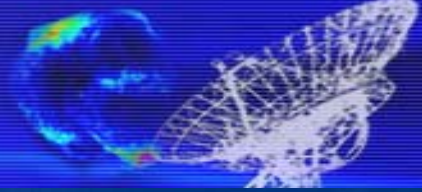


Calculation of phased array noise temperature

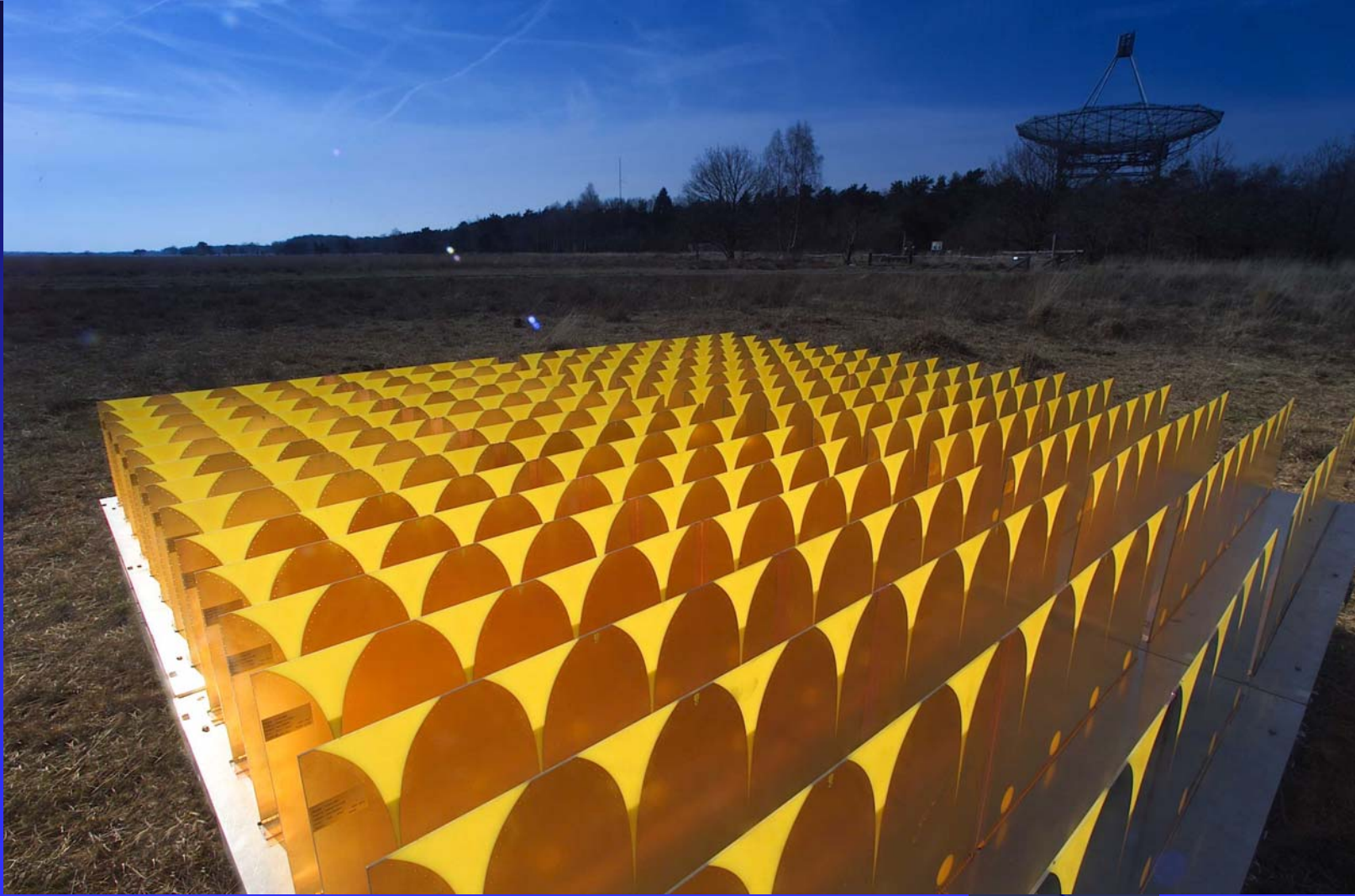
Bert Woestenburg

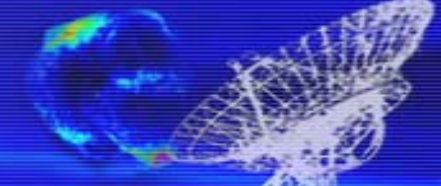




THEA

Aperture array

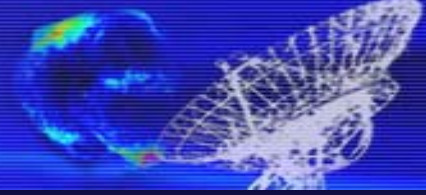




Digestif

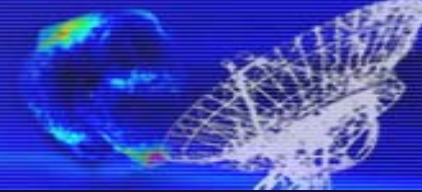
Focal Plane Array



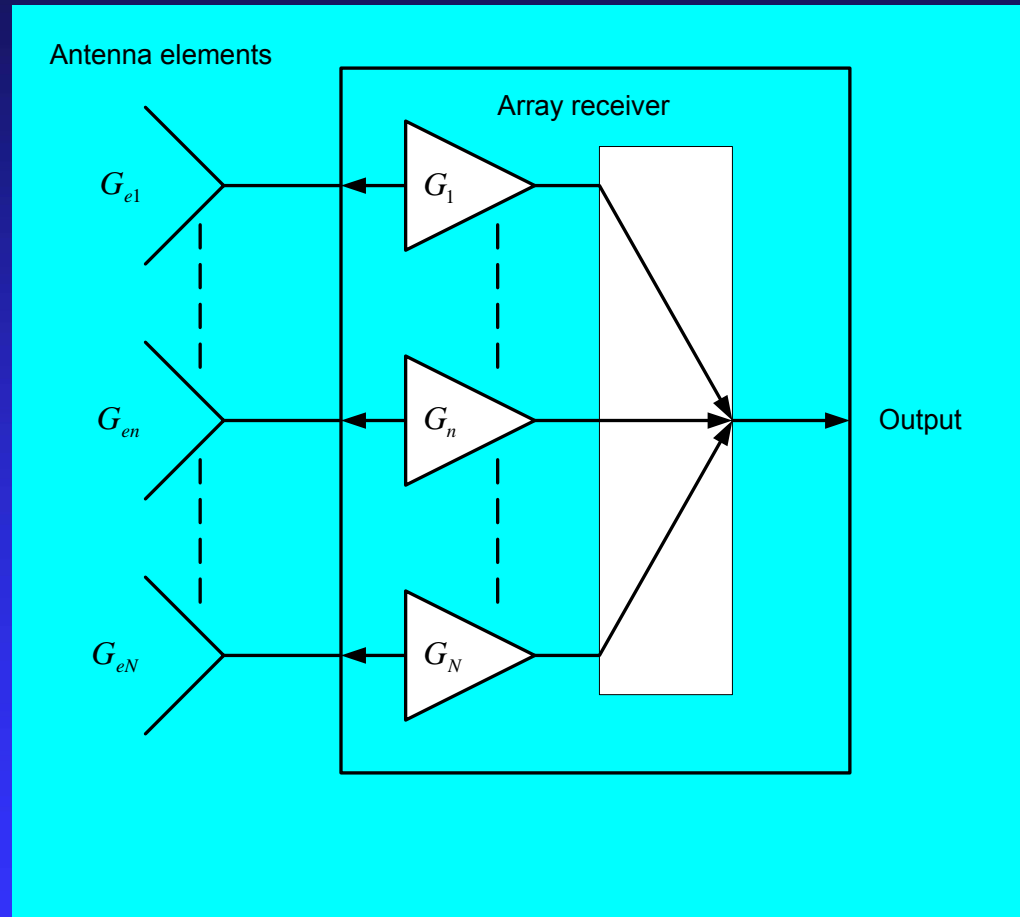


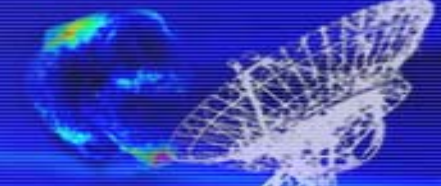
Outline

- Modeling of an array system with wave parameters
- Modeling an array system as a combination of two-ports
- Example of WW-procedure for FARADAY
- Calculation methods (4) for array noise temperature
- Examples of calculations for some simple arrays
- Conclusion



