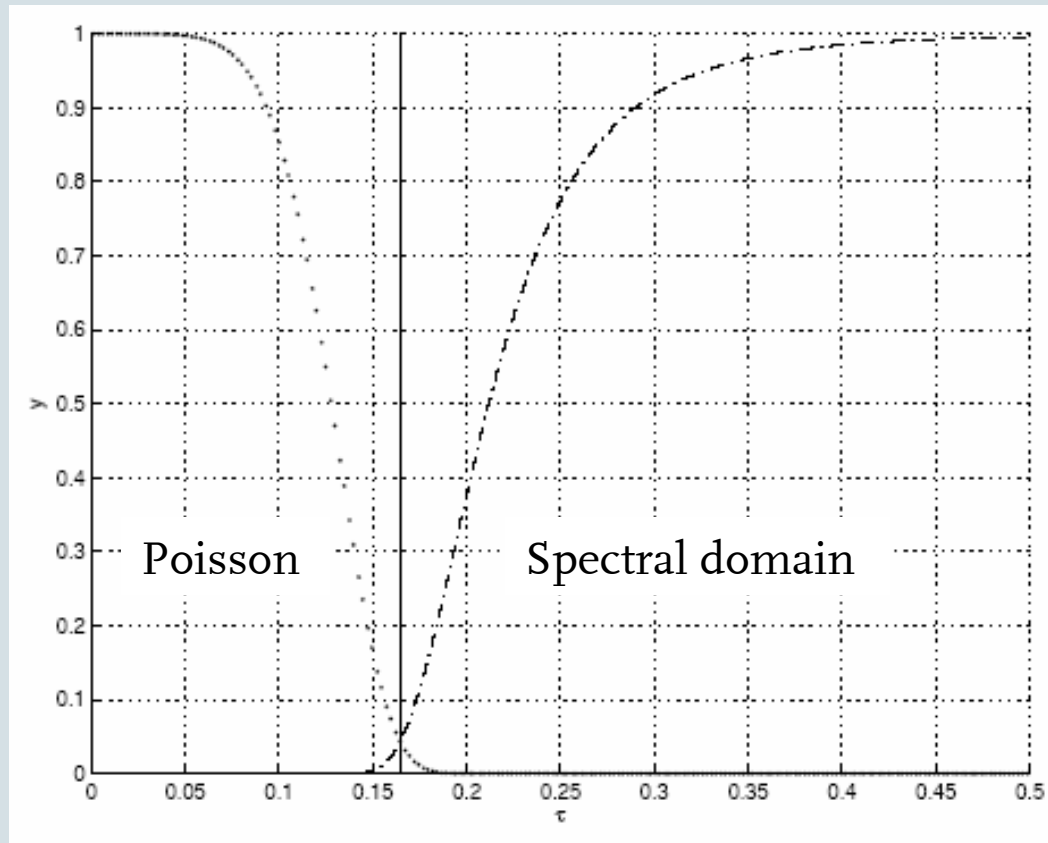
results in

$$\hat{A}^{(\alpha)}(k_t) = \frac{2\lambda}{\sqrt{\pi}} \int_0^\infty \tau^{\lambda-1} \sum_j (c_1^{(\alpha)} + 2\tau^{2\lambda} c_2^{(\alpha)}) \exp(-k_{t,j}^2 \tau^{2\lambda}) \gamma_{v,r}^{(\alpha\alpha)}(\vec{k}_j) d\tau$$

Large argument: exponential convergence

Small argument: Poisson summation

Transition point: leading-order term



Success depends on the determination of

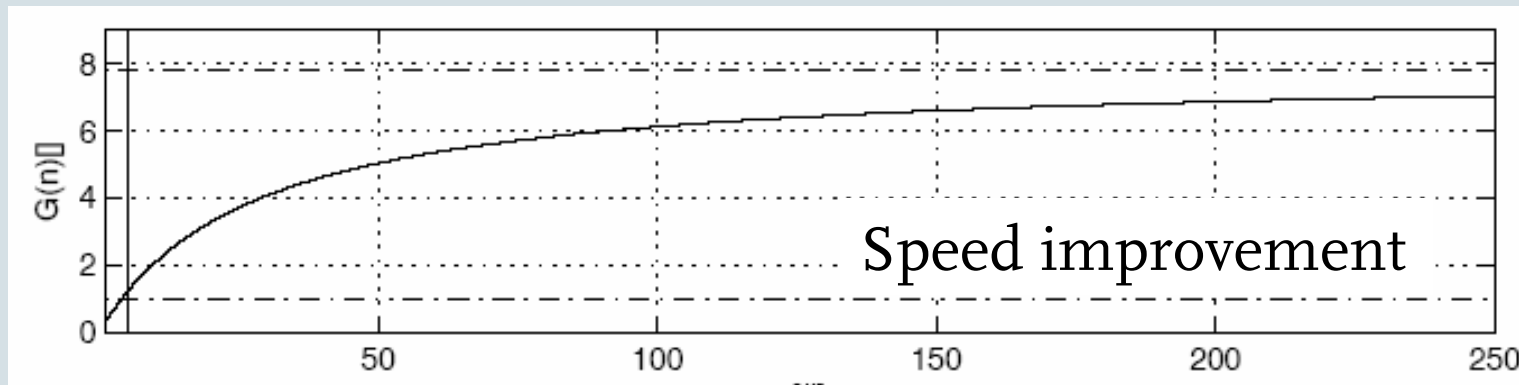
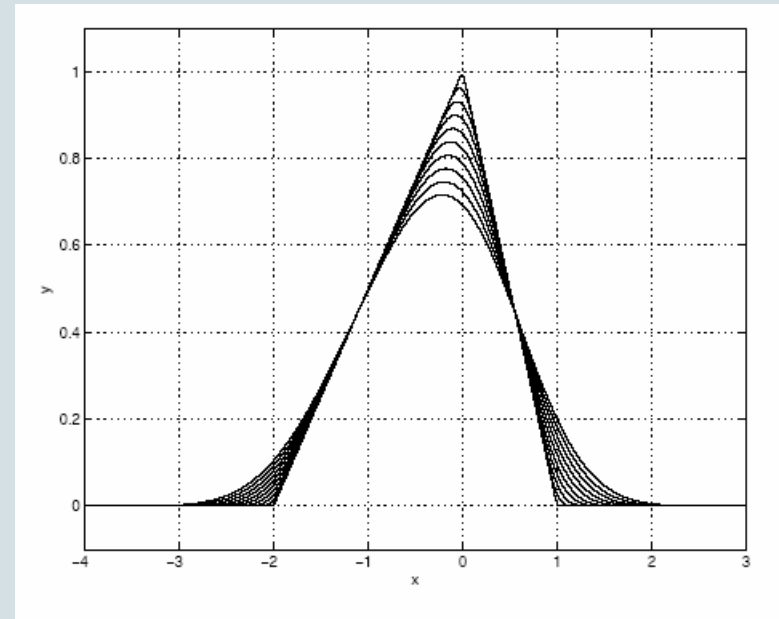
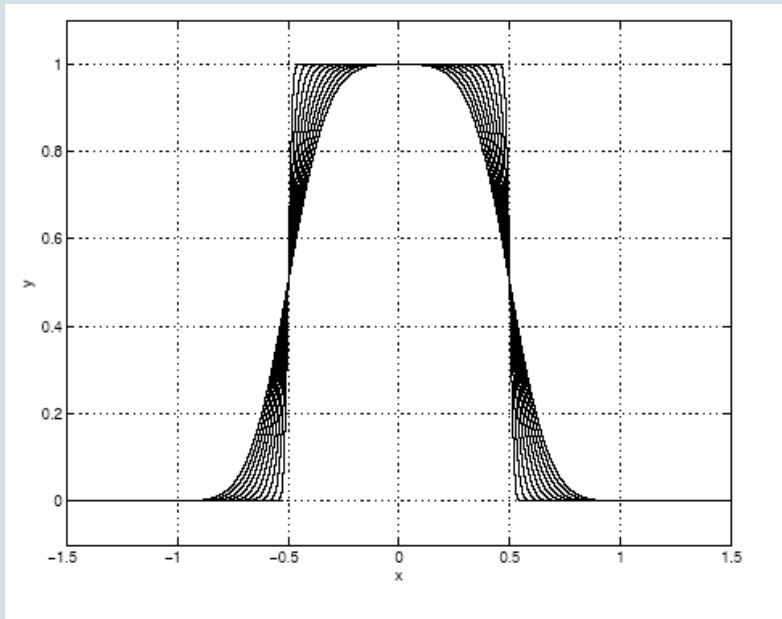
$$\gamma_{v,r}^{(\alpha)}(\vec{\rho}, \tau) = F^{-1}\{\exp(-k_t^2 \tau^{2\lambda}) \hat{\gamma}_{v,r}^{(\alpha)}(\vec{k}_t)\}$$

Can be identified as solution of **heat equation**

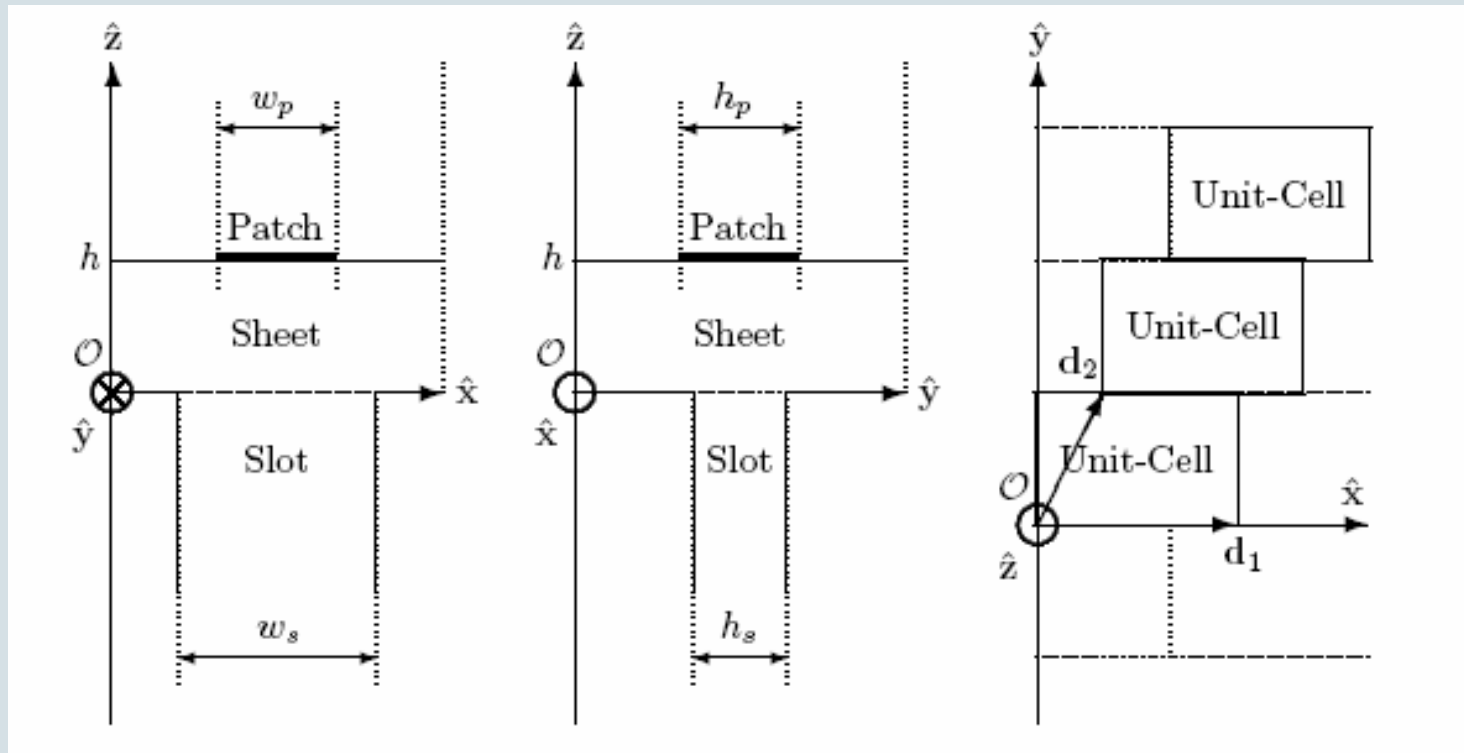
Example: **rooftop function**

- product of pulse and triangular function
- regularized function also as product

