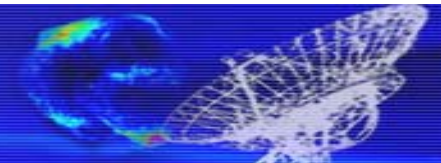


# Receiver architectures

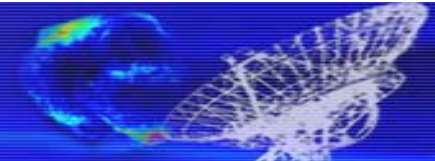
## APERTIF (DIGESTIF) as a design example

Laurens Bakker



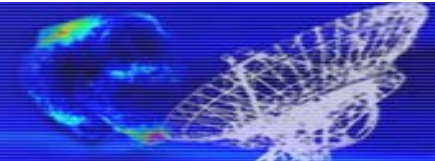
# outline

- APERTIF specifications
- System partitioning
- RFI environment at Westerbork
- Receiver architectures
- Some first receiver specifications for APERTIF
- DIGESTIF as a first step



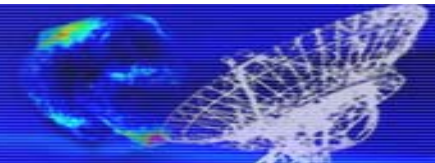
# *Some basic parameters*

- APERTIF stands for APERture Tile In Focus
- Target frequency range 850-1750MHz
- Target Instantaneous bandwidth is 300MHz
- FoV of 8 sq. degrees
  - 220 active elements required
  - 25 dual polarized beams (50 beams total)
- $A_{\text{eff}}/T_{\text{sys}} > 100 \text{m}^2/\text{K}$  (all telescopes)
  - $T_{\text{sys}} < 50\text{K}$
- Beamforming is performed all digital
- $1\text{m}^2$  available at focal point telescope for FPA
- Should be installed on the WSRT in 2010
  - Uses existing mechanical infrastructure



# Resulting order magnitude numbers

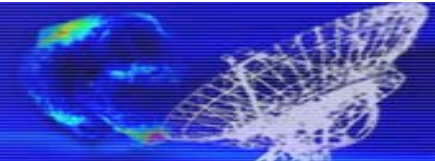
- Total datarate with 8 bits digitization is  $\sim 1.4\text{Tb/s}$ 
  - Sampling at 800MHz
- After beamforming this reduces to 120Gb/s (4 bits)
- Required processing (rough estimate):
  - Subband partitioning one element (polyphase filtering):
    - Prefilter: 16 MAC/sample
    - FFT:  $3 * \log(\text{nr\_subbands}) / \log(4) = 3 * \log(64) / \log(4) = 9$  MAC/sample
    - Total MAC/element:  $25\text{MAC/sample} * \text{sampling\_clock} = 20\text{GMAC}$
  - Beamforming (all elements used for all beams):
    - $\text{Nr\_beams} * \text{nr\_elements} * 4\text{MAC/sample} * 2 * \text{bandwidth} = 26400\text{GMAC}$
- Total required processing :  $220 * 20\text{GMAC} + 26400\text{GMAC} = 30800\text{GMAC}$
- $\sim 0.5$  of the largest FPGA /element currently available (early 2007) when running at full speed



# System partitioning

## Main drivers

- Max 375kg in prime focus of telescope
- Max 1m<sup>3</sup> in prime focus of telescope
  - Puts restriction on size of electronics
  - Puts restriction on power consumption (cooling)
- Cost
  - The main variable for partitioning are the signal transport cost and cost associated to mounting/shielding
  - Antenna cost, receiver cost and digital processing cost determined by requirements (FOV, BW, freq. range)
- Five main architectures were considered



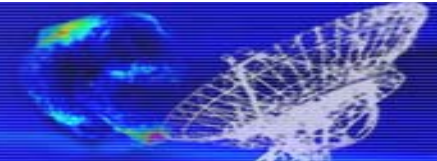
# Westerbork telescope



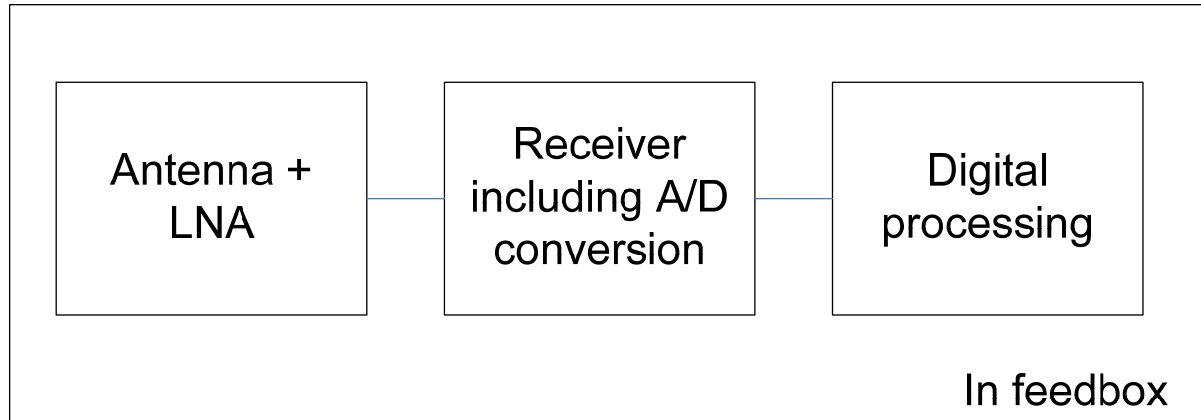
feedbox

Cable jumper

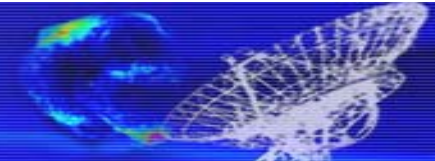
Ground based house



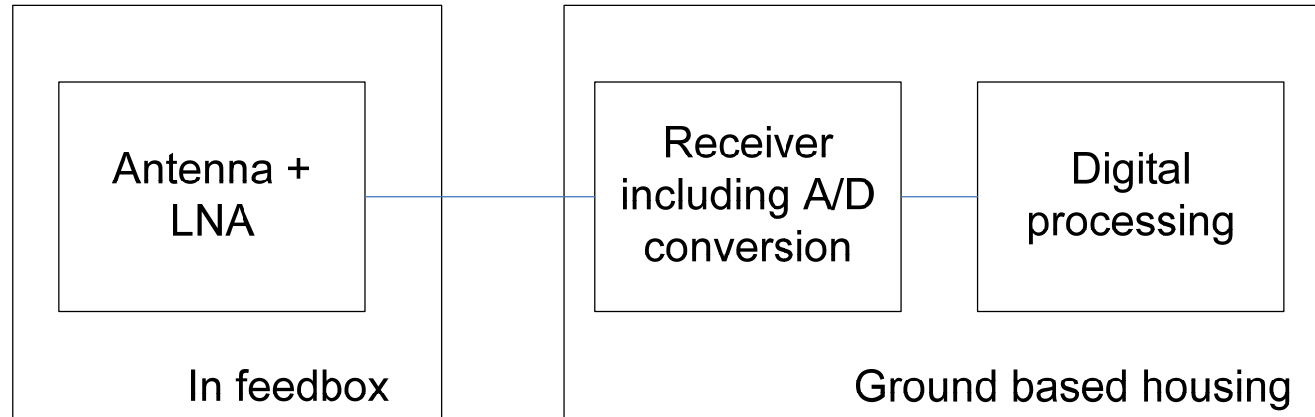
# Everything in feedbox



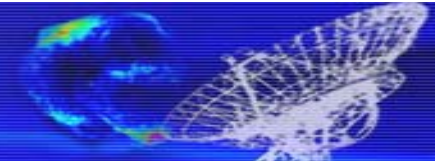
- Results in potentially the lowest cost solution
  - No signal transportation costs
- Available space will be limiting factor
  - Large level of integration required
- Cooling will become problematic
  - Some 6kW of power consumption in 1m<sup>3</sup>
- Self generated RFI of major concern