A practical array receiver system





Array modeling with wave parameters

- Noise output due to internal noise sources

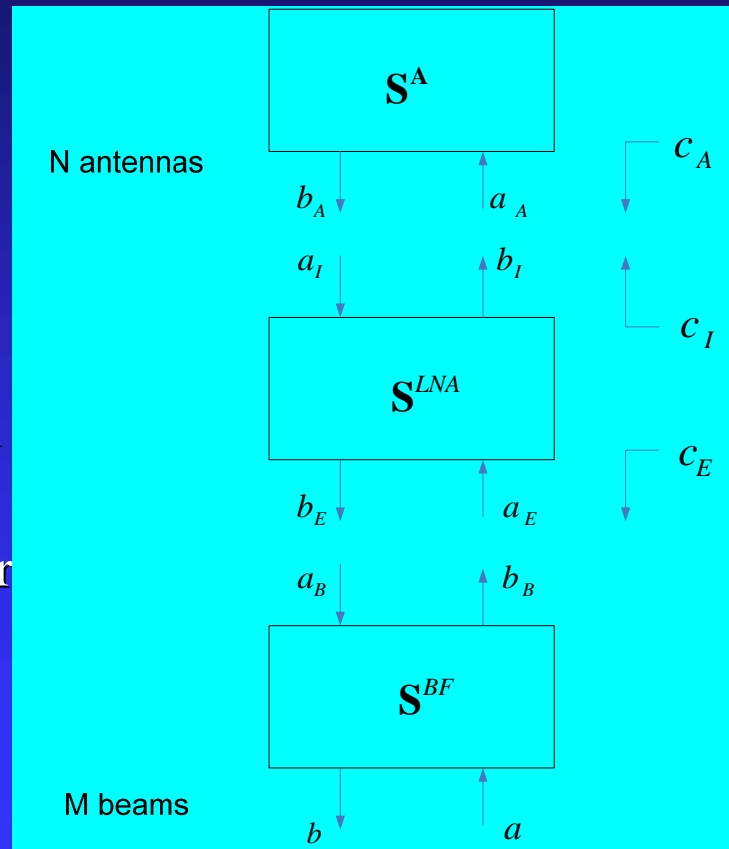
$$\underline{b}_E = \underline{c}_E + \mathbf{S}_{21}^{LNA} \mathbf{S}^A (\mathbf{I} - \mathbf{S}_{11}^{LNA} \mathbf{S}^A)^{-1} \underline{c}_I$$

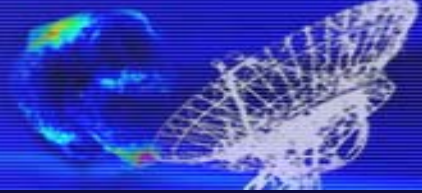
$$\underline{b} = \mathbf{S}_{21}^{BF} \underline{b}_E$$

fully describes array noise behavior; basis for array noise calculations with CAESAR

- Matrix description using active reflection coefficient does not easily show criteria for optimization

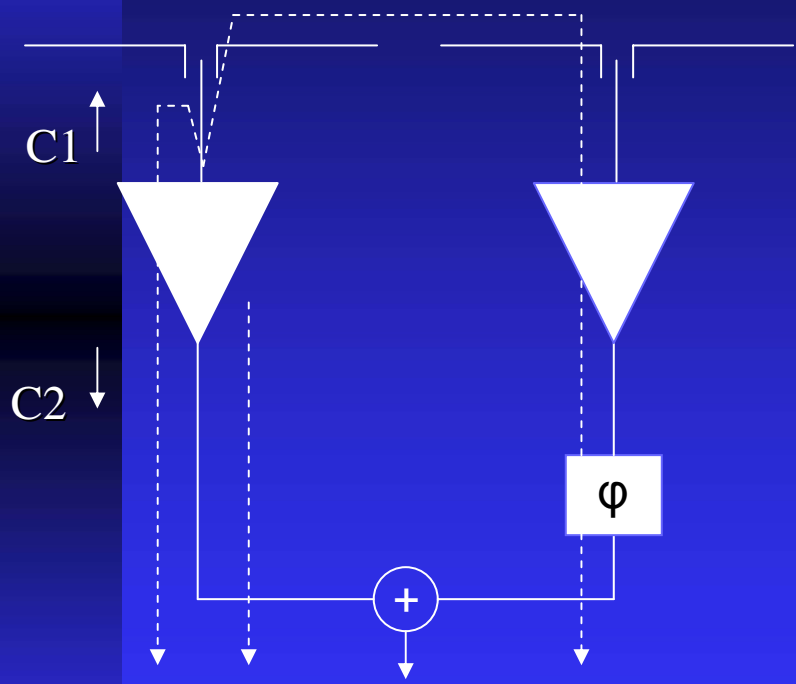
$$\underline{b} = \underline{w} \mathbf{C}_E + S_{21} \underline{w} \mathbf{R}_{act} (\mathbf{I} - \mathbf{S}_{11}^{LNA} \mathbf{S}^A)^{-1} \mathbf{C}_I$$



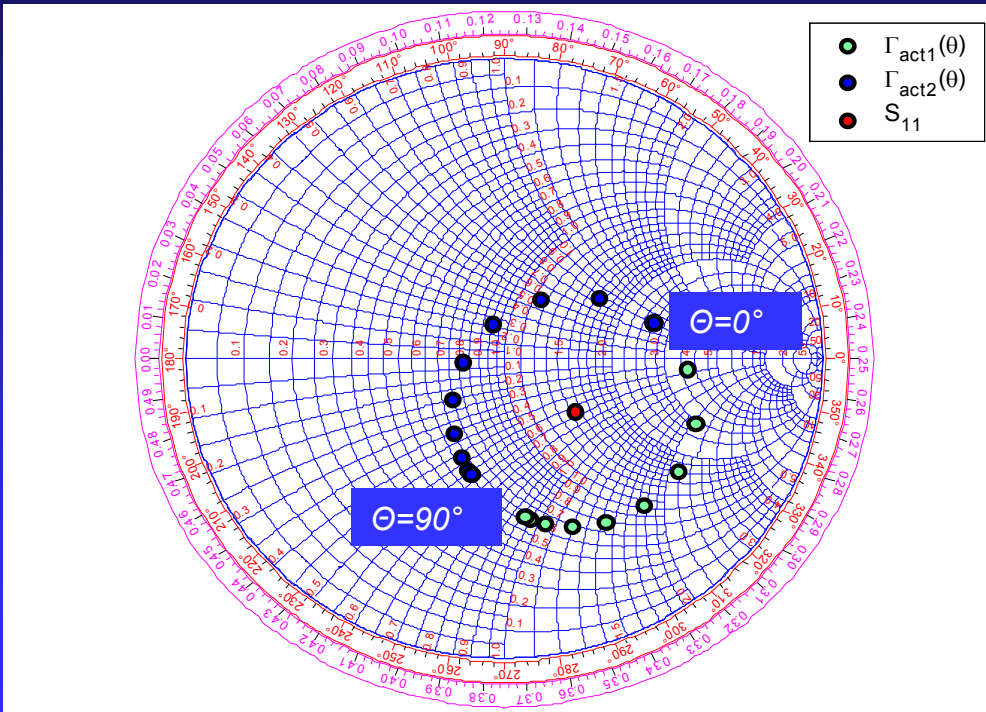


A two-element dipole array

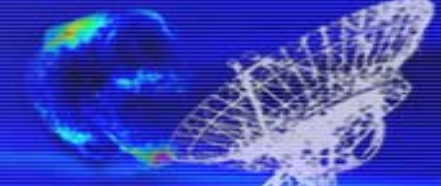
Reflection coefficients in a 2x1 dipole array