Recent development: **RWG functions**



Example: open-ended waveguide with patch

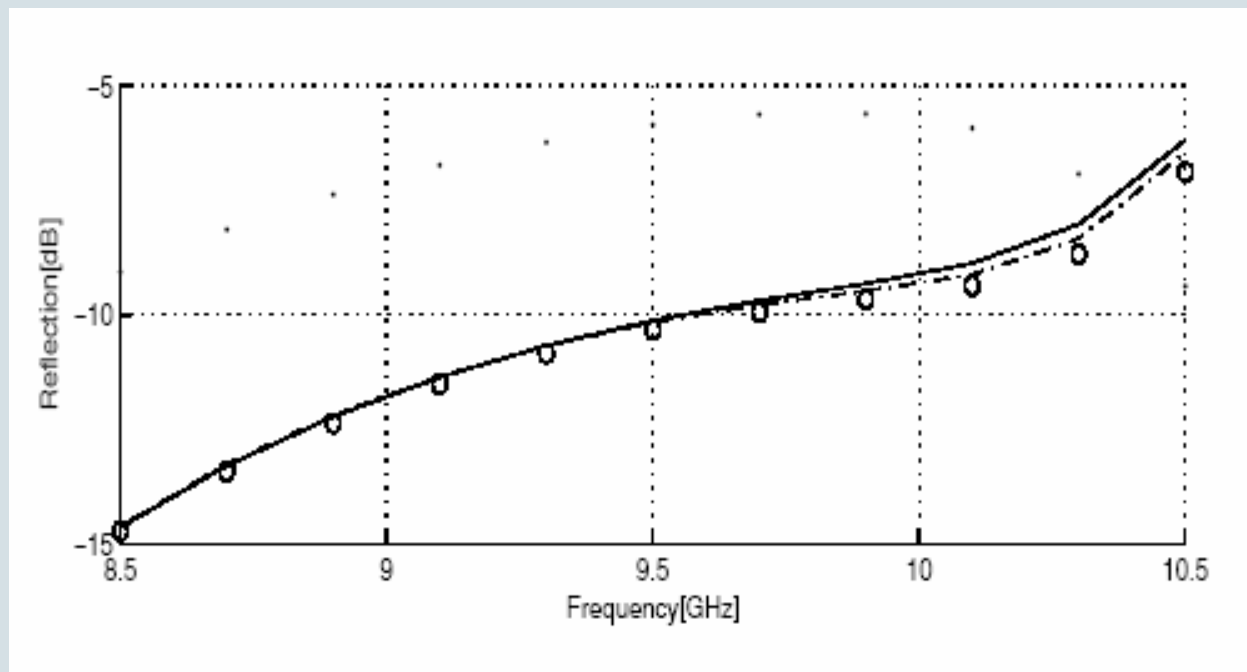


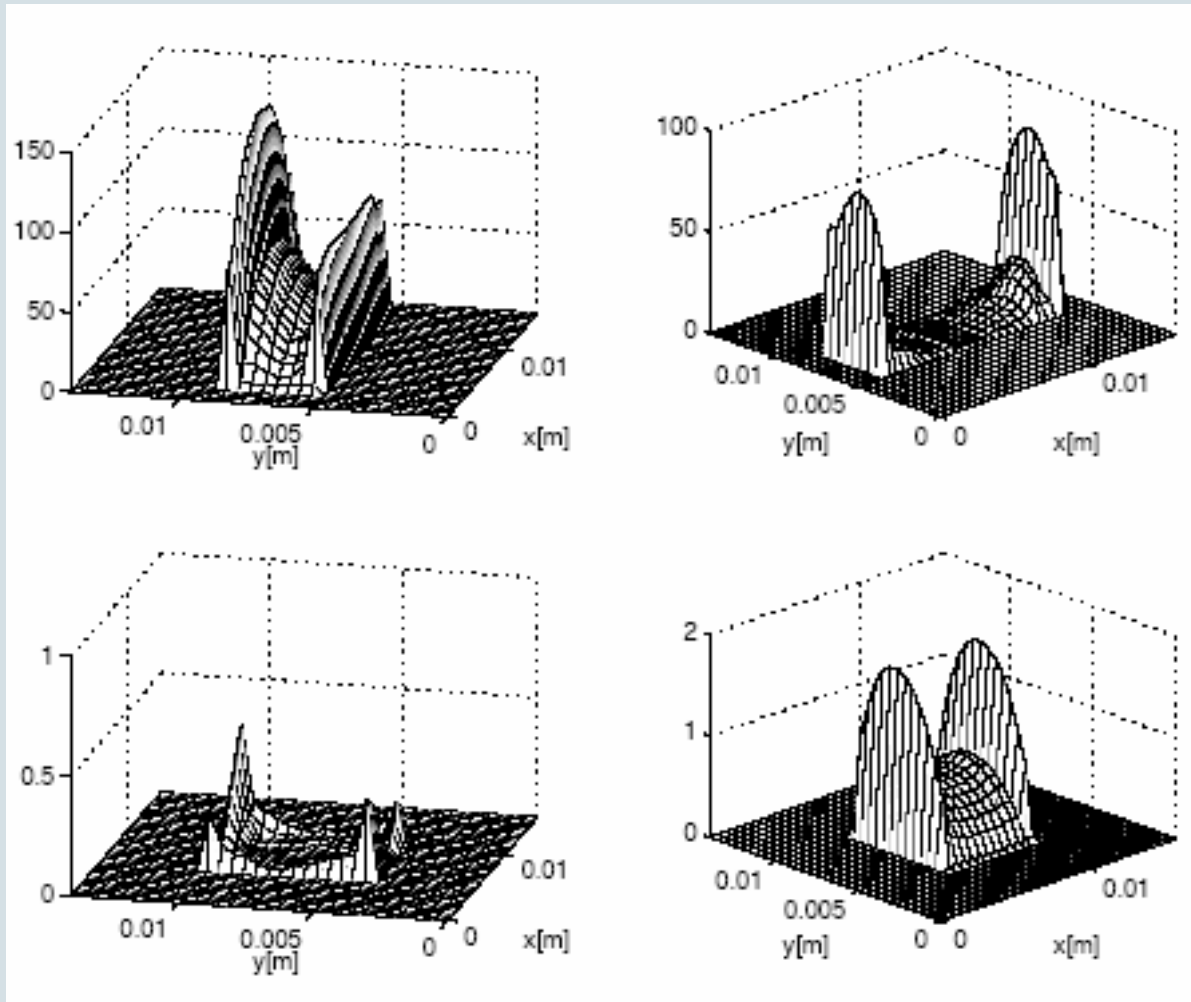
Side view x

Side view y

Top view

Power reflection coefficient





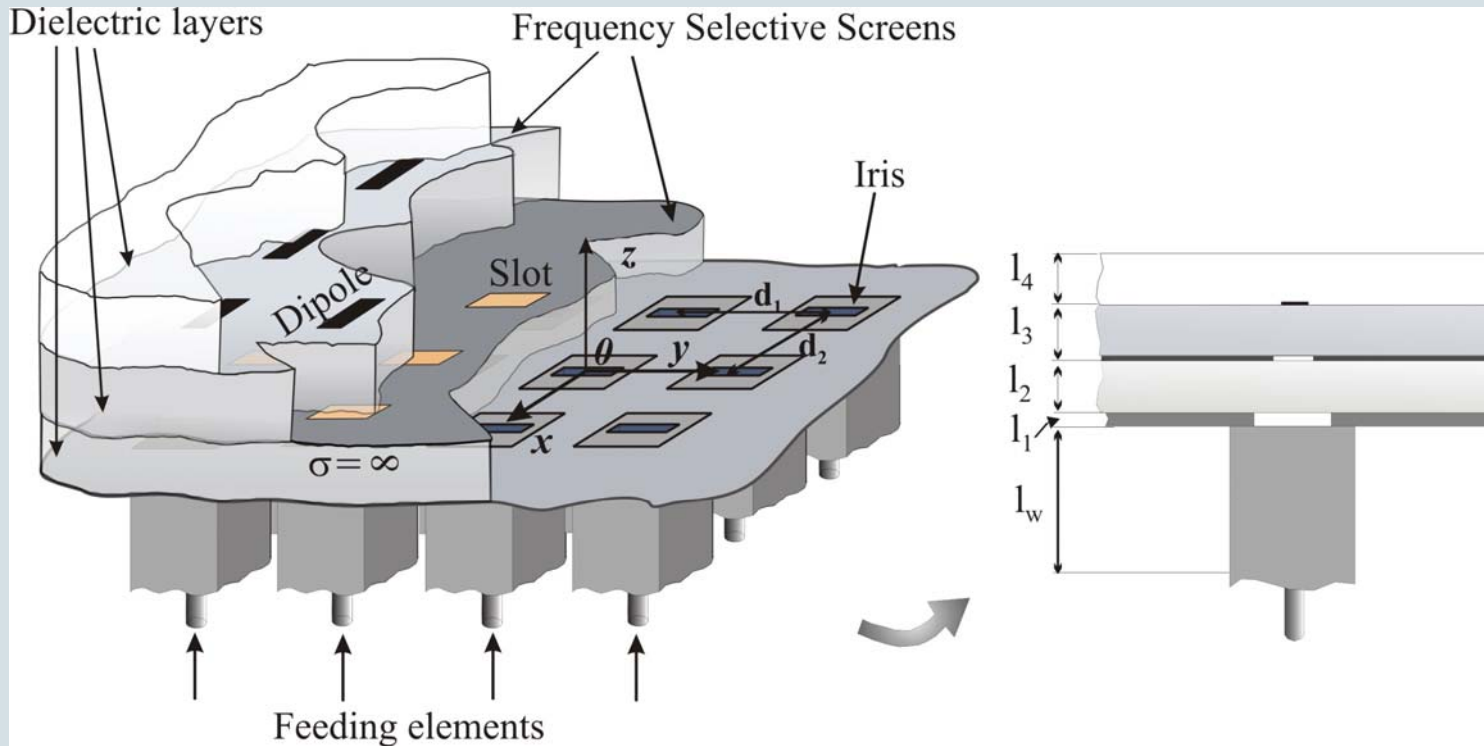
M_x and M_y
on aperture

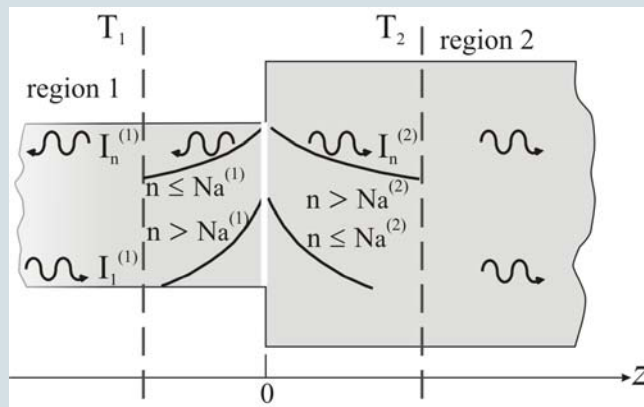