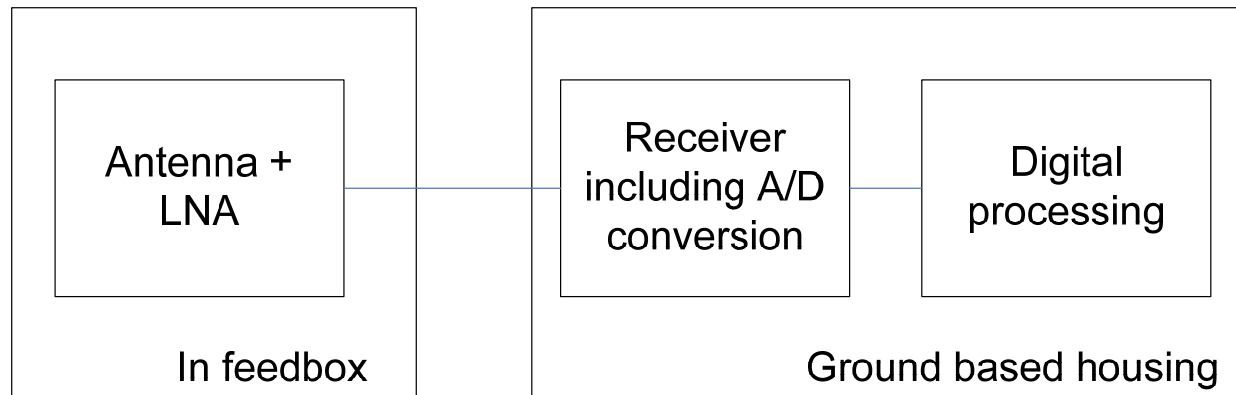
# Coax to ground based house



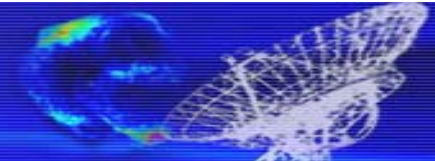
- Coax all the way to ground based house (20-50Euro element)
- Size, weight, power consumption no major concern
- Digital processing at ground
- Major issue is getting the cable all the way to the ground
  - Difficult for large amount of cables (>100)



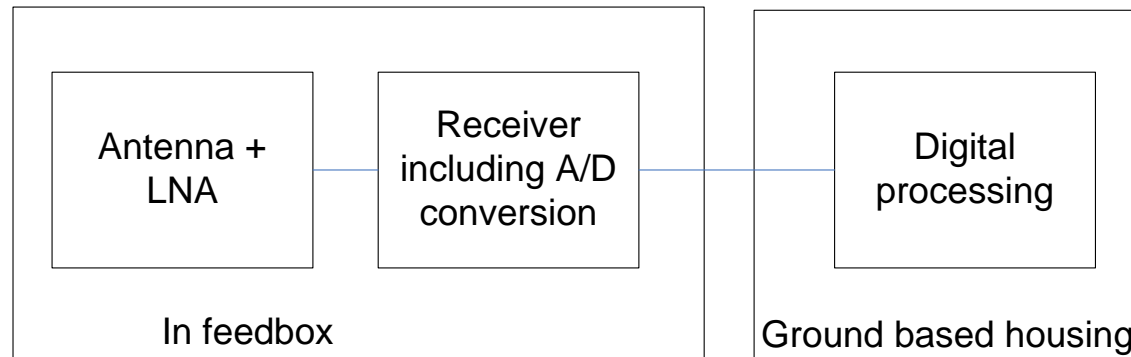
# Analog fiber link to ground based house



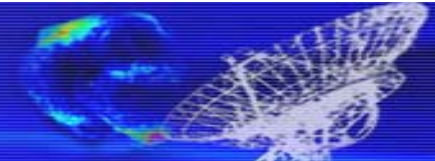
- Solves problem of large bundle of coax across jumper
- Digital processing at the ground
- Increase in cost compared to coax cable (500-1000Euro/element)
- Development of low cost analog optical link required



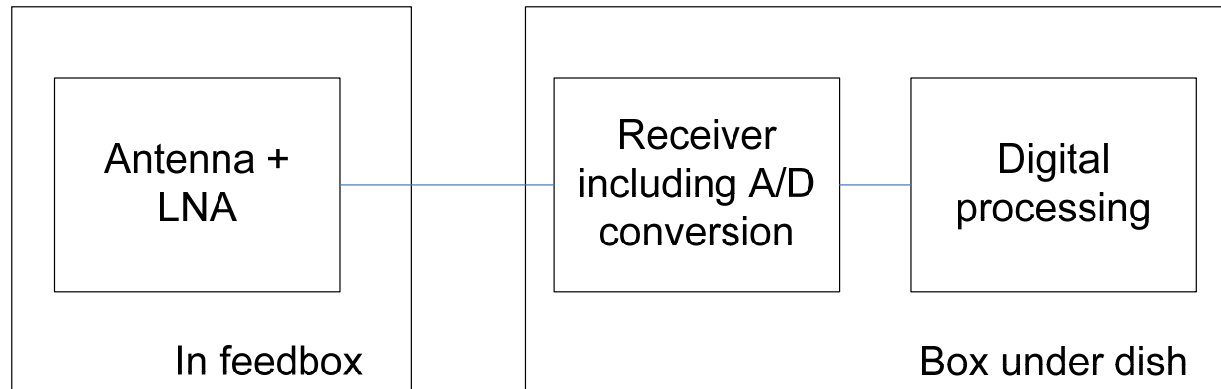
# Digital fiber link to ground based house



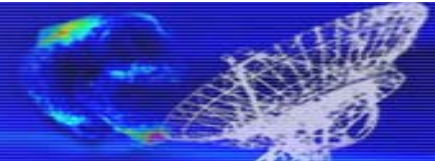
- No problem with analog cables all the way to the ground
- Processing at ground
- Self generated RFI of concern
- Technical proven technology; COTS
- High cost solution (700-1500Euro/element for transport)



# New box under reflector



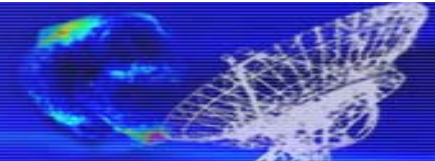
- Coax cables to new box under reflector (10-30Euro/element)
- Solves space/thermal problem in feedbox
- Avoids cables all the way down
- Question mark if it is possible to mount the box
- Box might introduce maintenance issues
- Seems most attractive, if mechanical feasible