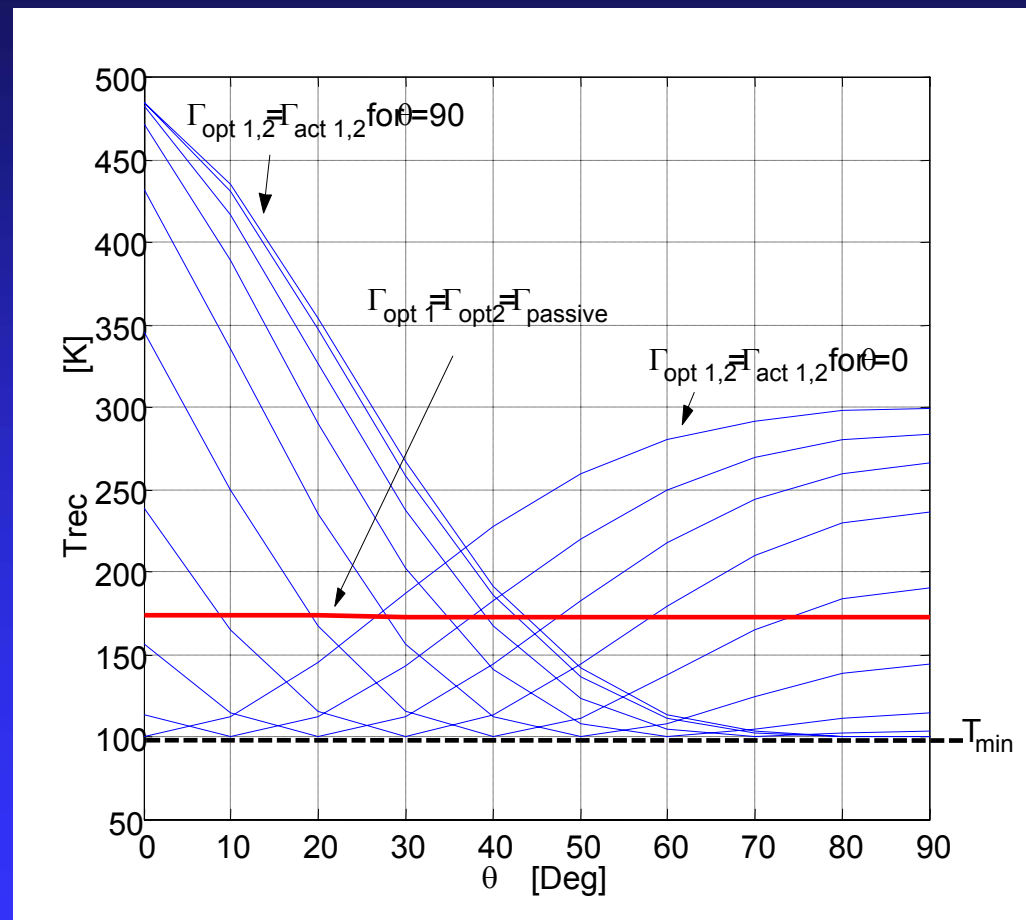
$$C_{tot} = Direct + Refl + Coupl$$

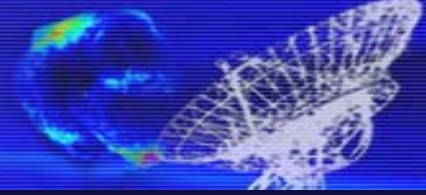


$$\Gamma_{act_i} = \sum_j S_{ji} e^{j(\Phi_j - \Phi_i)}$$



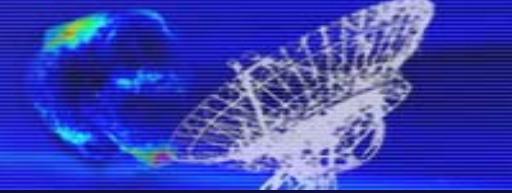
T_{rec} as a function of scan angle (results with CAESAR)





Modeling the array as a combination of two-ports

- Replace the array system by an equivalent single-channel system
- The single antenna represents the array beam properties
- The single receiver represents the array gain and noise temperature
- The receiver gain and noise temperature are expressed in the properties of the individual two-port receivers, embedded in the array environment
- Thus the effects of mutual coupling in the array on the two-port noise and gains are included, effectively defining independent two-port receiver channels

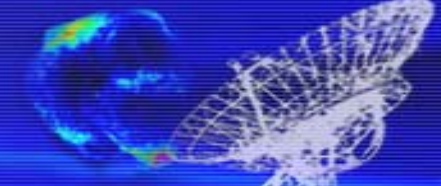


Embedded two-port noise temperature and available gain formulas

$$T_n = T_{\min} + \frac{4R_n T_0}{Z_0} \frac{|\Gamma_{act} - \Gamma_{opt}|^2}{|1 + \Gamma_{opt}|^2 (1 - |\Gamma_{act}|^2)}$$

$$|\Gamma_{act}| \neq 1$$

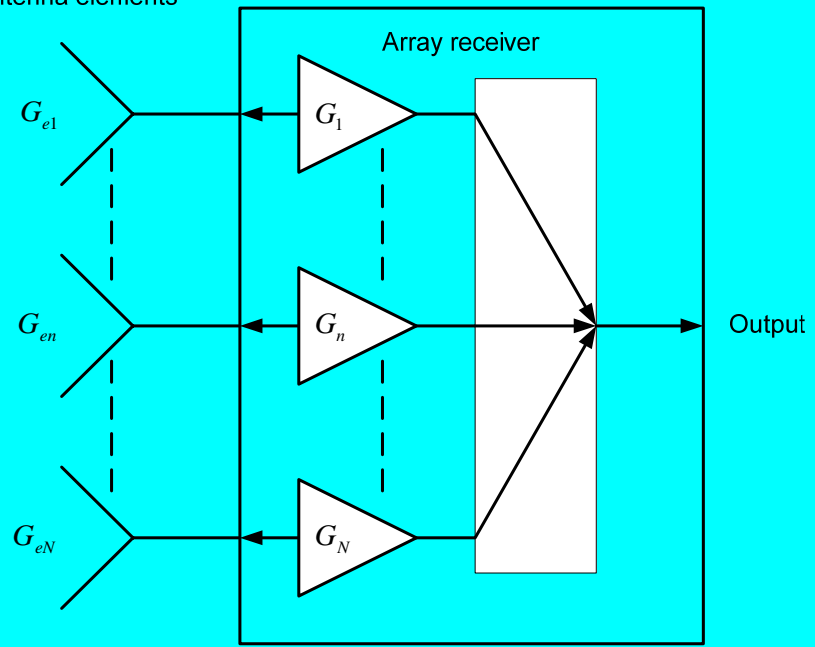
Available gain: $G_{av} = \frac{|S_{21}|^2 (1 - |\Gamma_{act}|^2)}{|1 - \Gamma_{act} S_{11}|^2 (1 - |S'_{22}|^2)}$, $S'_{22} = S_{22} + \frac{S_{12} \Gamma_{act} S_{21}}{1 - \Gamma_{act} S_{11}}$



Definition of array receiver gain

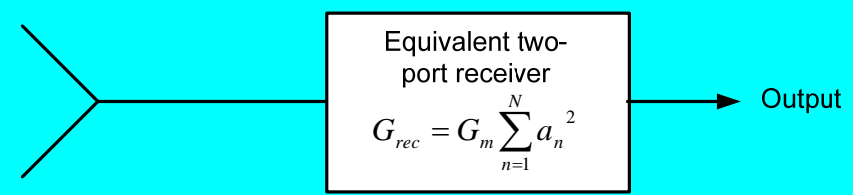
Replace the array system by an equivalent single-channel system

Antenna elements



$$S_o = P_o G_m \left| \sum_{n=1}^N (\sqrt{G_{en}}) a_n \exp(j\theta_n) \right|^2$$

$$a_n = \sqrt{\frac{G_n}{G_m}}$$

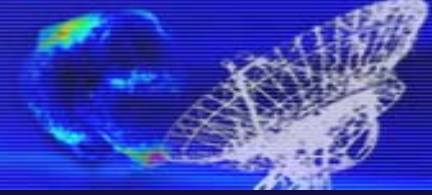


Equivalent single-output antenna

$$G_a = \left| \sum_{n=1}^N (\sqrt{G_{en}}) a_n \exp(j\theta_n) \right|^2 / \sum_{n=1}^N a_n^2$$

$$S_o = P_o G_a G_m \sum_{n=1}^N a_n^2$$

$$G_{rec} = G_m \sum_{n=1}^N a_n^2 = \sum_{n=1}^N G_n$$



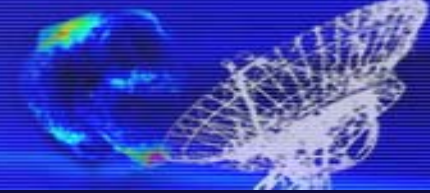
Definition of array receiver noise temperature

Calculate from excess output noise density N_o

of receiver: $T_{rec} = \frac{N_o}{kG_{rec}}$

Array receiver noise temperature is weighted average of the individual embedded receiver channel noise temperatures

$$T_{rec} = \frac{\sum_{n=1}^N G_n T_n}{\sum_{n=1}^N G_n}$$



Calculation of equivalent receiver gain and noise, including combiner

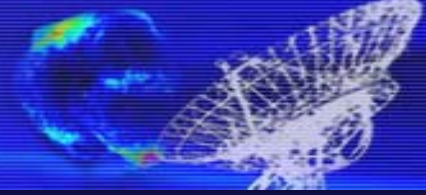
- Consider simple N-channel array of identical two-ports, formed by a cascade of two-ports, and one combiner with amplitude taper. Gain of two-port:

$$G_{rec} = \sum_{n=1}^N G_n = G \sum_{n=1}^N G_{cn} = GG_c$$

- Excess output noise density $N_o = kTG_{rec} + kT_0(1-G_c)$ gives noise temperature, according to cascade noise formula for one channel two-port with a second stage passive lossy two-port, representing the combiner