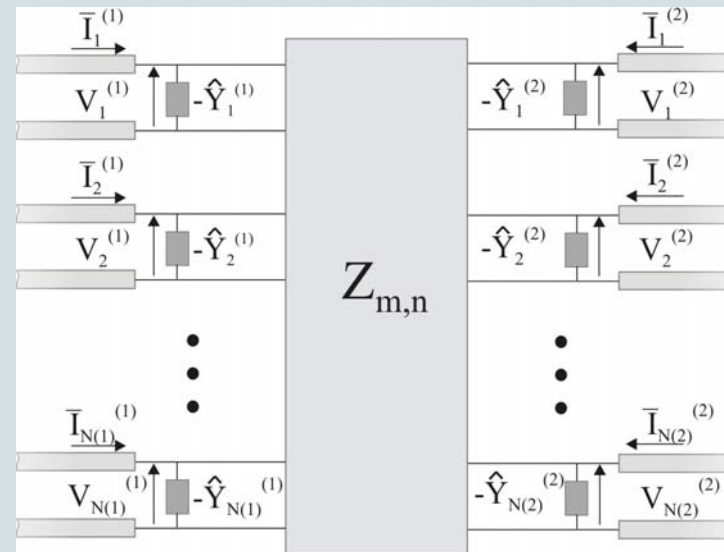
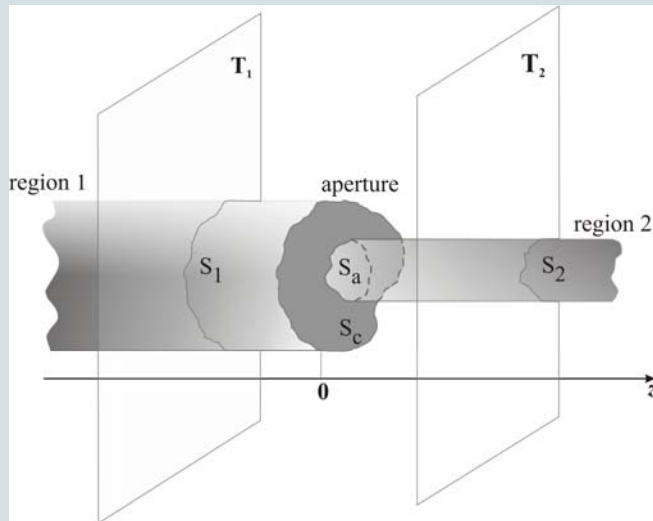
J_x and J_y
on patch

Integration of Waveguide Array and FSS

Stefania Monni, June 27, 2005
with Giampiero Gerini and
Andrea Neto, TNO FEL

Structure to be analyzed/designed

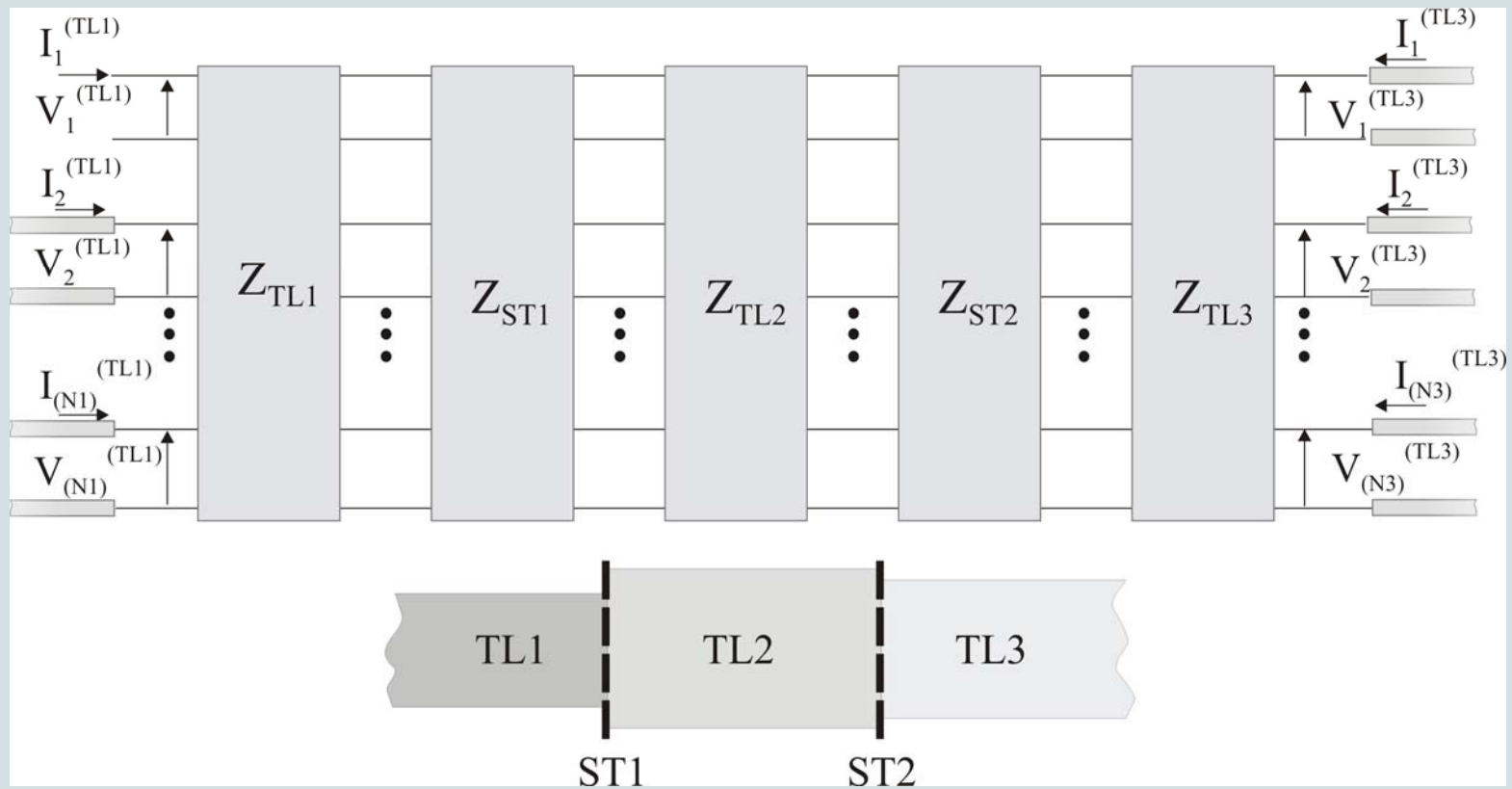


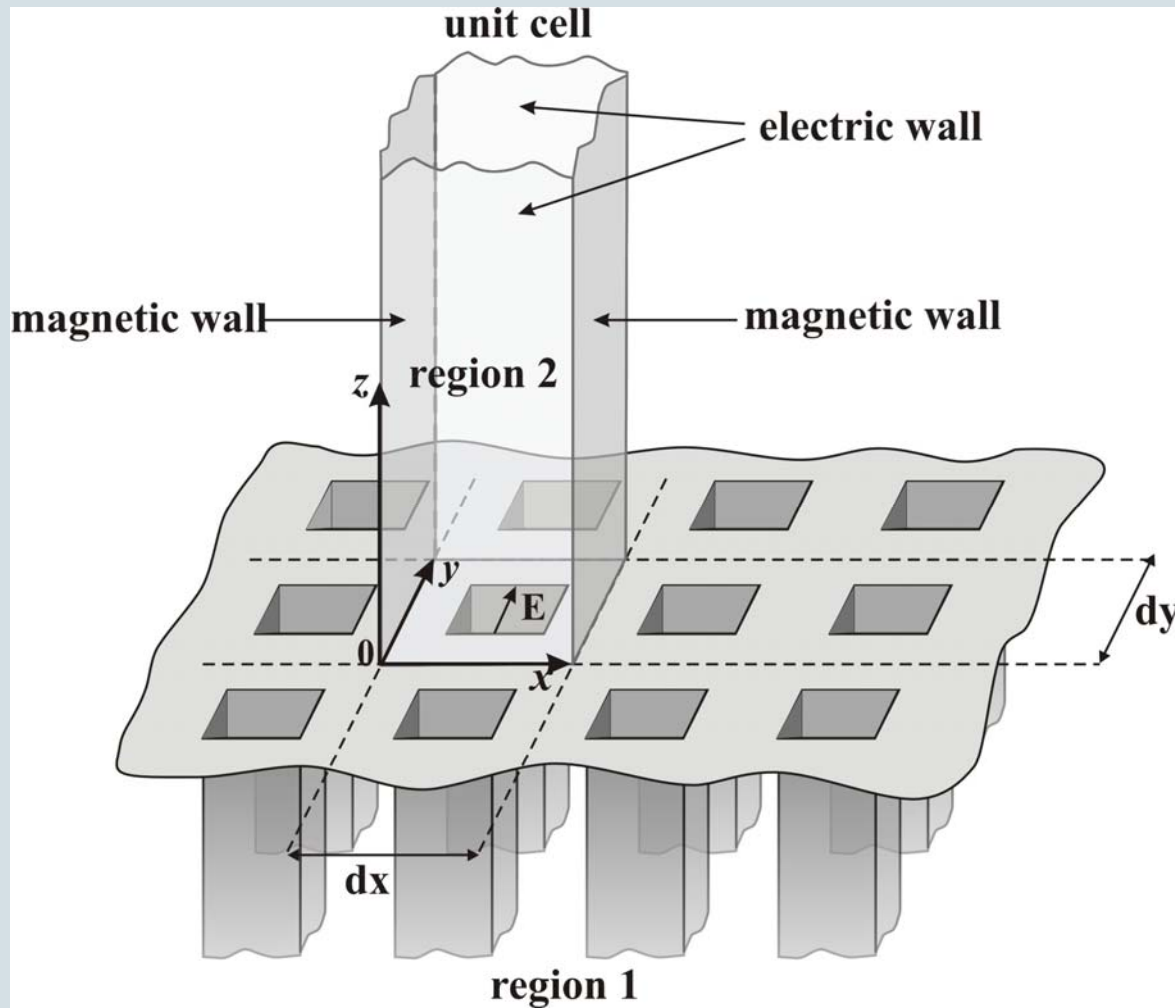


Procedure:

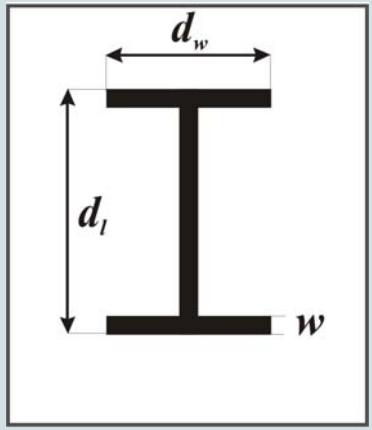
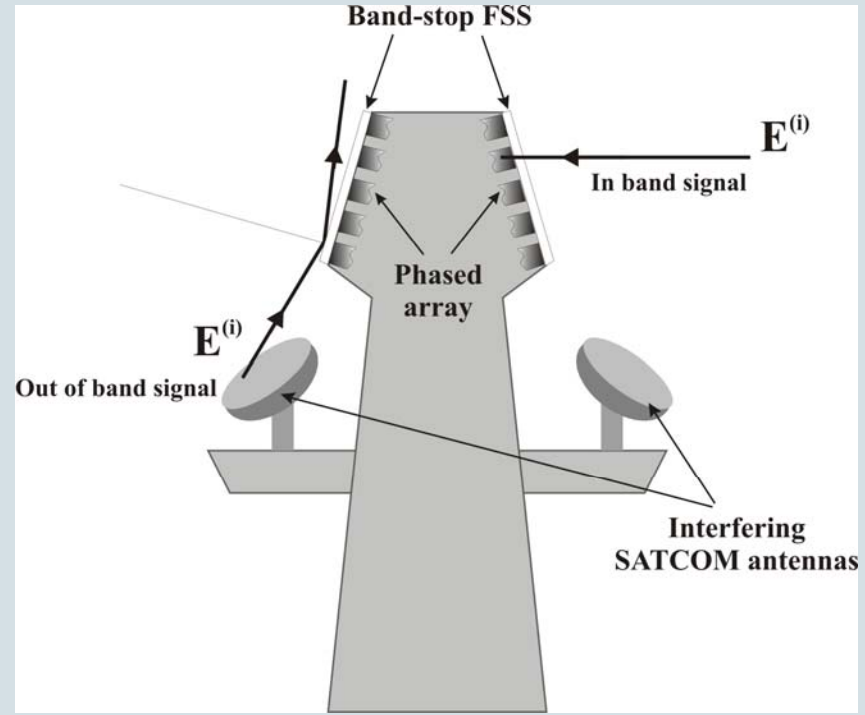
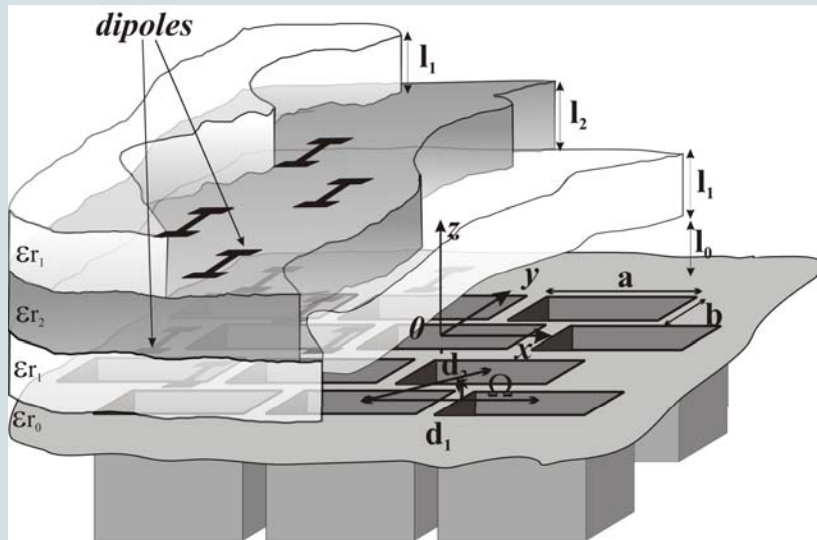
- identify accessible modes
- boundary conditions at $z=0$
- identify multimode network

Multiple sections

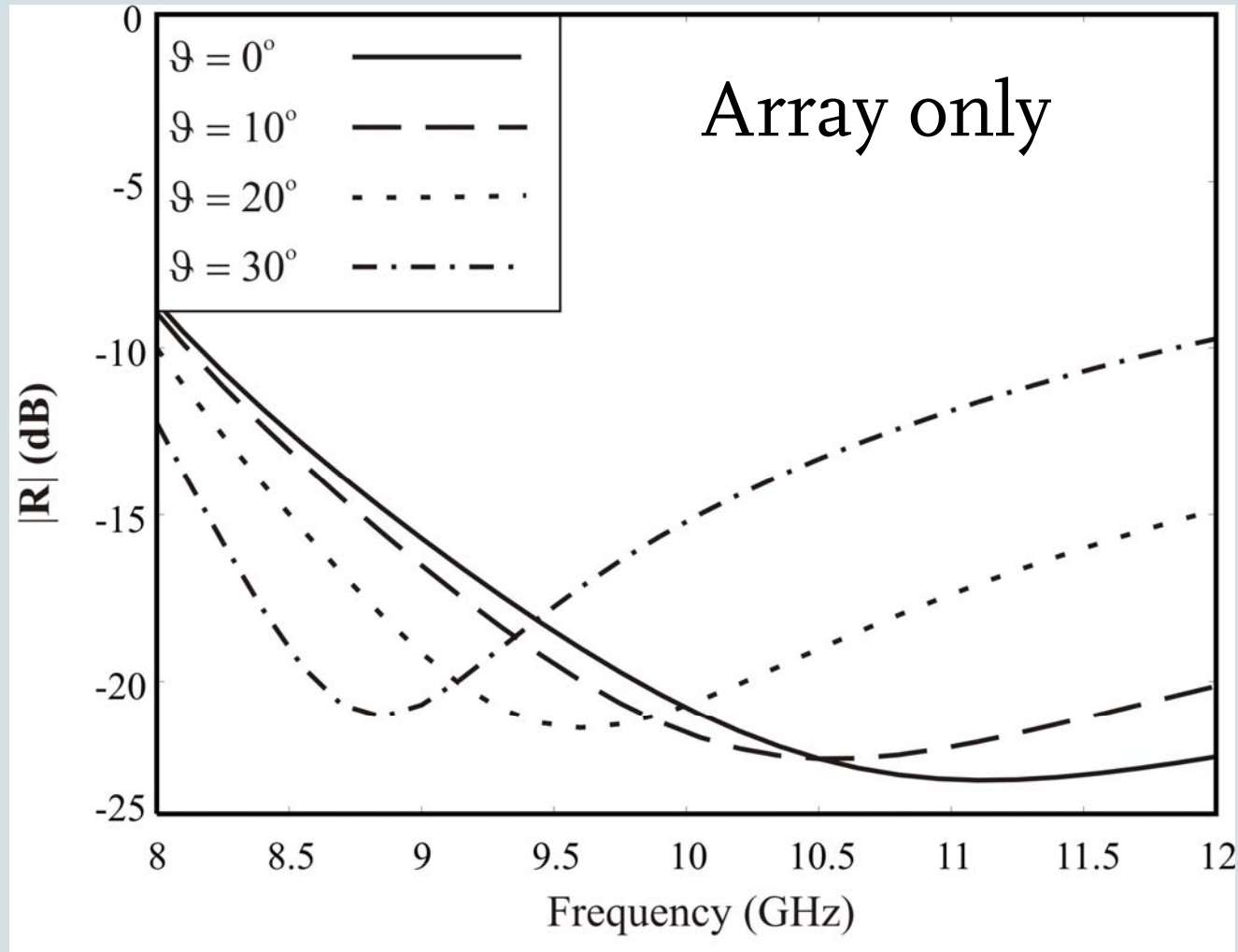


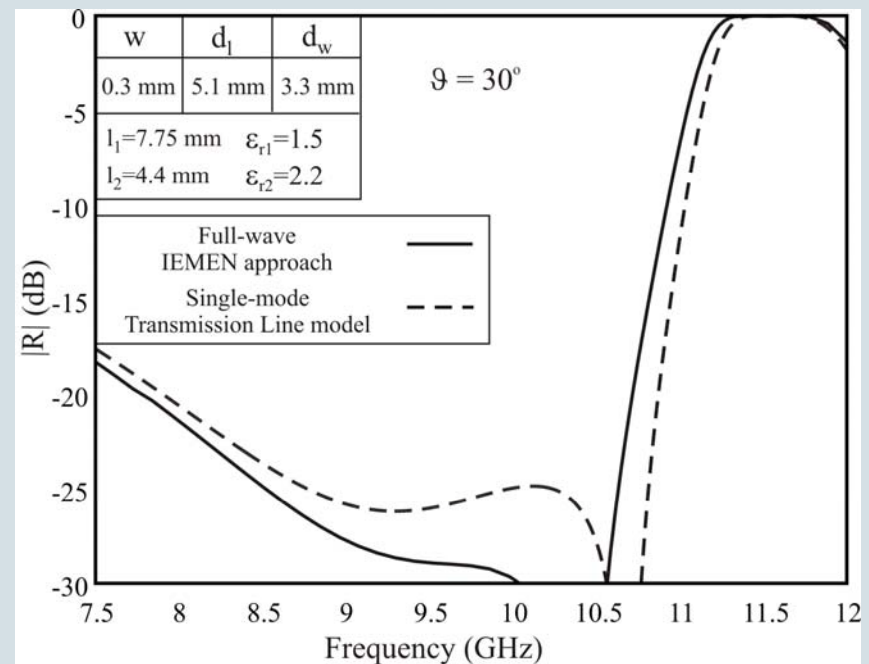
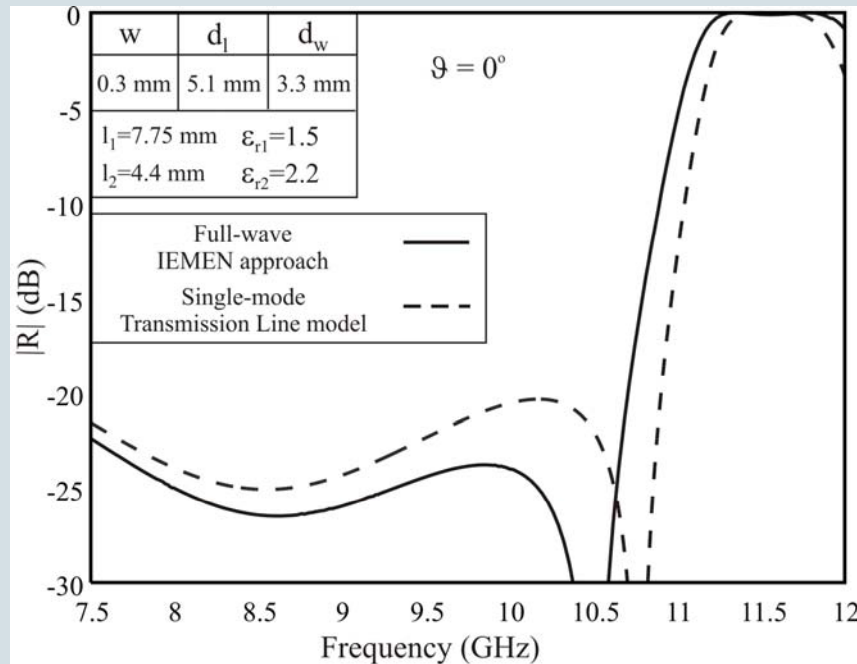