# Westerbork telescope

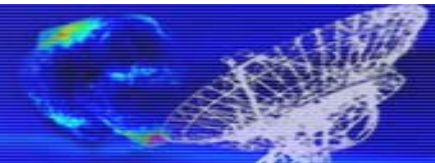


New box

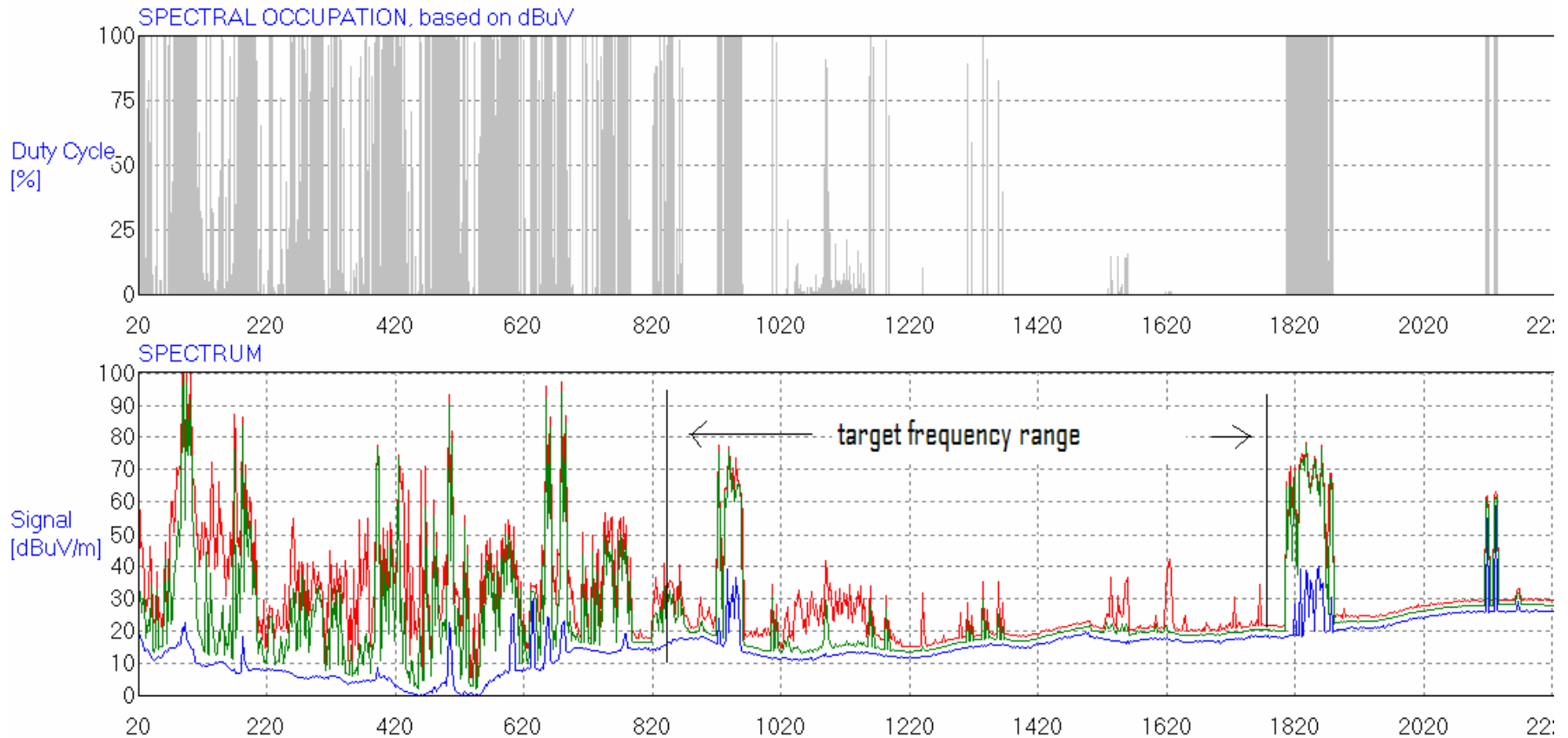


# Conclusion on system partitioning

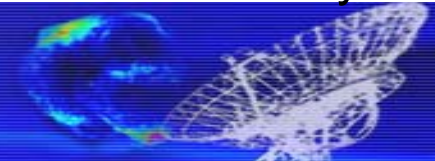
- Processing should be located as close as possible to antennas to limit signal transportation cost
- Locating everything in the feedbox is hard to realize
  - Size
  - Power consumption / cooling
  - Self generate RFI
- Digitization and processing should be co-located
- RF signal transport over coax cable seems most suitable
- New box under telescope seems most promising
  - No integration of all RF functions in one chip



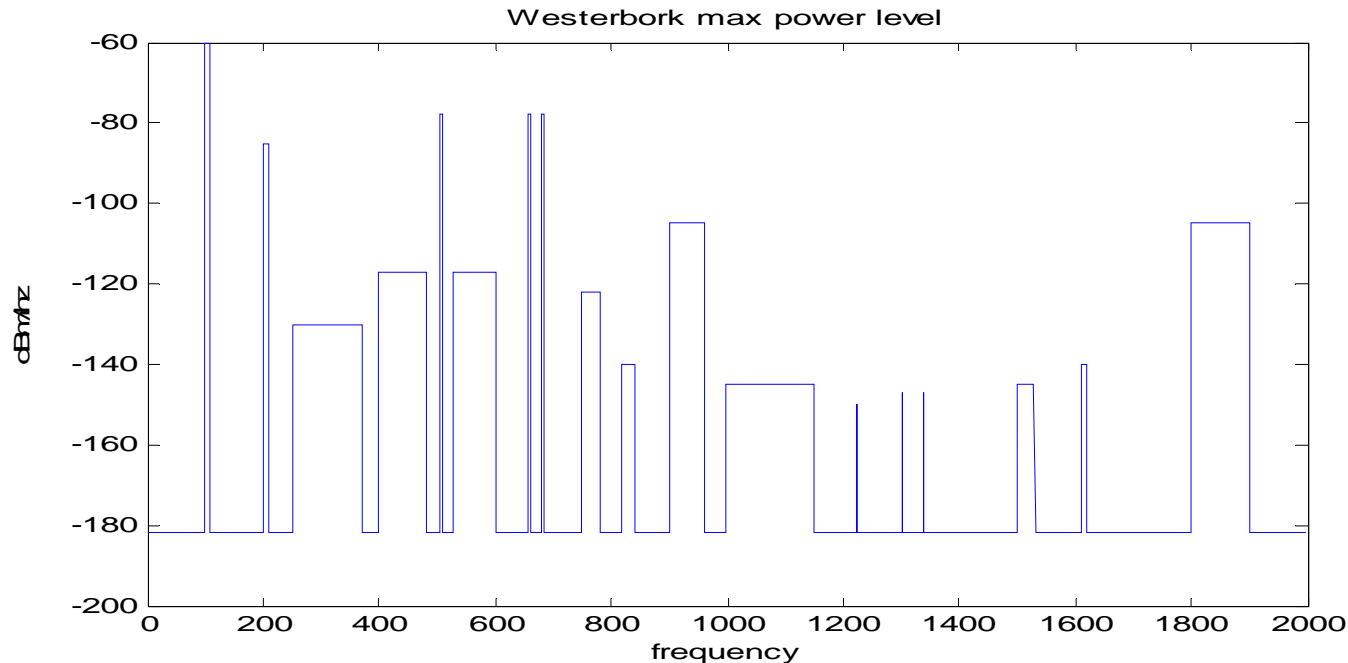
# RFI at Westerbork



- Measurements with WSRT monitoring station (30m height)
- TV channels (highest peaks) are out of band. Signals turned off recently



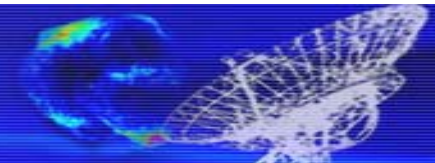
# Translation to dBm/Hz



using

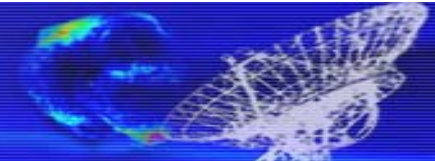
$$P_{re} = \frac{\lambda^2 E^2}{480\pi^2}$$

- Worst case analysis (max power)
- Integrate over required band and compare with KTB noise
- 8 bits are insufficient to handle GSM
- 8 bits sufficient for other parts of required frequency range