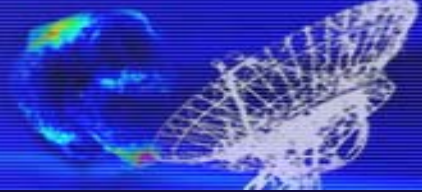
$$T_{rec} = T + \frac{T_0}{G} \left(\frac{1}{G_c} - 1 \right), \text{ with}$$

$$T = \frac{\sum_{n=1}^N G_n T_n}{\sum_{n=1}^N G_n}$$

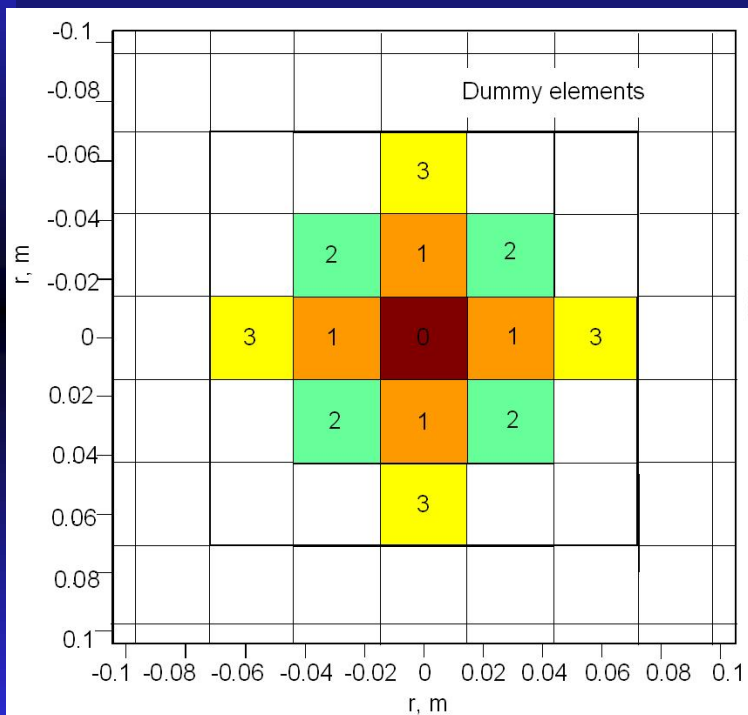


References

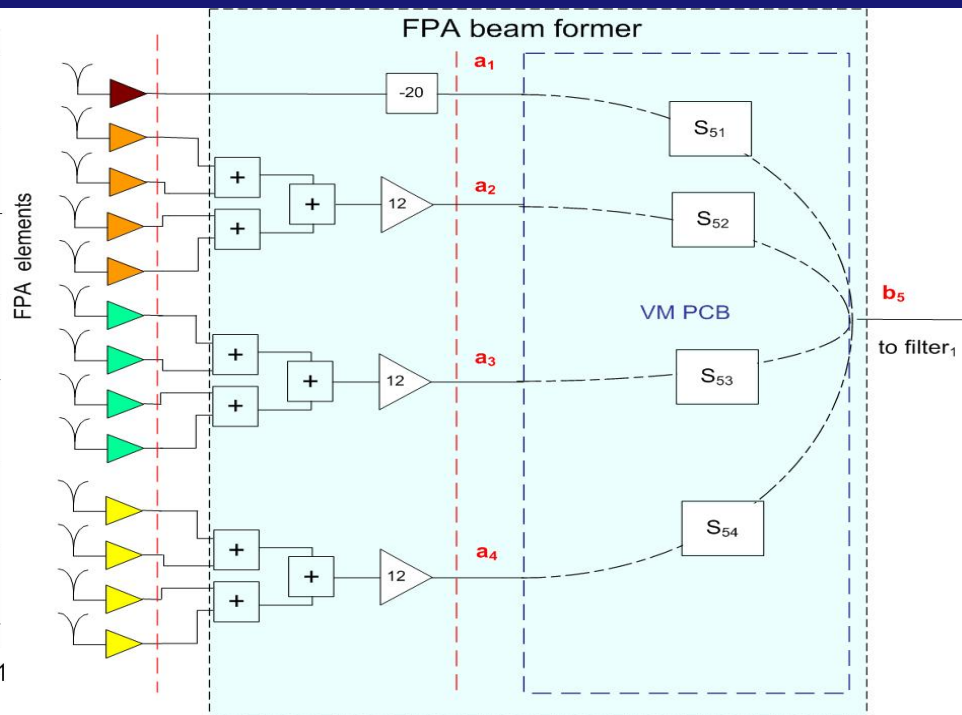
- Waldman, A., Wooley, G.J., “Noise Temperature of a Phased Array Receiver”, The Microwave Journal, pp. 89-96, September 1966
- Lee, J.J., “G/T and Noise Figure of Active Arrays”, IEEE Transactions on Antennas and Propagation, Vol.41, No. 2, pp. 241-244, February 1993
- Woestenburg, E.E.M., “Definition of Array Receiver System Gain and Noise Temperature”, Poster at Paris SKA Workshop, September 2006, ASTRON Doc.nr. RP-123, August 2006
- Maaskant, R., Woestenburg, E.E.M., “Applying the Active Antenna Impedance to Noise Match in Receiving Array Antennas”, Proc. APS, June 2007, Hawaii
- Maaskant, R., Yang, B., “A Combined Electromagnetic and Microwave Antenna System Simulator for Radio Astronomy”, Proc. EuCAP, November 2006, Nice



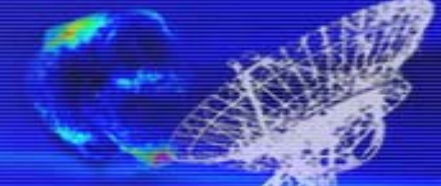
Example of analysis of the FARADAY array with the WW-procedure