Boundary conditions for infinite half space

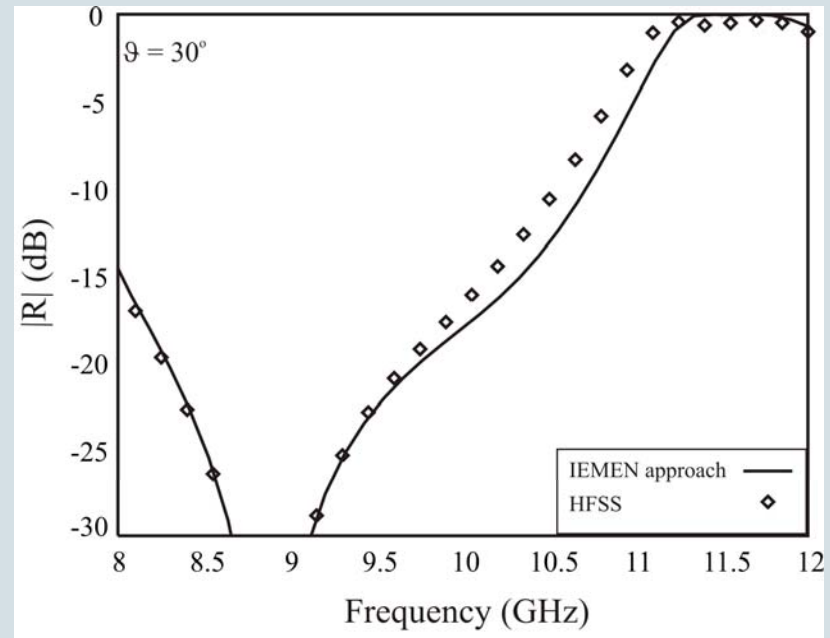
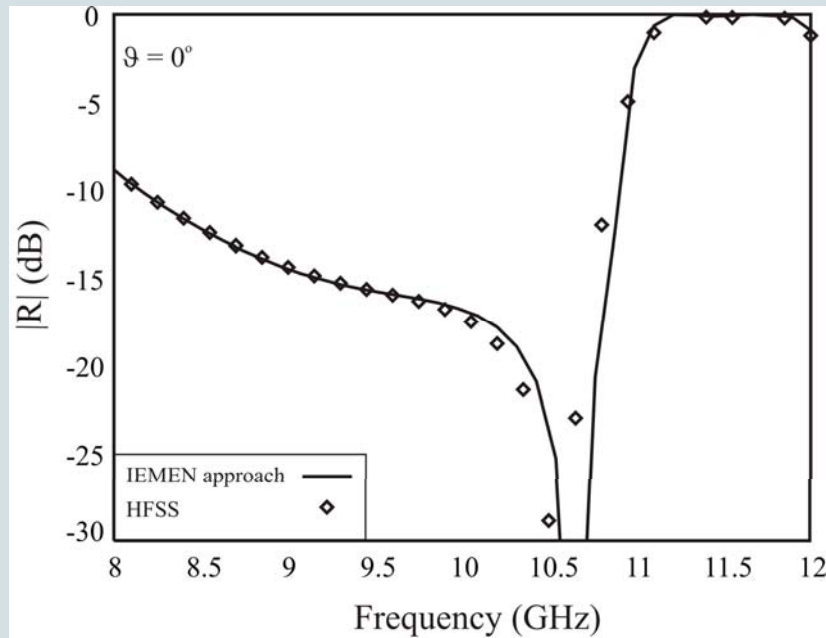


Example for synthesis: interference between MFR array and satellite communication antenna

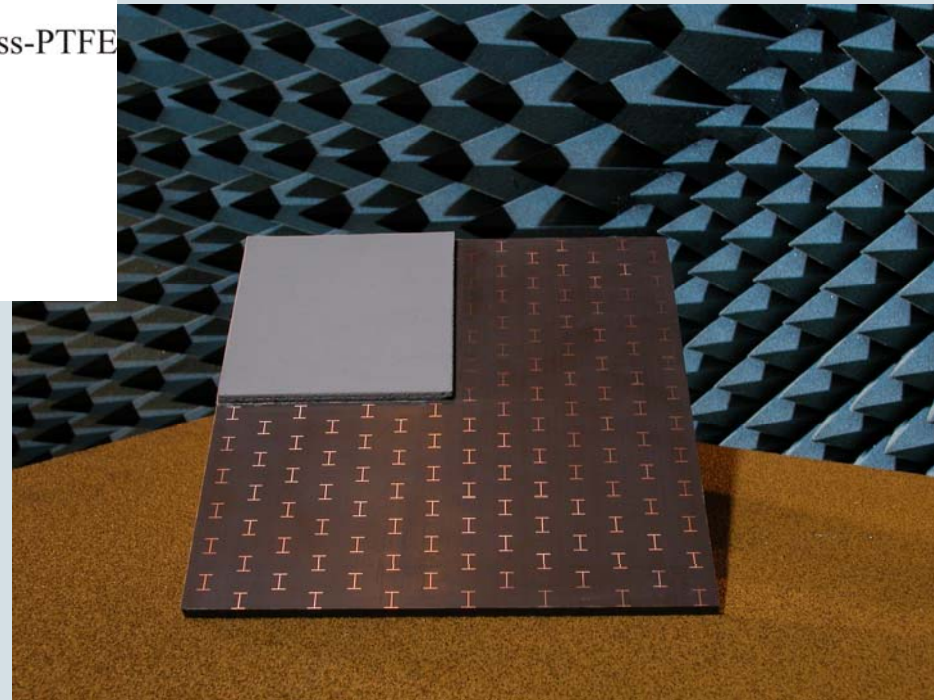
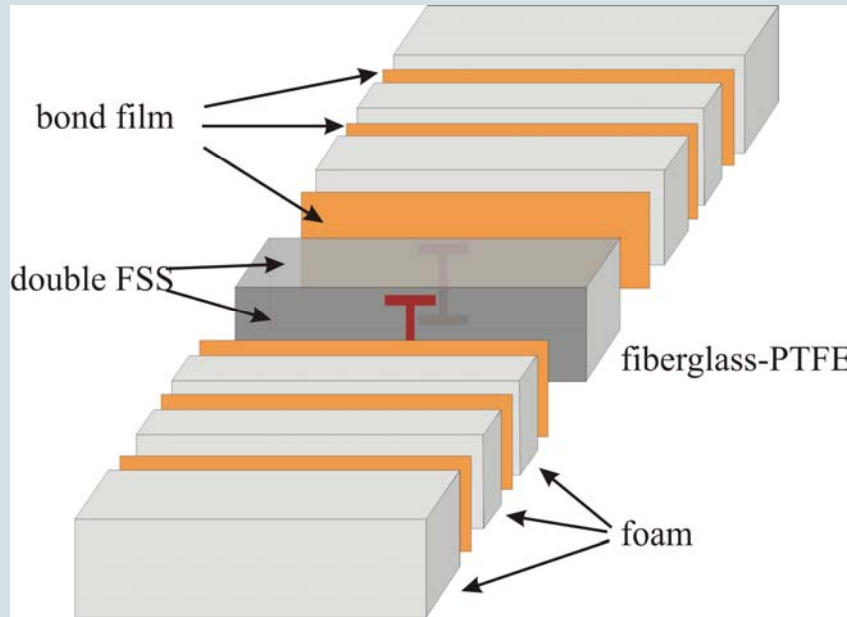