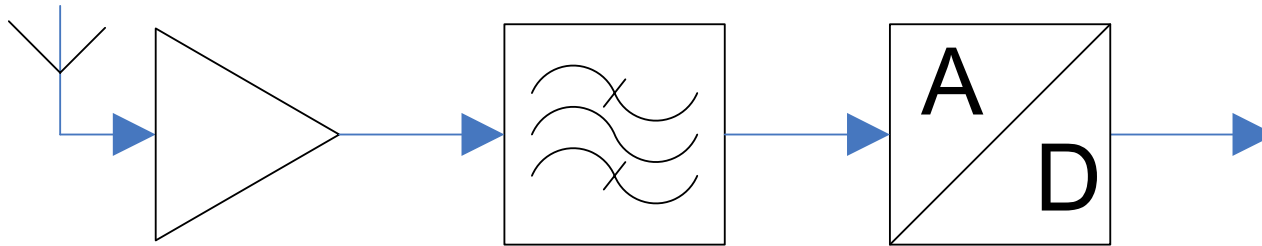


# Conclusion on RFI/nr. of bits

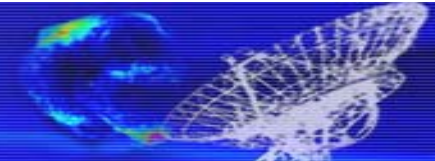
- GSM from 900-960MHz can not be sampled with 8 bits
- 8 bits OK for other part of frequency range
- Limit required number of bits -> exclude 900-960MHz
- When digitizing with 8 bits, 0.85-1GHz not useable
- The antennas will be designed for 0.85-1.75GHz
- The RF system will be designed for 0.85-1.75GHz
  - GSM band will determine linearity requirements after LNA



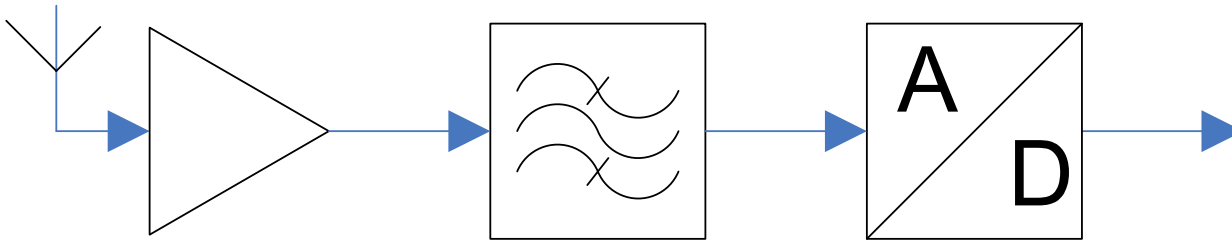
# Direct sampling



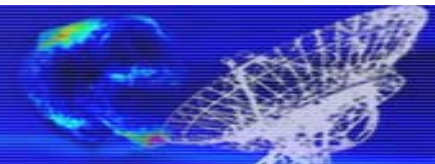
- High speed sampling at 3.5GHz
- High resulting data rate
- Large digital processing power required
- With RFI environment at Westerbork bandpass filter required
- Quality of the A/D clock will be an issue
- Required A/D converter currently not commercially available



# Undersampling

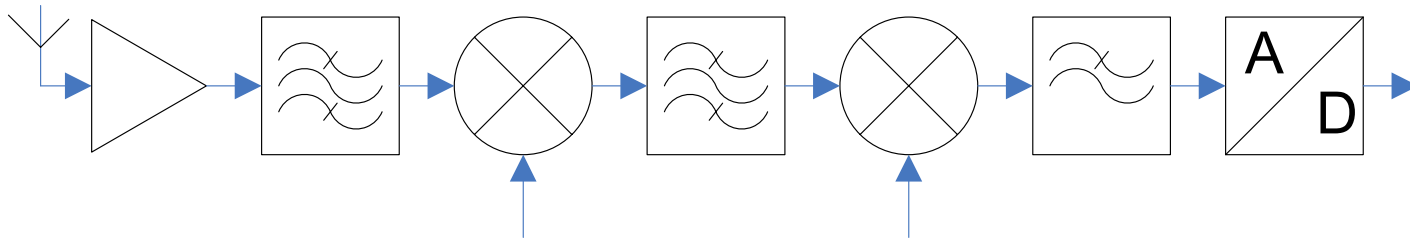


- Reduction of required sampling rate compared to direct sampling
- Analog input bandwidth of A/D should be sufficient
- A switchable bank of bandpass filters is required
  - High cost
  - Difficult to integrate
- Use only 1 filter and sample at 1750MHz
  - For reduced frequency range from 1000-1750MHz

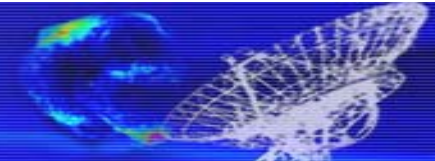


# Superheterodyne receiver

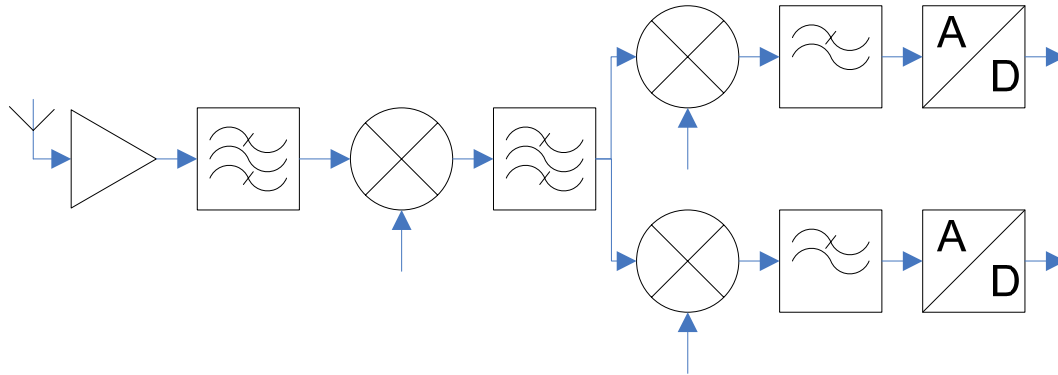
without IQ sampling



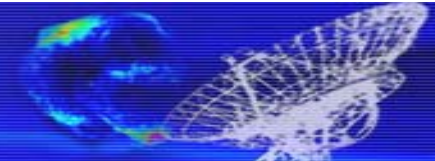
- Good handling of RFI
- Only band of interest converted and processed
- Several filters-> hard to integrate
- 2 Local oscillators
- Sampling at 800MHz (depending on quality of BP filter)
- Can be built with “discrete” components
- No large technical risks foreseen
- No large potential for cost reduction !!???