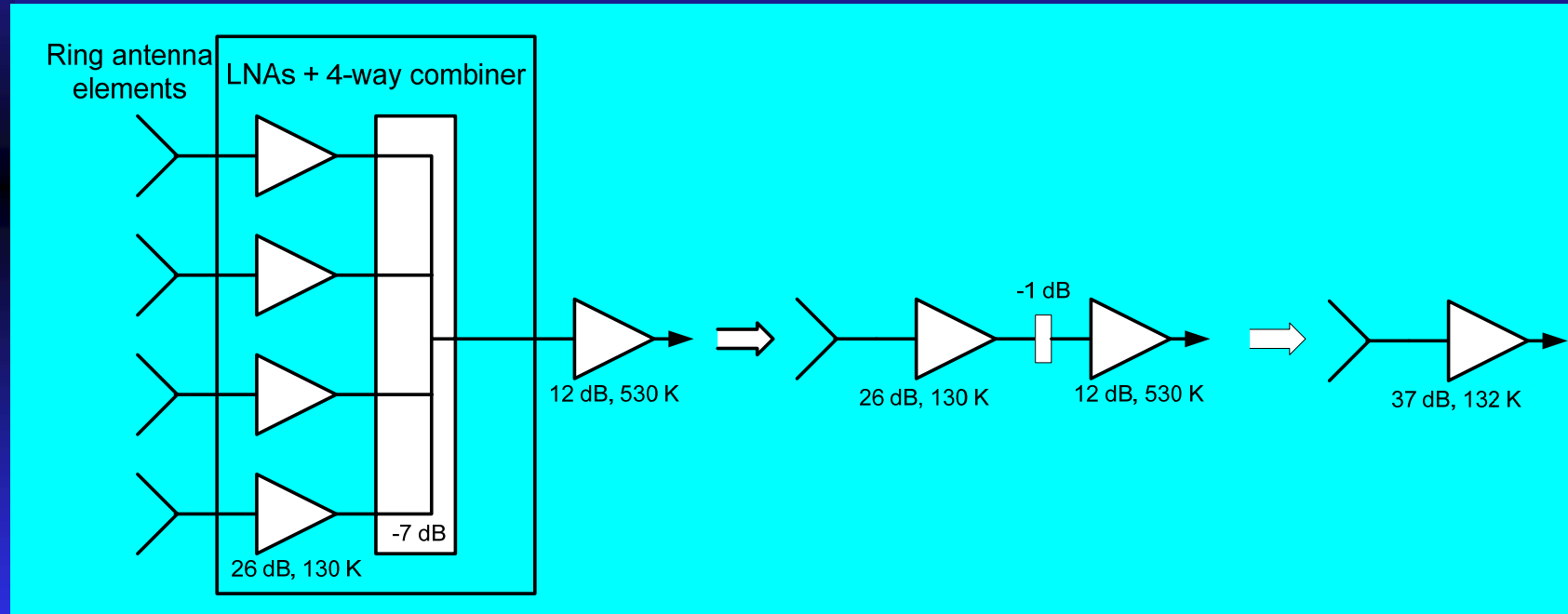
Top view of the array

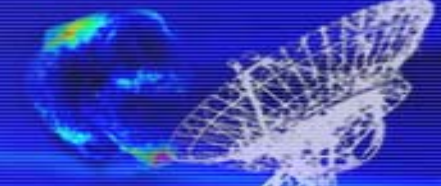


Beam former configuration 2005

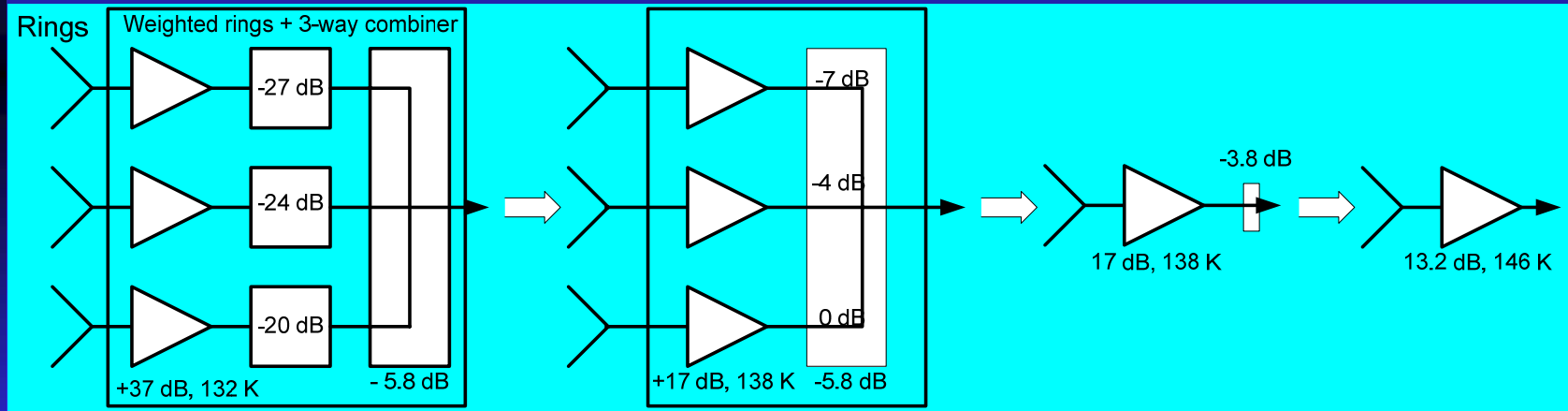


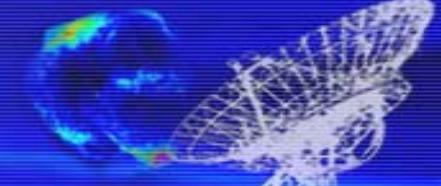
Transforming a ring system into an equivalent two-port receiver



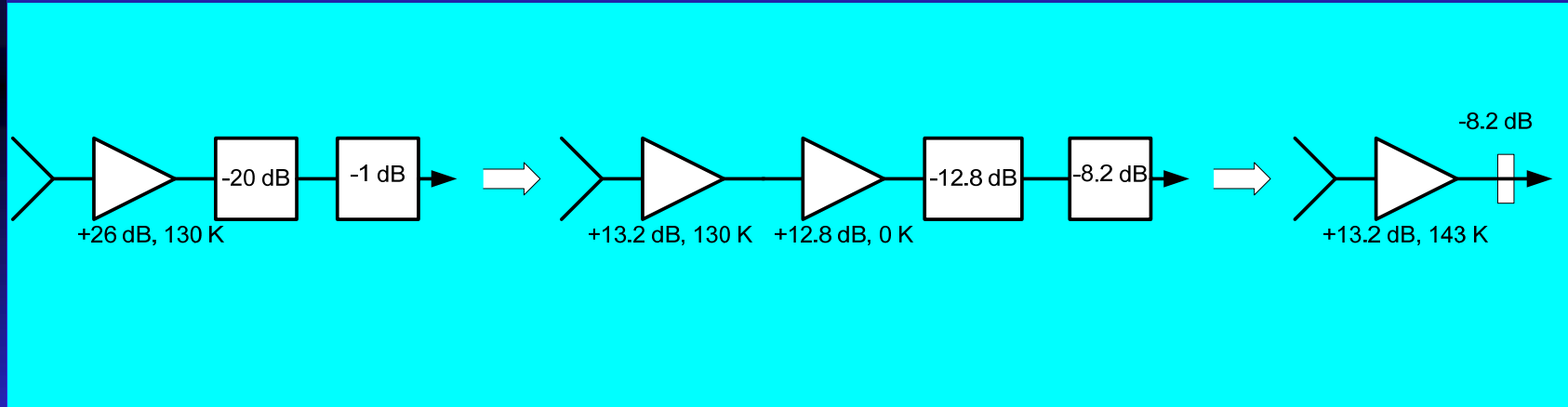


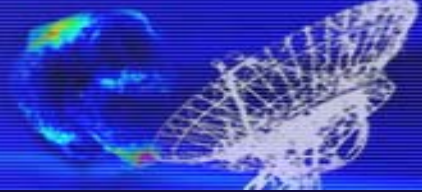
Combination of the weighted rings



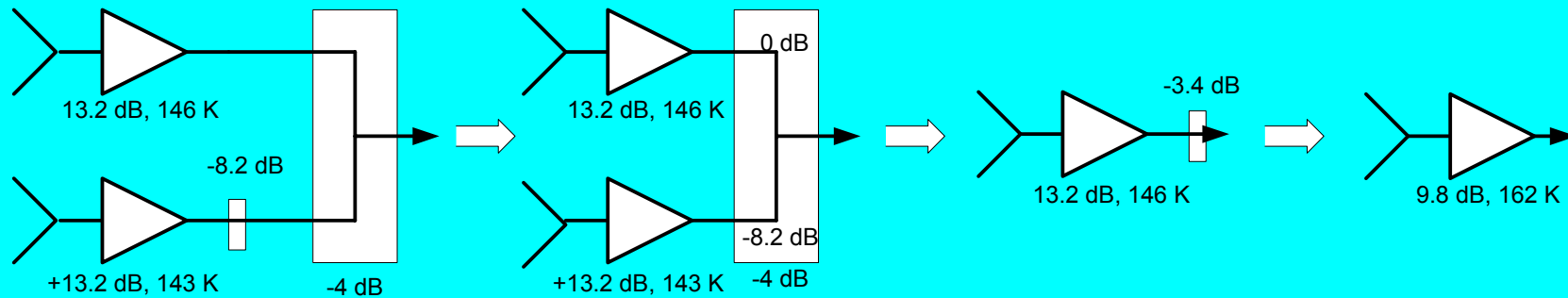


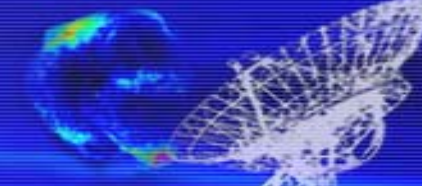
The equivalent central element





The central element added, to give the equivalent FPA receiver





Calculation of array noise temperature (Method 1)

- Weighted average of embedded two-port noise temperatures, using active reflection coefficients

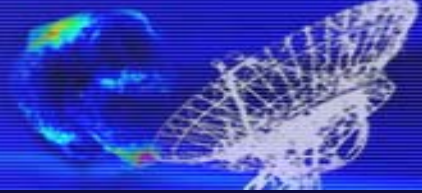
$$T_{array} = \frac{\sum_{n=1}^N G_{av_n} T_{rec_n}}{\sum_{n=1}^N G_{av_n}}$$

$$T_{rec_n} = T_{min_n} + \frac{4R_{n_n} T_0}{Z_0} \frac{|\Gamma_{act_n} - \Gamma_{opt_n}|^2}{|1 + \Gamma_{opt_n}|^2 (1 - |\Gamma_{act_n}|^2)}$$

$$\Gamma_{act_n} = \sum_{m=1}^N \frac{G_m}{G_n} S_{nm} e^{j(\phi_m - \phi_n)}$$

$$G_{av_n} = G_{rec_n} (1 - |\Gamma_{act_n}|^2)$$

- Represents noise temperature at array input, as would result from a measurement with a hot and cold load at the array input
- Calculation method implemented in Excel spread sheet

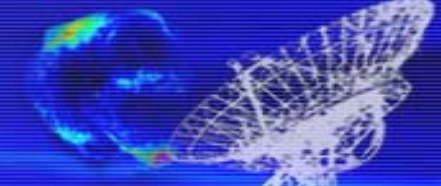


Calculation of array noise temperature (Method 2)

- CAESAR, numerical tool based on MM for modeling antenna structures, combined with microwave system simulator, using noise wave correlation and S-matrices to calculate array output noise for a certain sky brightness temperature. Program input includes LNA parameters
- Two calculations are performed to determine the array noise temperature, as the array response to a sky signal, thus by definition the noise temperature at the input of the antenna array

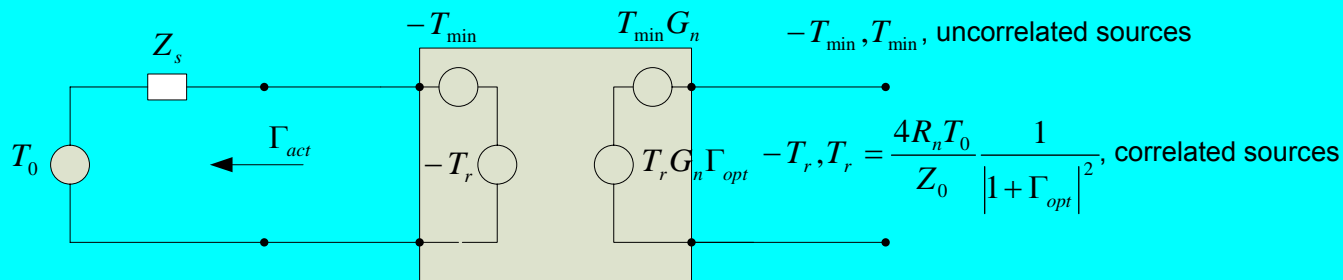
$$T_{array} = \frac{P_{out} (T_{sky} = 0)}{P_{out} (T_{sky} = 1, T_{rec} = 0)}$$