Single-mode design and full-wave model



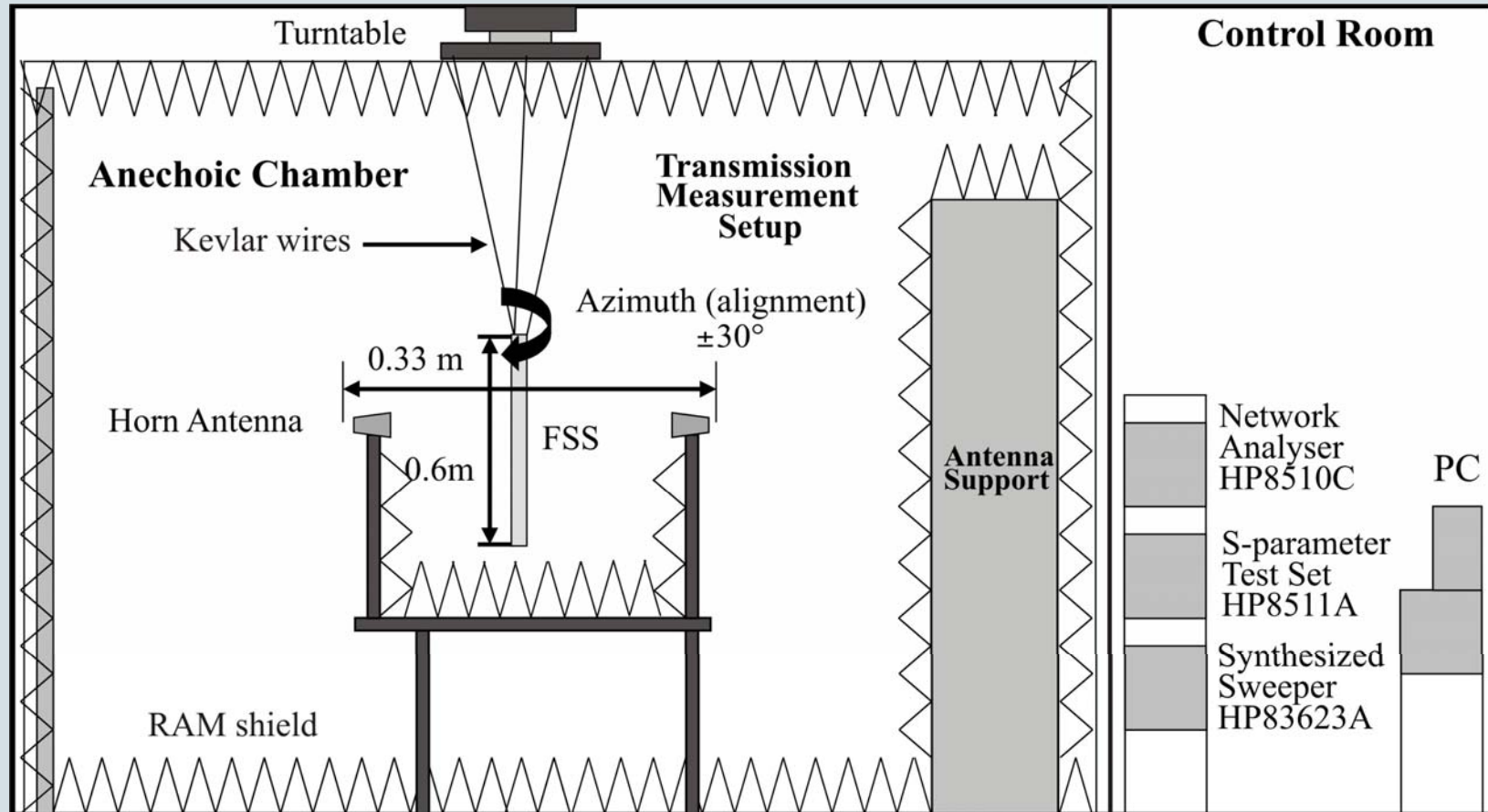
IEMEN method and HFSS

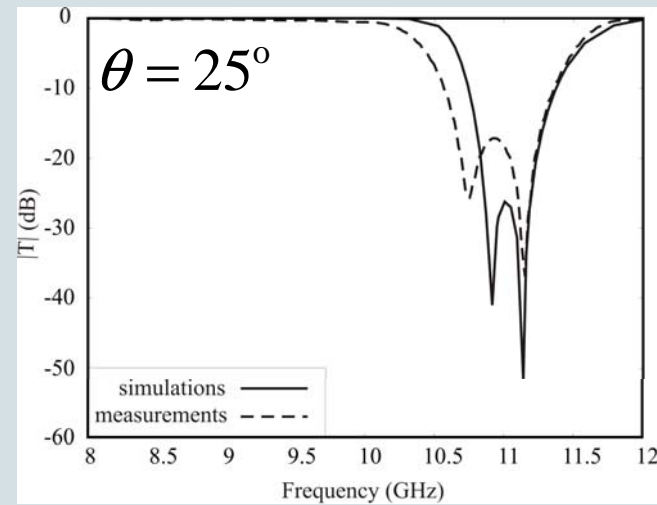
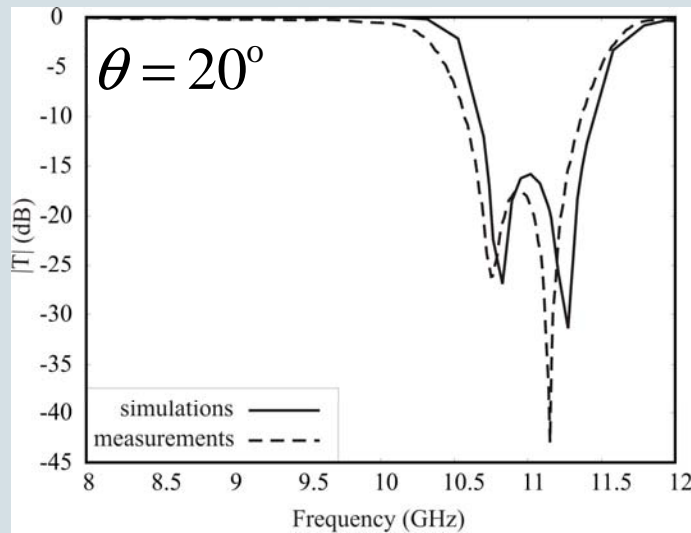
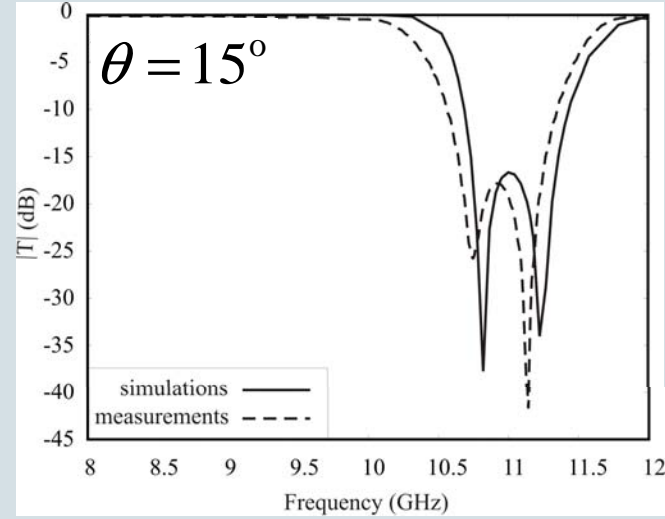
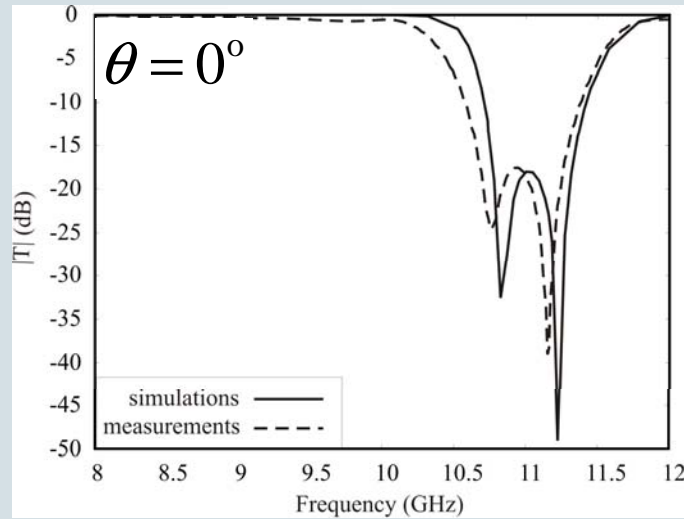


Realization



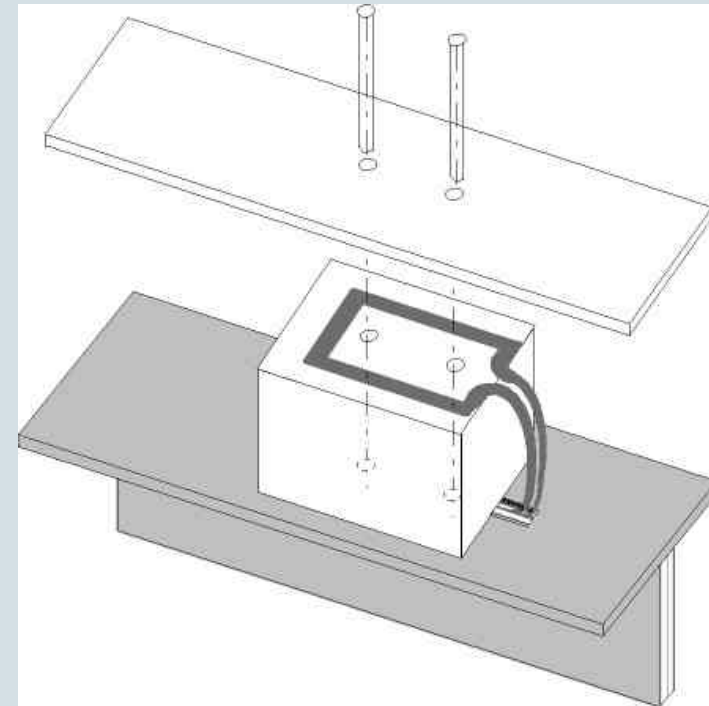
Transmission
measurement





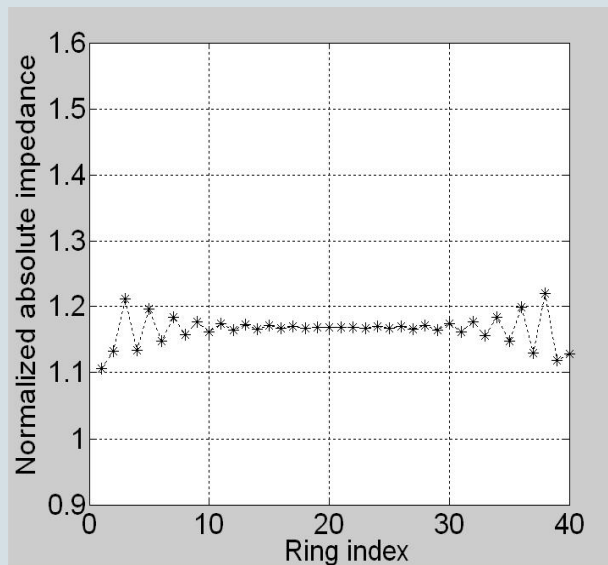
Finite Antenna Arrays: Eigencurrent Approach

Dave Bekers, December 13, 2004
with Stef van Eijndhoven, Fons van
der Ven en Peter Paul Borsboom
Sponsored by Thales Nederland

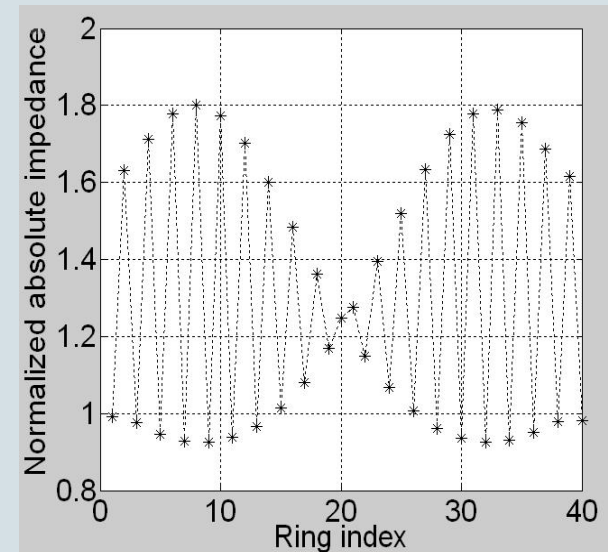


Phased array radar systems:
1000 elements in an array

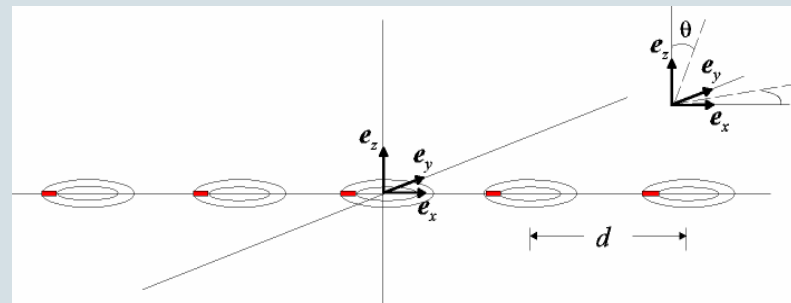
Desired in arrays:
Small variation of impedance