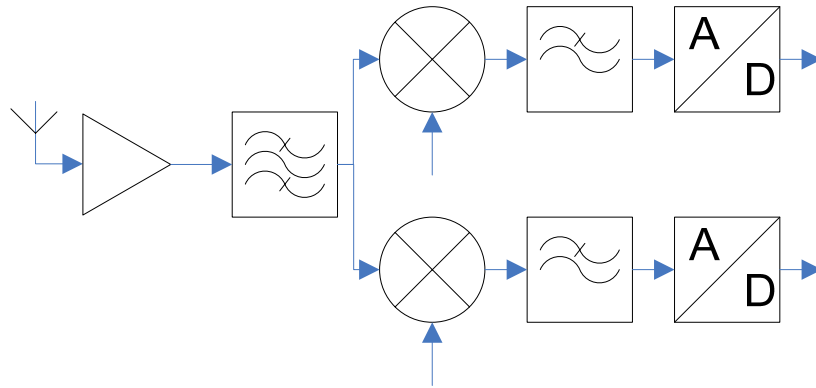
# Superheterodyne receiver with IQ sampling



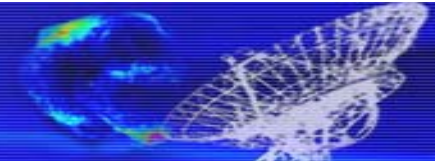
- Relaxed requirements on second filter
- “Low” speed A/D converters (400MHz, 2 required)
- Only band of interest converted and processed
- Still 2 Local oscillators
- IQ balance will be biggest challenge
  - Only at one frequency required
- Better potential for integration
  - Integration is likely required in order for this receiver to be feasible



# Zero-IF receiver

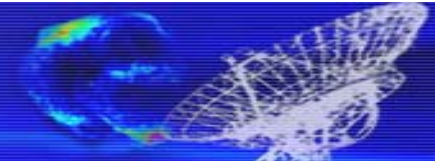


- Good potential for integration
- Only one Local oscillator required
- “low” speed A/D converter (400MHz, 2 required)
- IQ balance over large frequency range
- Only band of interest converted and processed
- Integration is required
  - Potential for low cost solution



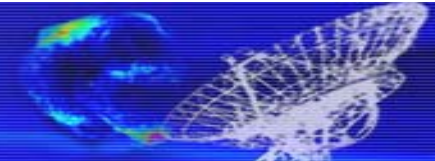
# Rx specification from antenna to A/D converter

- 66dB of net gain required between antenna and A/D
  - for 8 bits A/D, input 0.5V and 100ohm input impedance
- $T_{\text{sys}} < 50\text{K}$
- Second and third order distortion set 10dB below noise level
- For LNA TV-channels Smilde important
  - Analog signals Smilde recently turned off but now digital
- other components IP2 and IP3 requirements can be reduced significantly by sufficient filtering

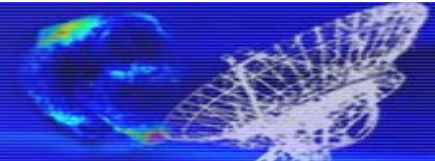
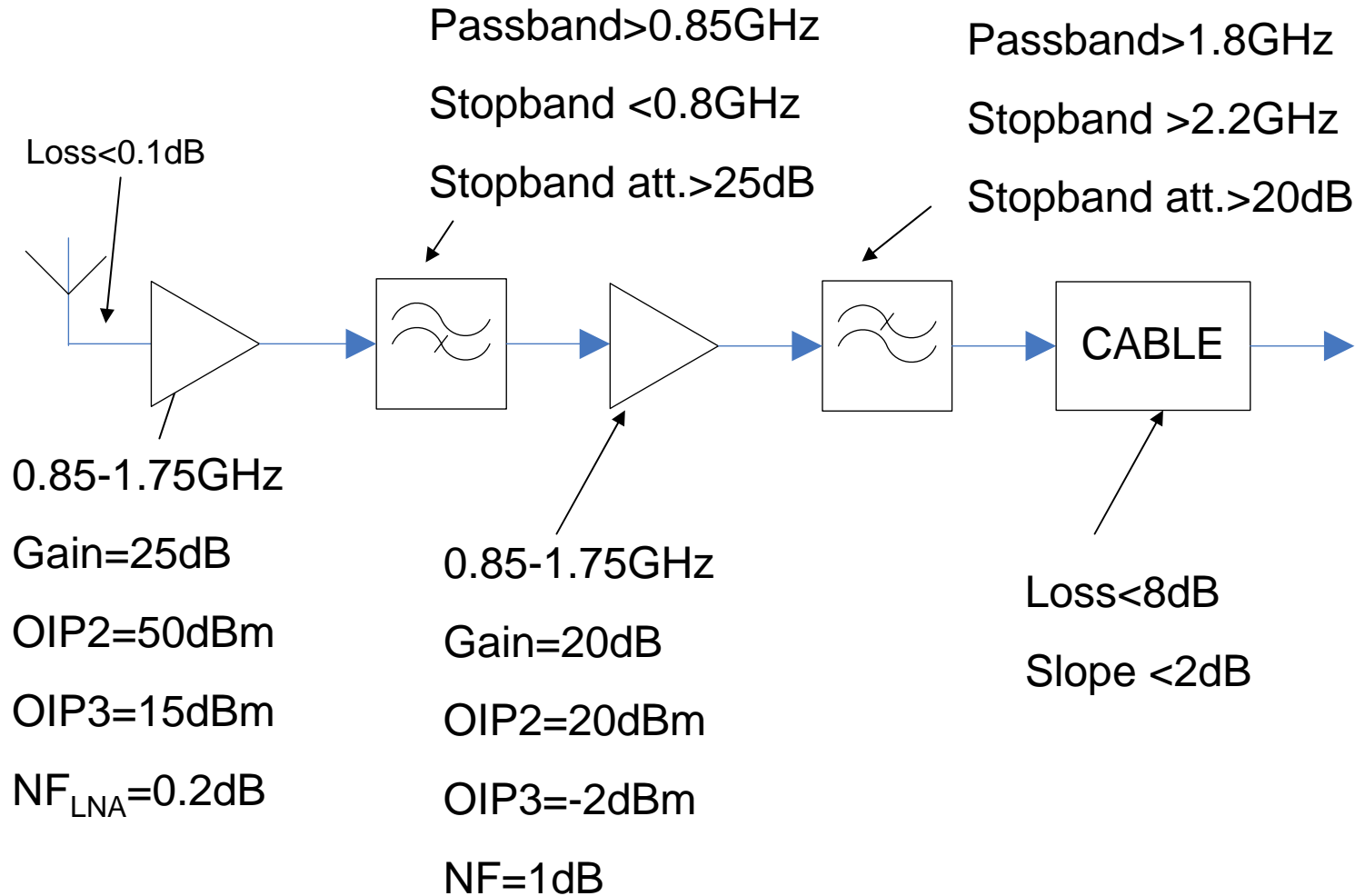


# LNA requirements

- Spillover=15K
- Noise coupling =5K
- $T_{\text{sky}} = 5\text{K}$
- Loss between antenna and LNA 0.1dB
- Means  $T_{\text{Ina}}$  needs to be 18K (0.25dB) in order to achieve  $T_{\text{sys}}$
  
- Cooling might be needed achieve the required  $T_{\text{sys}}$
- With  $1\text{m}^2$  of antenna-elements opening of cooler very large and/or long cables
  - Likely unrealistic for  $1\text{m}^2$  of antenna-elements
- Local cooling ??