Calculation of array noise temperature (Method 3)

Calculate array output noise from individual channel two-port contributions, using 4 noise sources derived from noise wave formulas given by Wedge

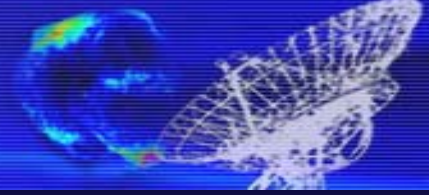


Correlated output noise density $kT_r G_n |\Gamma_{opt} - \Gamma_{act}|^2$

Uncorrelated output noise density $(1 - |\Gamma_{act_n}|^2) kT_{\min} G_n$

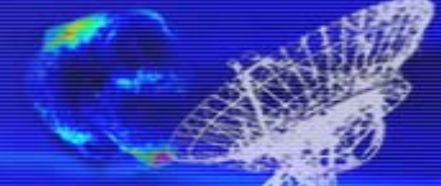
Use array available gain $G_{av_{array}} = \sum_{n=1}^N G_{av_n} = \sum_{n=1}^N G_n (1 - |\Gamma_{act_n}|^2)$ to calculate array input noise temperature:

$$T_{array} = \frac{\sum_{n=1}^N \left[(1 - |\Gamma_{act_n}|^2) T_{\min_n} G_n + T_{r_n} |\Gamma_{opt_n} - \Gamma_{act_n}|^2 G_n \right]}{\sum_{n=1}^N G_{av_n}}$$



Calculation of array noise temperature (Method 4)

- Use of Y-parameter models with associated noise sources (O'Sullivan – CSIRO)
- Different from other three methods, but gives the same numerical results in comparing calculations for a 7-element array with other methods



Available array gain and coupling efficiency

- Array coupling and excitation are incorporated in the active reflection coefficient, which may be >1 in FPAs

$$\Gamma_{act_n} = \sum_{m=1}^N \frac{G_m}{G_n} S_{nm} e^{j(\phi_m - \phi_n)}$$

- Coupling efficiency as defined by Kehn et al is also a measure of array coupling and excitation

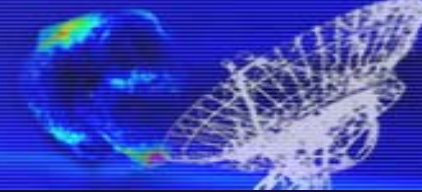
$$\varepsilon_c = \frac{P_{rad}}{P_{inc}} = 1 - \frac{P_c}{P_{inc}} = 1 - \frac{\sum_{n=1}^N |b_n|^2}{\sum_{m=1}^M |a_m|^2}, b_n = \sum_{m=1}^M S_{mn} a_m$$

- Relation

$$\varepsilon_c = 1 - \frac{\sum_{n=1}^N G_n |\Gamma_{act_n}|^2}{\sum_{m=1}^M G_m}$$

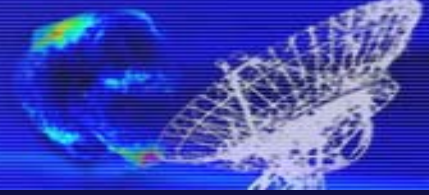


$$\varepsilon_c \sum_{n=1}^N G_n = \sum_{n=1}^N G_n (1 - |\Gamma_{act_n}|^2) = G_{av}$$



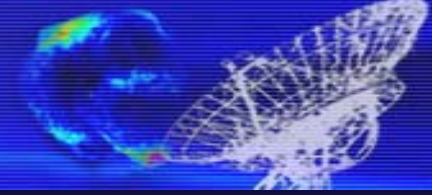
References

- Wedge, S.W., Rutledge, D.B., “Wave Techniques for Noise Modeling and Measurement”, IEEE MTT, Vol.40. No.11, November 1992, pp. 2004-2012
- Kehn, M.N.M., Ivashina, M.V., Kildal, P.-S., Maaskant, R.,”Coupling Efficiency of Wideband Dense Focal Plane Array Feeds for Reflector Antennas – Part I: Definitions and Theoretical Study on a Hypothetical Hard Waveguide Array”, submitted to IEEE Trans.Antennas Propagat., July 2007
- Woestenburg, E.E.M., “Calculation of the Noise Temperature of Focal Plane Array Receiver Systems”, ASTRON Doc. No. RP 269, September 2007

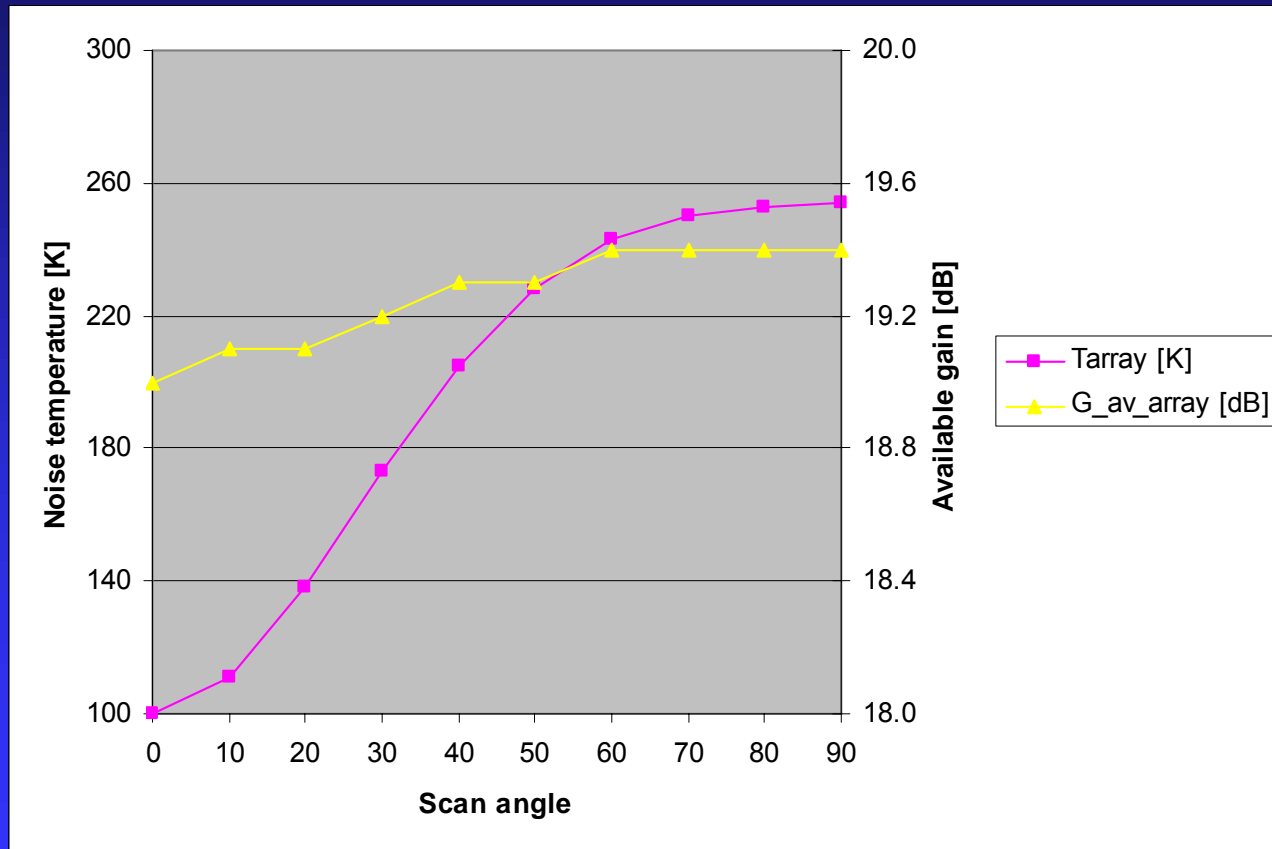


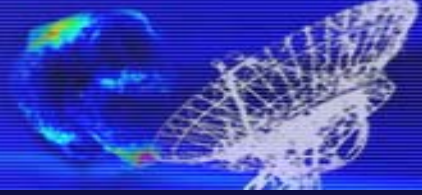
Calculation examples for simple FPAs

- Two-element dipole array with equal weighting gives the same results with simulator and noise formula calculations
- 3-element array with equal weighting of outer elements
- 5-element FPA at 2.3 GHz also gives consistent results (not presented here)
- 7-element array (Y-parameters O'Sullivan)