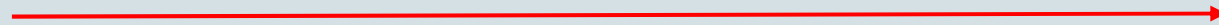
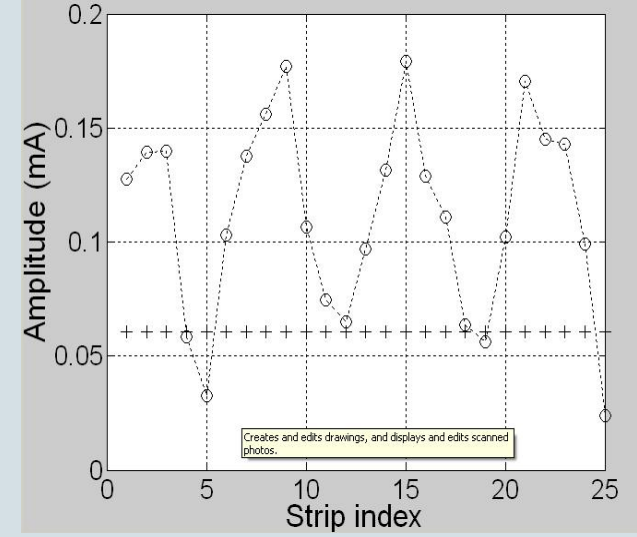
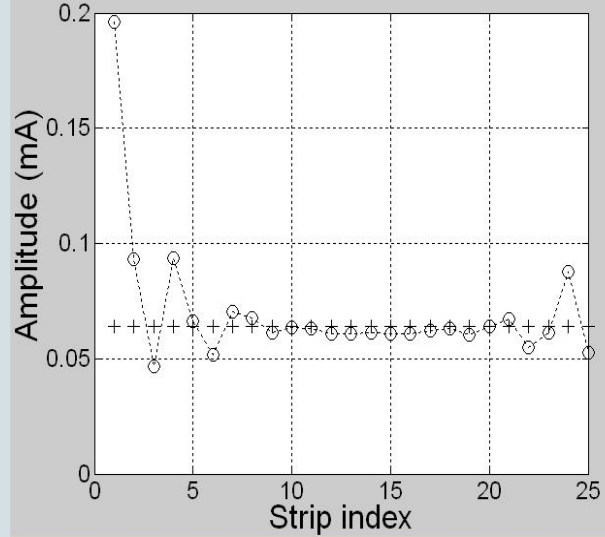
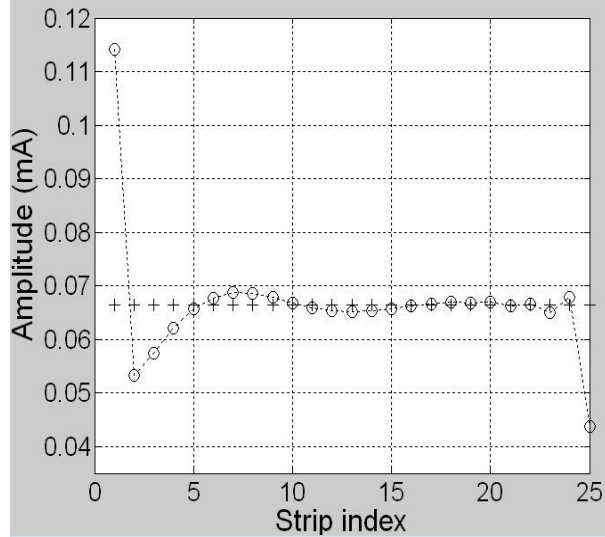
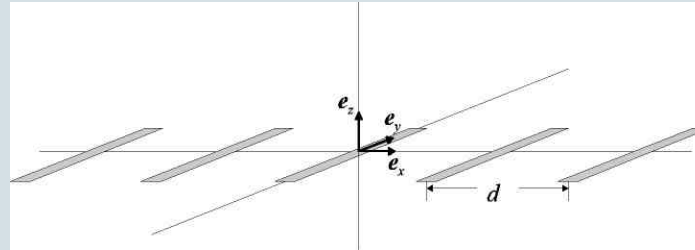
Not desired in arrays:
Large variation of impedance



Line array of 40 rings



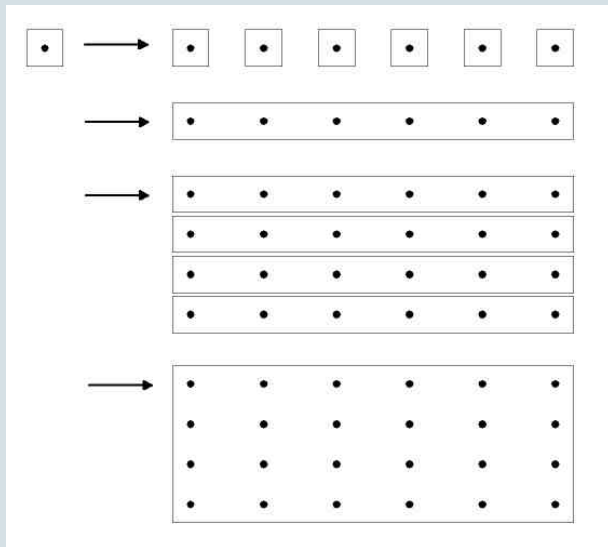
Line array of 25 strips



Decreasing frequency

General Idea

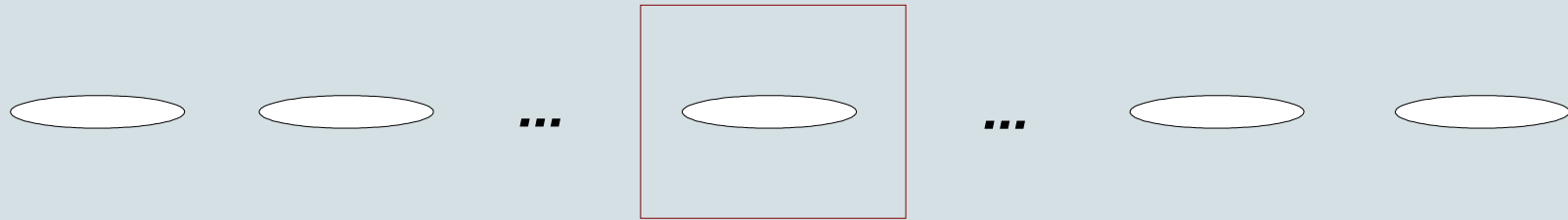
- Characteristics of arrays and their elements is reflected in their “eigenstates”
- Modularity: hierarchy of “subarrays”



“Brute force” method: 40 elements,
30 piecewise functions per element \Rightarrow
matrix size 1200×1200

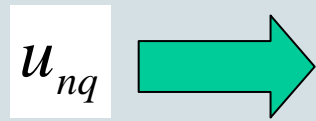
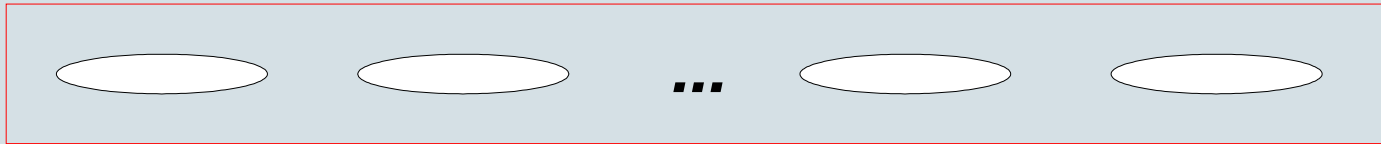
New approach: 30 piecewise functions
on element \Rightarrow matrix size 30×30
40 elements, 2 coupling eigencurrents
 \Rightarrow matrix size 80×80