# Preliminary Rx specifications

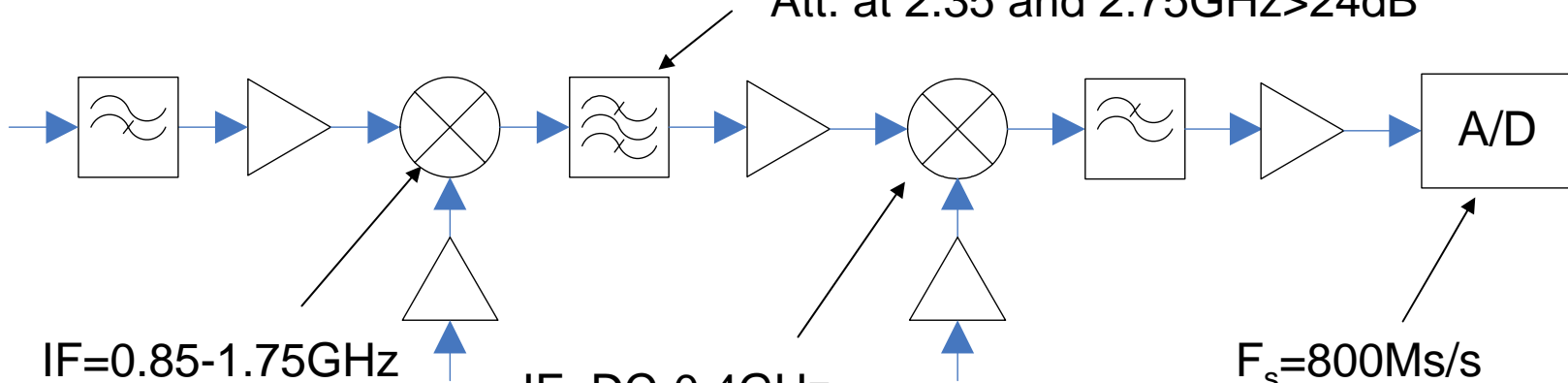


# Preliminary Rx specifications(2)

Passband 2.4-2.7GHz

Stopband 2.35-2.75GHz

Att. at 2.35 and 2.75GHz >24dB



IF=0.85-1.75GHz

RF=2.4-2.7GHz

LO=3.55-4.15GHz

IF=DC-0.4GHz

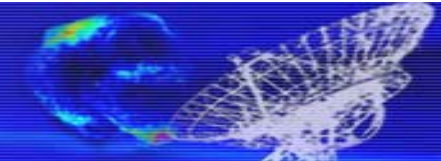
RF=2.4-2.7GHz

LO=2.75GHz

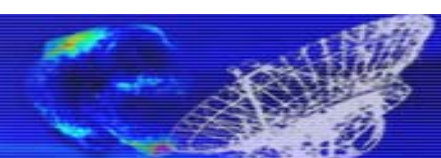
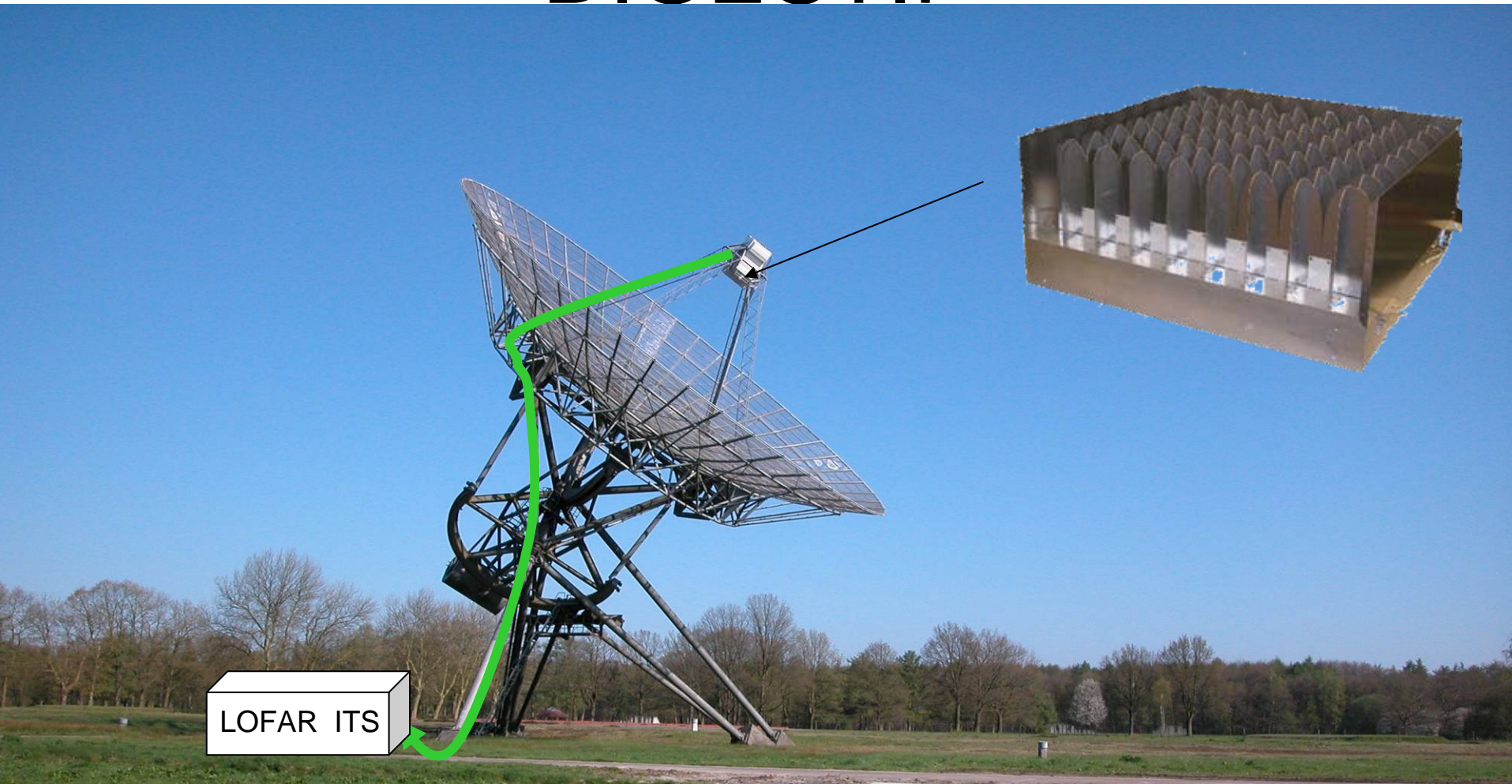
$F_s = 800\text{Ms/s}$

Signal gain complete receiver ~30dB

All specifications to be checked with DIGESTIF

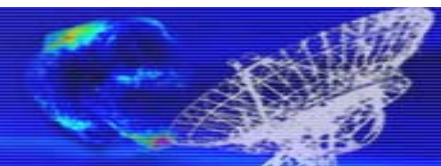


# DIGESTIF



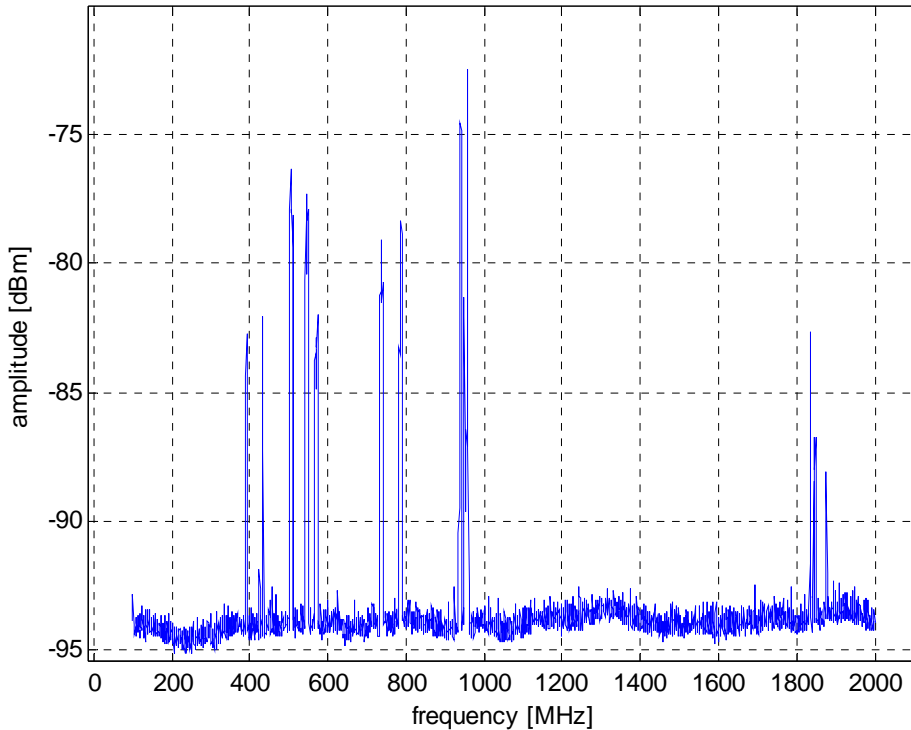
# DIGESTIF specifications

- Frequency range                      **1.0 – 1.7 GHz**
  - Array size                                80 x 80 cm
  - Dual linear polarization            (vivaldi elements)
  - Number of elements                 8 x 7 x 2 = 112
  - Instantaneous bandwidth          30 MHz
  - Number of receivers                 60
  - Data storage                            max 6.7 seconds per measurement
  - System temperature                 100 K
  - Dish diameter                         25 m
- 
- Additional requirement : no structural changes to WSRT