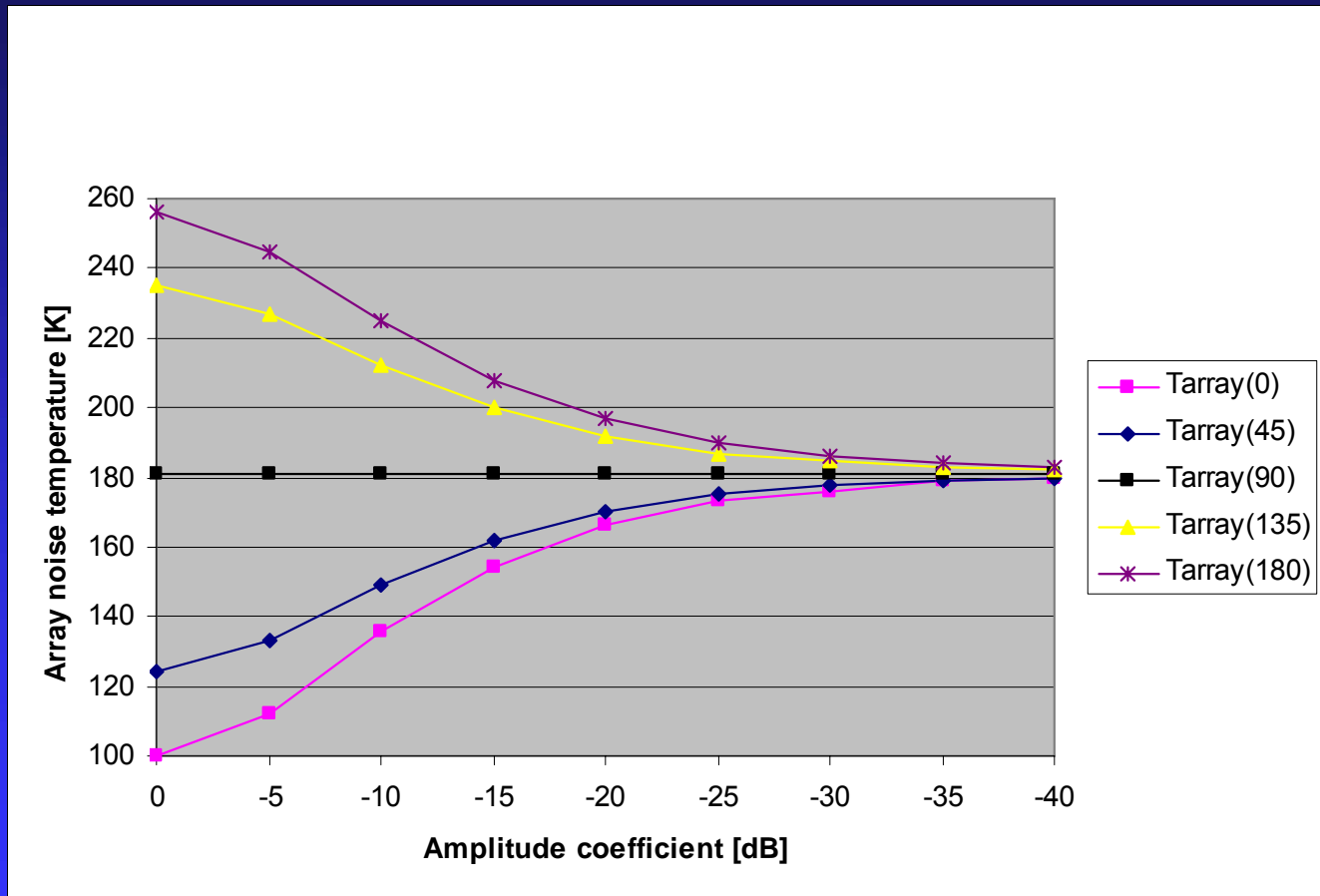


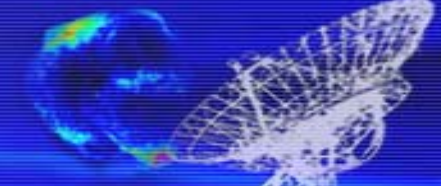
Noise temperatures and available gains as a function of scan angle for the two-element array (method 2) for methods 1 and 3, showing overlapping results



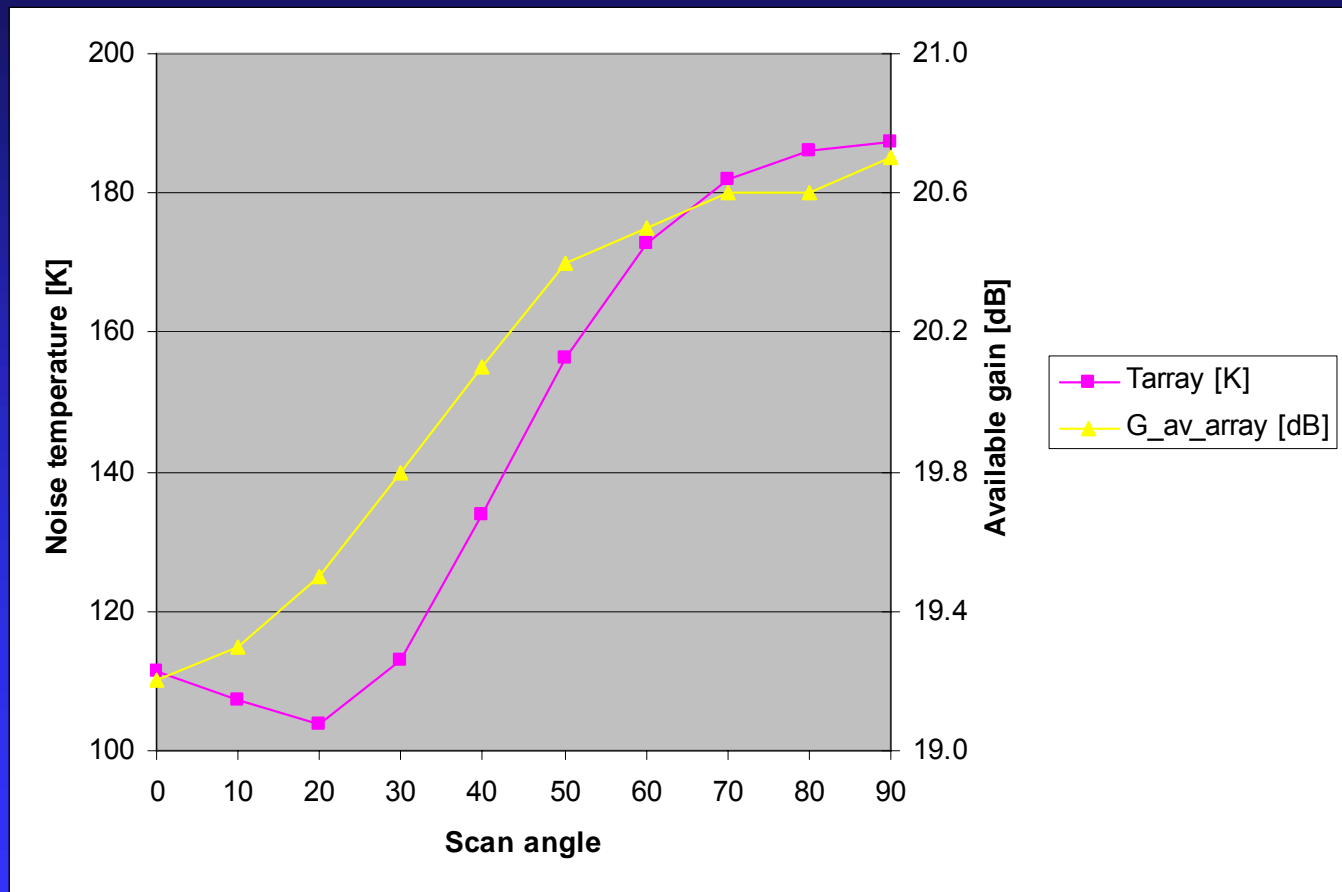


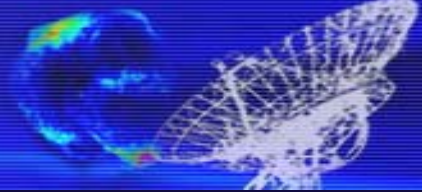
Two-element array noise temperature as a function of amplitude coefficients for five different phases (Methods 1 and 3)