General Idea



$$V_n, u_n$$

$$V_{nq} = V_n (1 + \epsilon_{nq})$$



$$u_{nq}$$

$$u_n$$

$$\alpha_{n1}$$

$$u_n$$

$$\alpha_{n2}$$

$$u_n$$

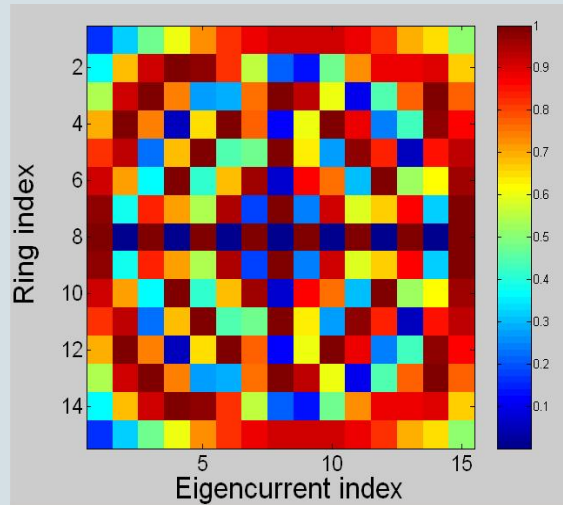
$$\alpha_{n(Q-1)}$$

$$u_n$$

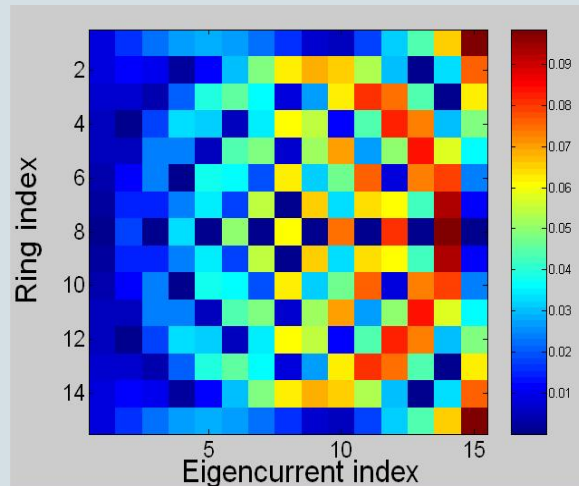
$$\alpha_{nQ}$$

+ perturbation

Line array of 15 rings with 2 groups of eigencurrents

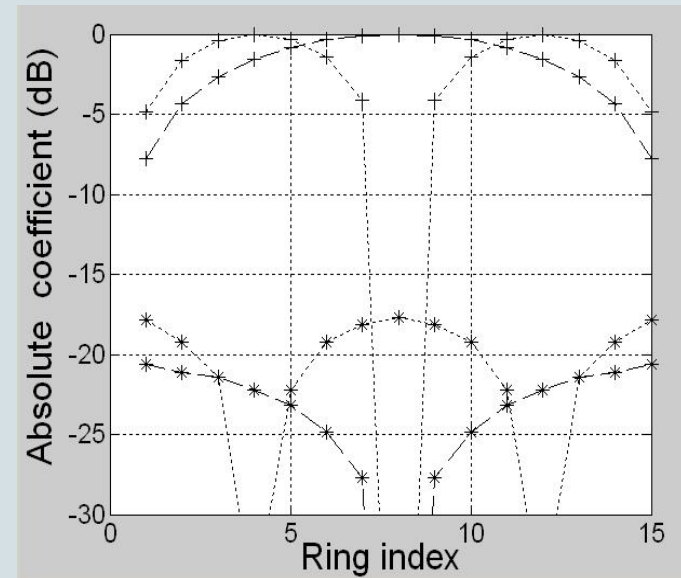


First group:
dominant coefficients



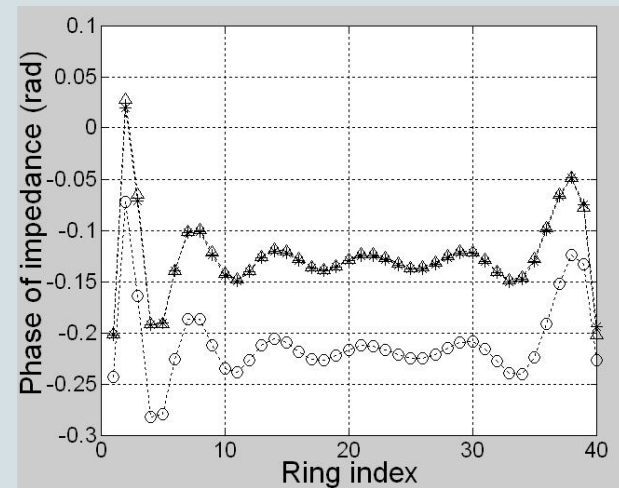
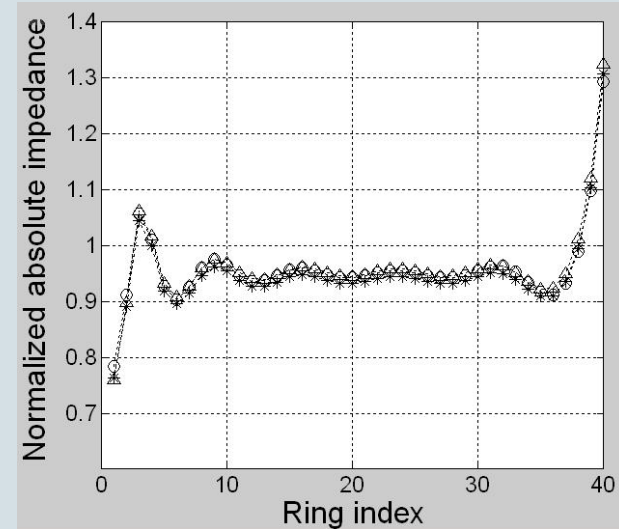
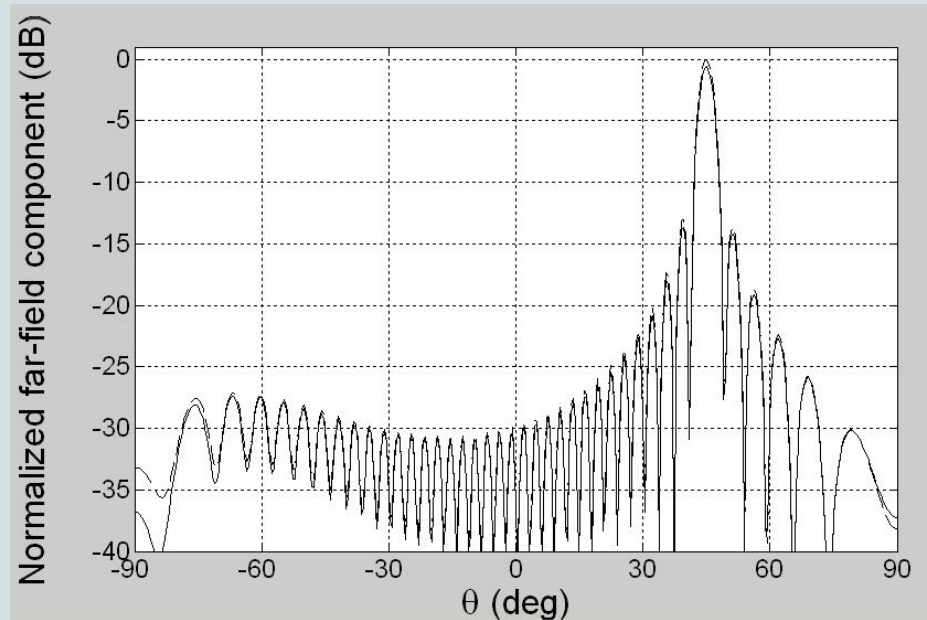
Second group:
'perturbation'

First 2 eigencurrents



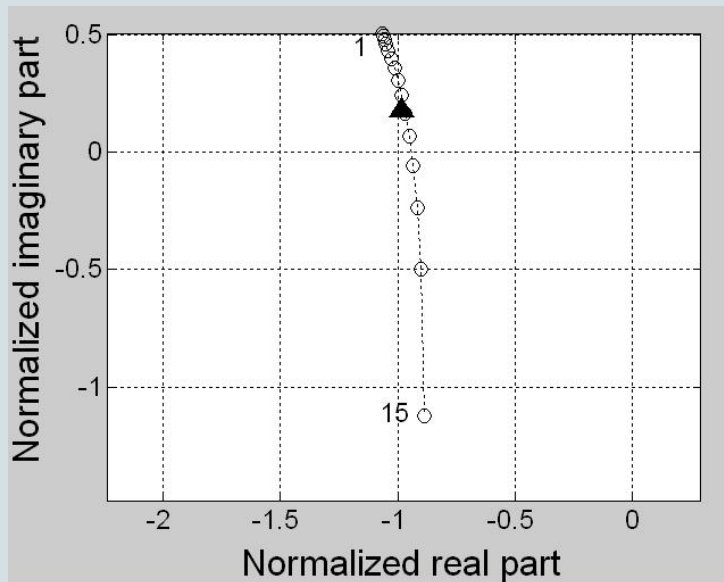
Results

Line array of 40 rings, scan at 45°, far field and normalized impedance

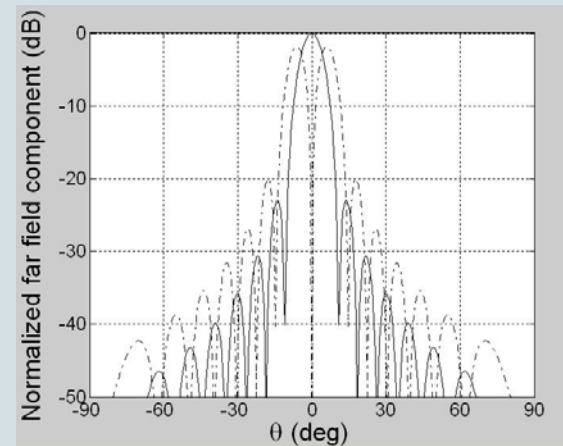


Eigencurrents: only computational ?

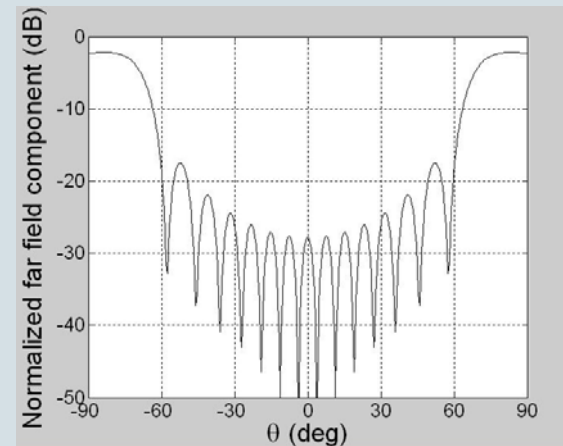
Line array of 15 rings



Eigenvalues



1st and 2nd
eigencurrent

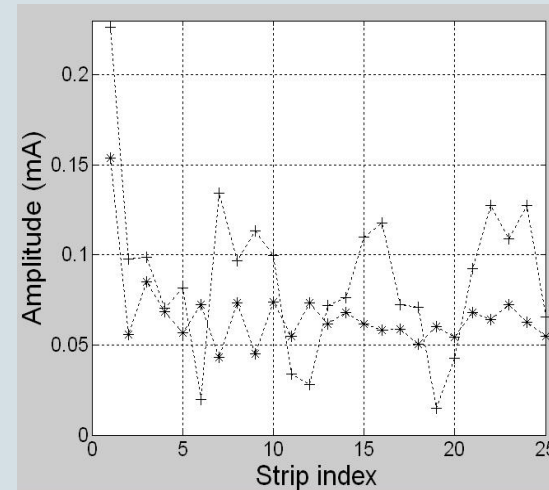
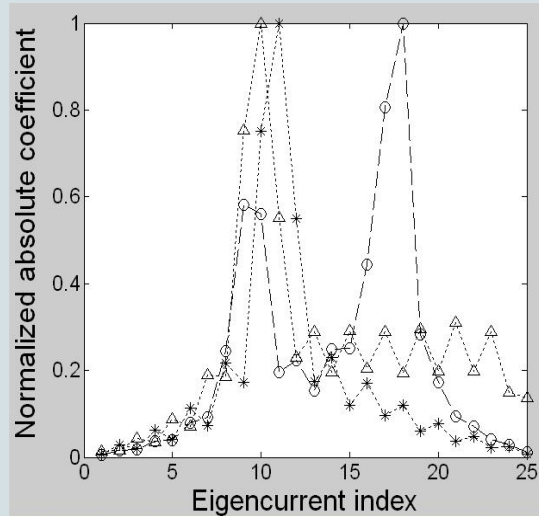


15th eigen-
current

Impedance variations

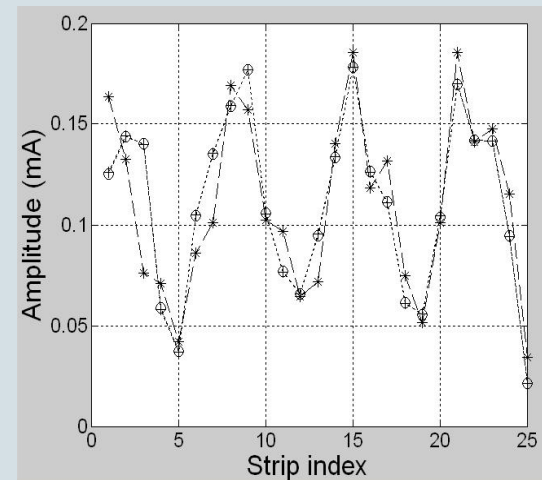
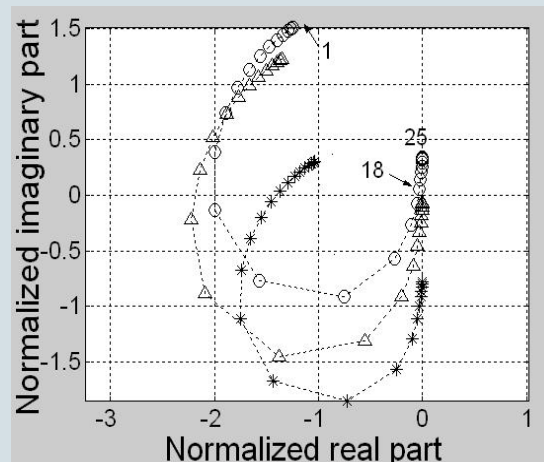
Line array
of 25 strips

Expansion
coefficients
1st group



First 14 and
first 17
eigencurrents

Eigenvalues



First 18
eigencurrents
and all
eigencurrents