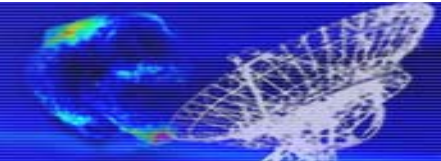
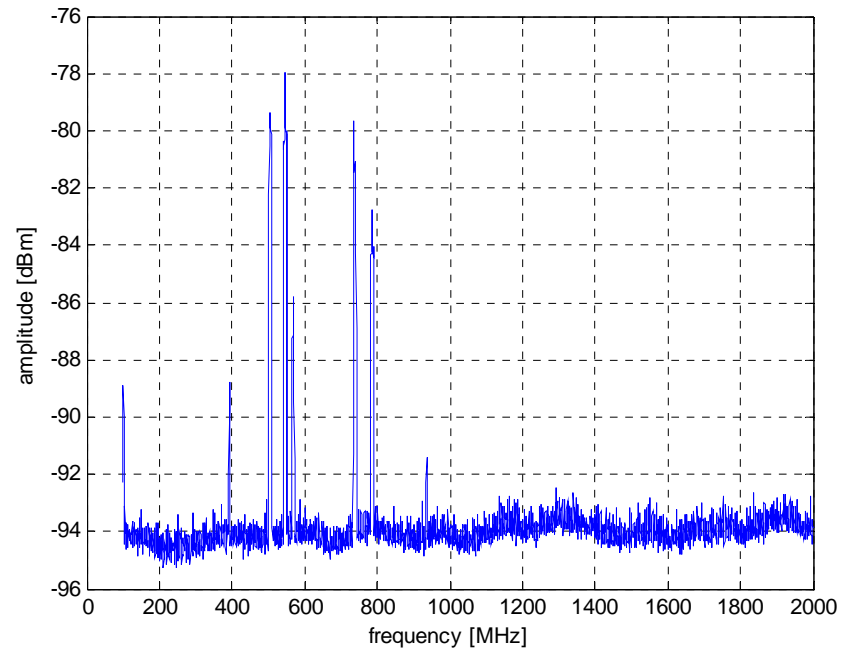
# Measurement RFI with antenna

0-37.DAT 30.Jul 07 rbw:1000000 vbw:30000 swt:6.50e-002 det:MAXPEAK Att:0 dB  $P_{tot}$ : -58.6 dB

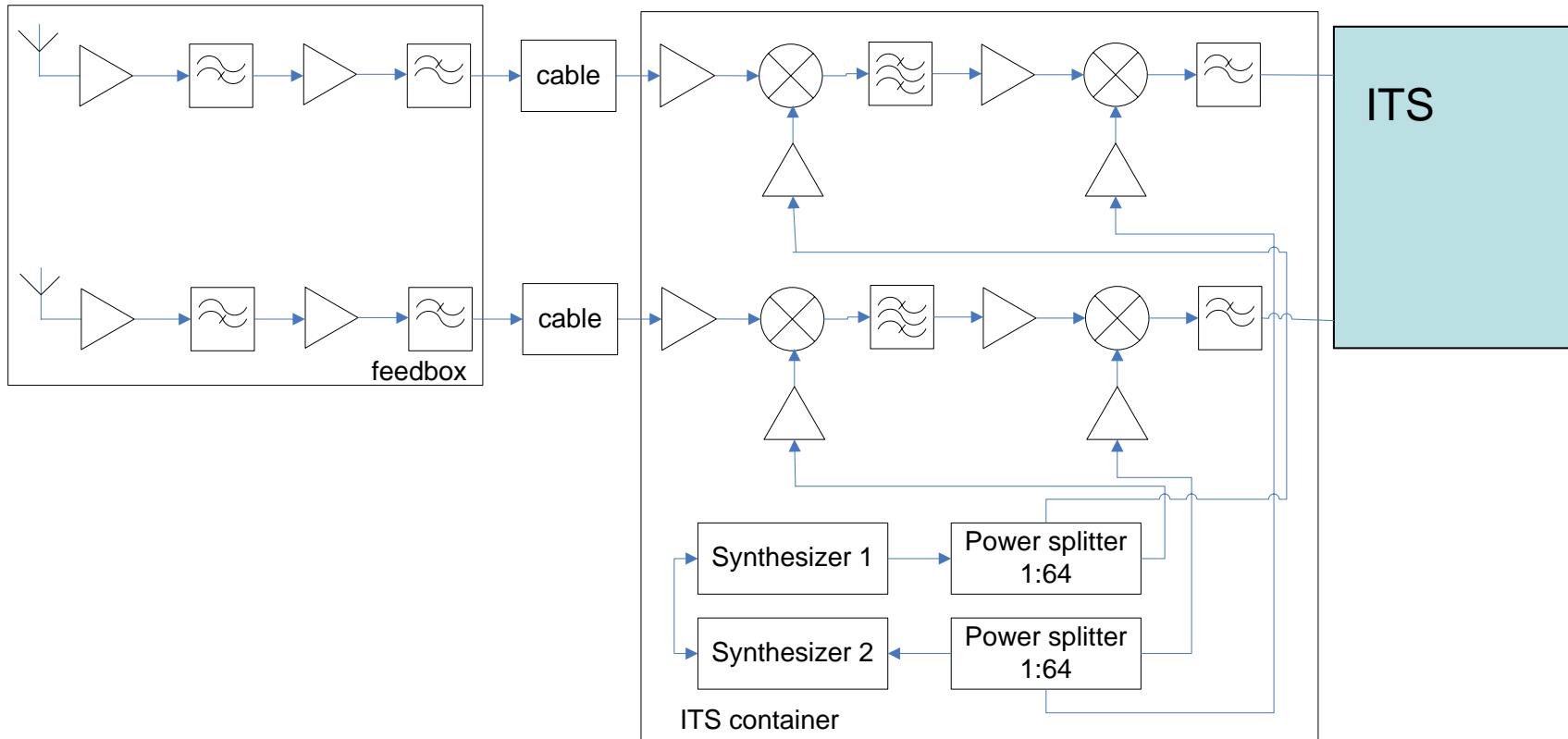


Measured in container (add ~20dB for cable loss and 50-75ohm transformers)

0-20.DAT 30.Jul 07 rbw:1000000 vbw:30000 swt:6.50e-002 det:MAXPEAK Att:0 dB  $P_{tot}$ : -60 dB



# DIGESTIF schematic

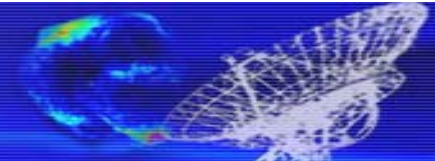


- No control outside ITS container
- Uses as much as possible APERTIF architecture



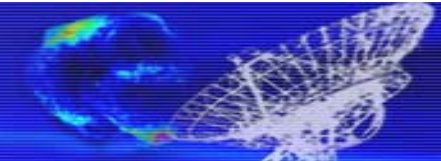
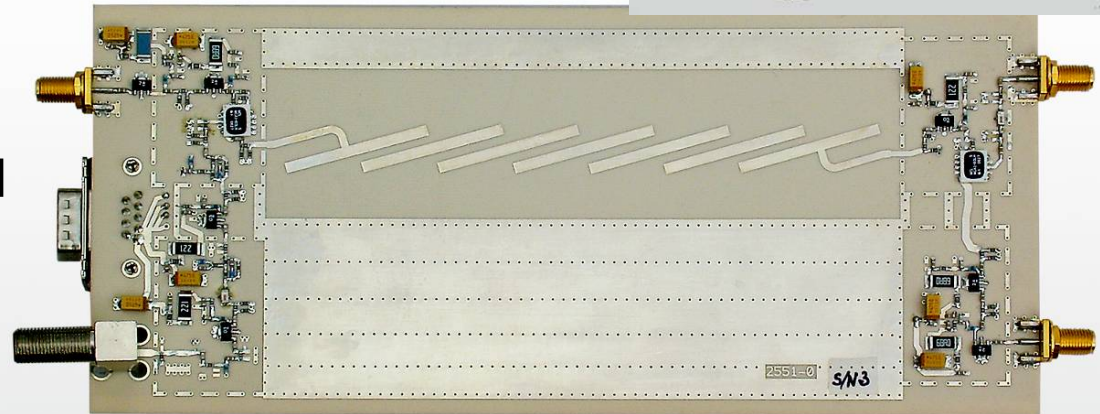
# LNA and cable

- Design for 50ohm input impedance
  - 40 dB gain (flat)
  - HPF with 30dB attenuation at 800MHz
  - 3 stages, with filtering after first stage
  - $T_{LNA} = 65$  K (over entire band, measured)
    - With reasonable S11 (-7.5dB worst case)
  - IP2 first stage  $\sim 40$ dBm
  - Sma input connector (50ohm)
  - F-connector output (75 ohm)
- 
- Cable loss at 1.7Ghz 15 dB
  - Gain slope from 1 GHz to 1.7 GHz 5 dB
  - Corrected in Down Conversion Unit (DCU)
  - Cables are tuned to  $< 100$ ps



# Down Conversion Unit

- Input band 1-1.7GHz
- Bandpass filter 2.5GHz center, 80 MHz BW, 60dB down for  $|f_c - f| > 100\text{MHz}$ , one steep slope required
- Convert to 40-80 MHz band
- 13 dB gain from input to output (1.4GHz input)
- LO1 3.5-4.1 GHz, input level  $> -13\text{dBm}$
- LO2 fixed at 2.56 GHz, input level  $> -13\text{dBm}$
- 10dBm mixers
- Cable equalizer included
- Additional shielding used
- Single supply voltage
- Power consumption 5W



# Receiver unit (ITS)

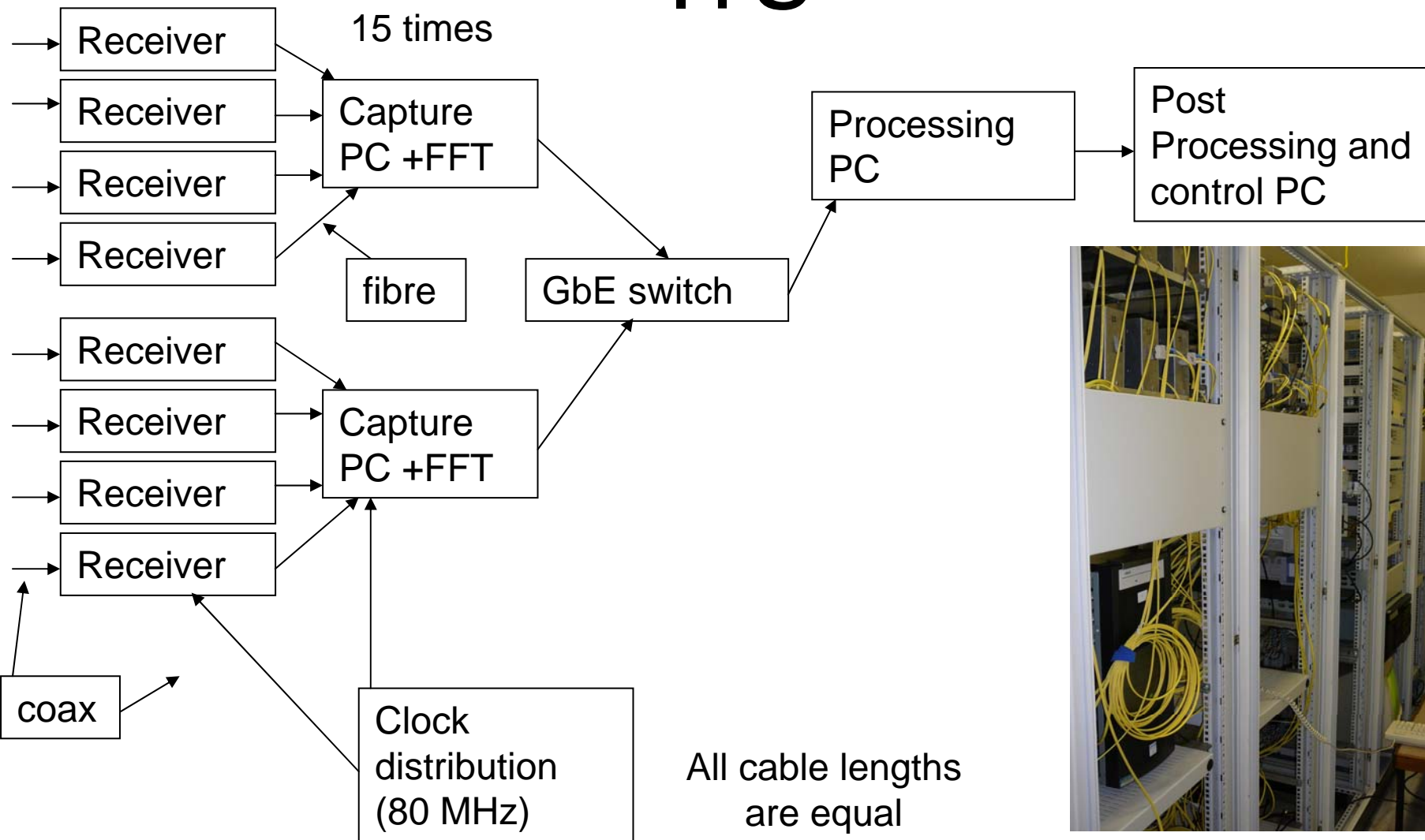
- Sampling at 80 MHz
- 12 bits (10 effective)
- Sampling in second nyquist zone
- 40-80MHz bandpass filter
  - >60dB down at 38 and 82MHz
- Signal gain receiver 25dB
- High Speed Optical Link to computer



# ITS

60 times

15 times



# Conclusions

- A new box under dish seems best solution
- 0.85-1GHz can likely not be used because of RFI
  - For a limited scan range of telescope this frequency range can be used, but then more stringent requirements on filtering
- A super-heterodyne receiver will be used
- $T_{\text{sys}}$  (not  $T_{\text{Ina}}$ ) of 50K will be challenging to achieve
- DIGESTIF:
  - 4 channels are currently operational (one with good LNA, 4 next week)
  - 32 channels will become operational in the next couple of weeks
  - Full system operational next couple of months
- First results with DIGESTIF ->presentation of Wim