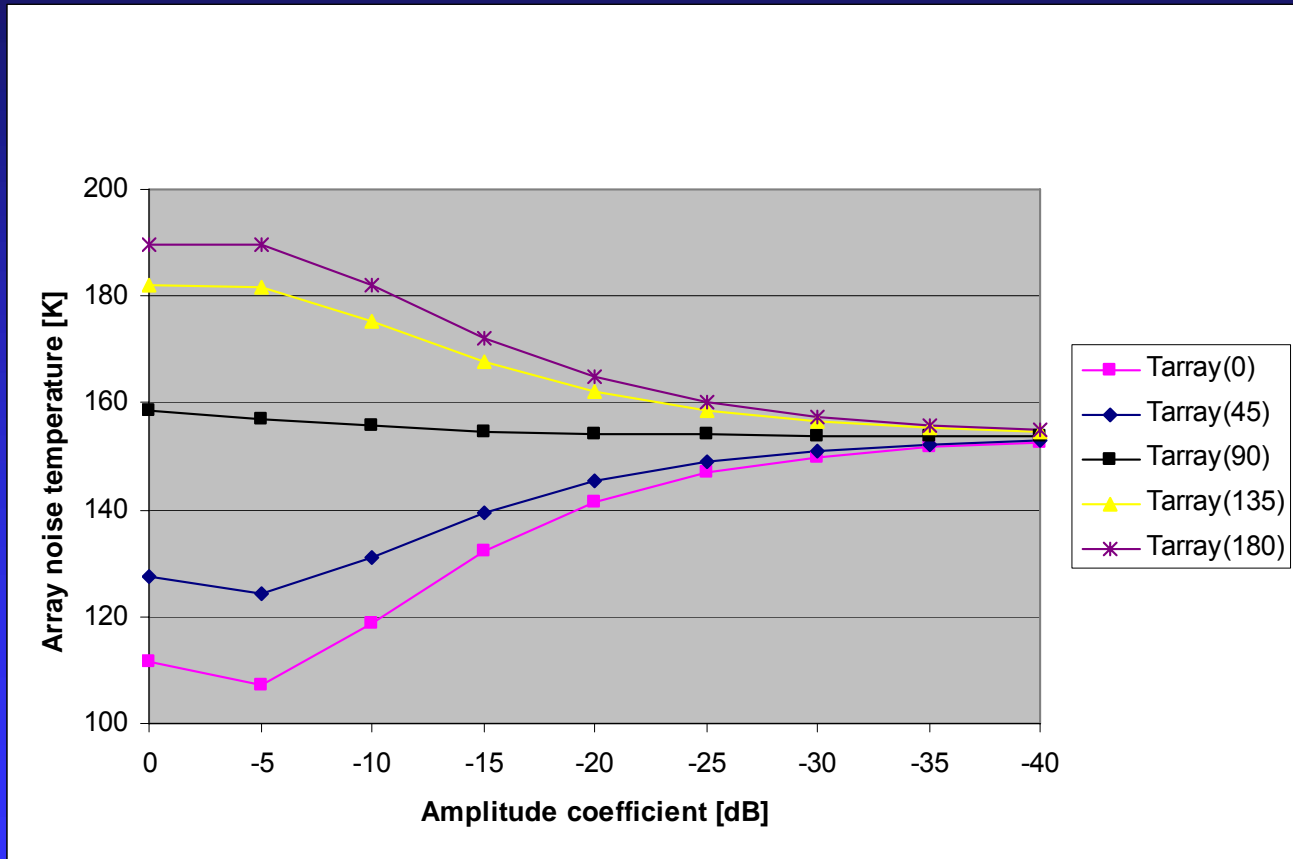


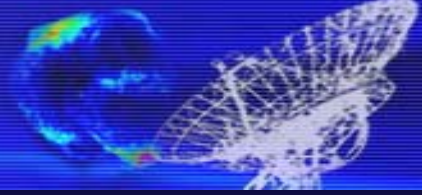
Noise temperature and available gain as a function of scan angle for a three-element array (Methods 1 and 3)



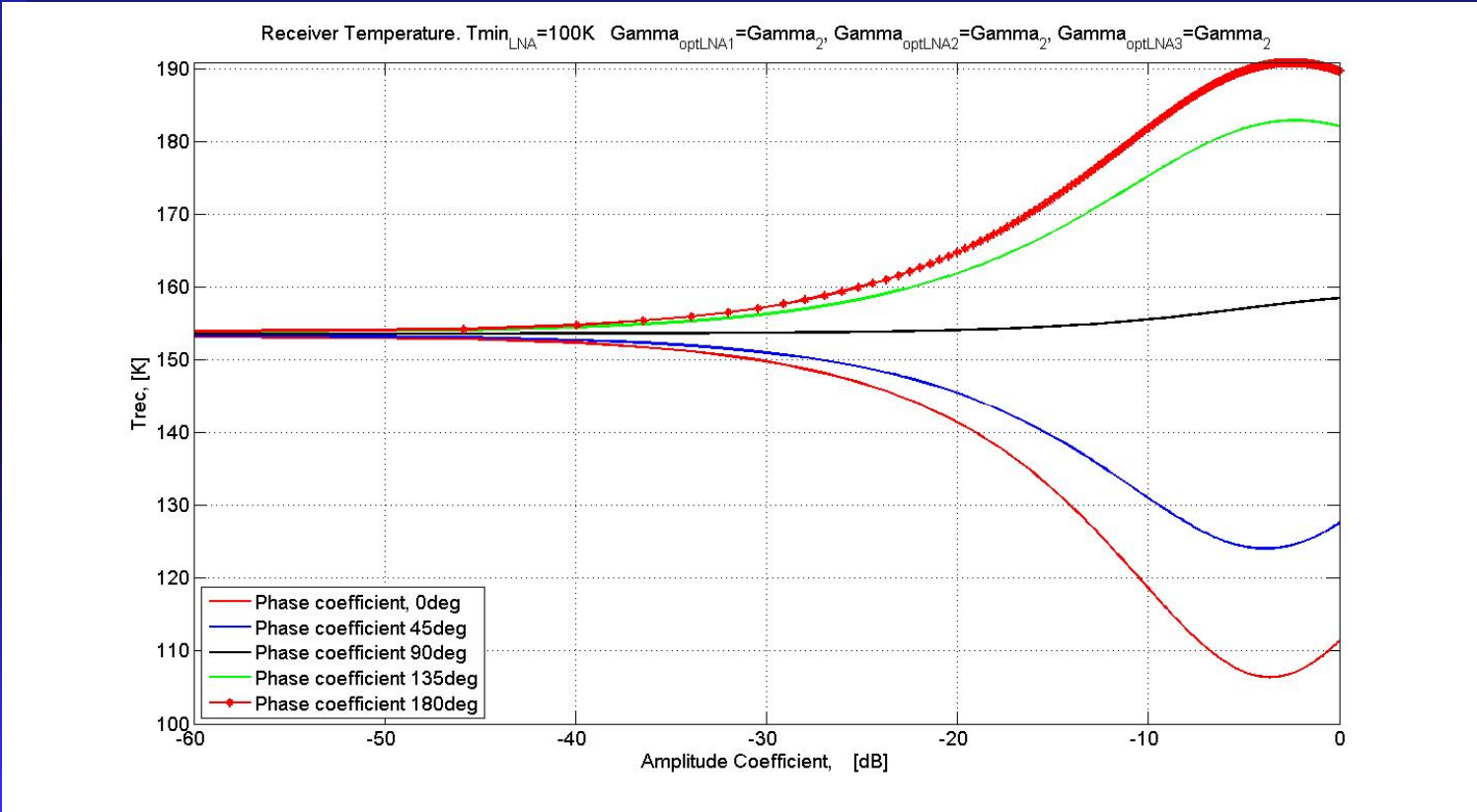


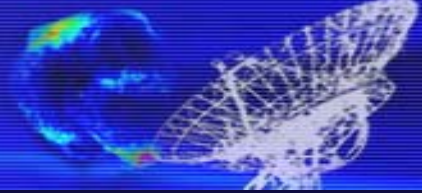
Three-element array noise temperature as a function of amplitude coefficient for the outer elements, at five different phases (Method 1 and 3)





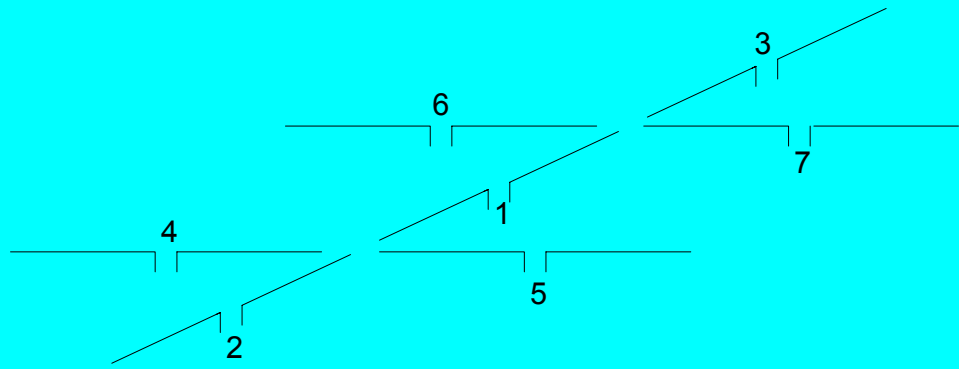
Results obtained with CAESAR for the three-element array for comparison with previous picture. Results are exactly the same for Methods 1, 2 and 3



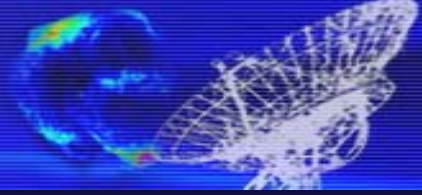


Geometry and configuration of a 7-element array

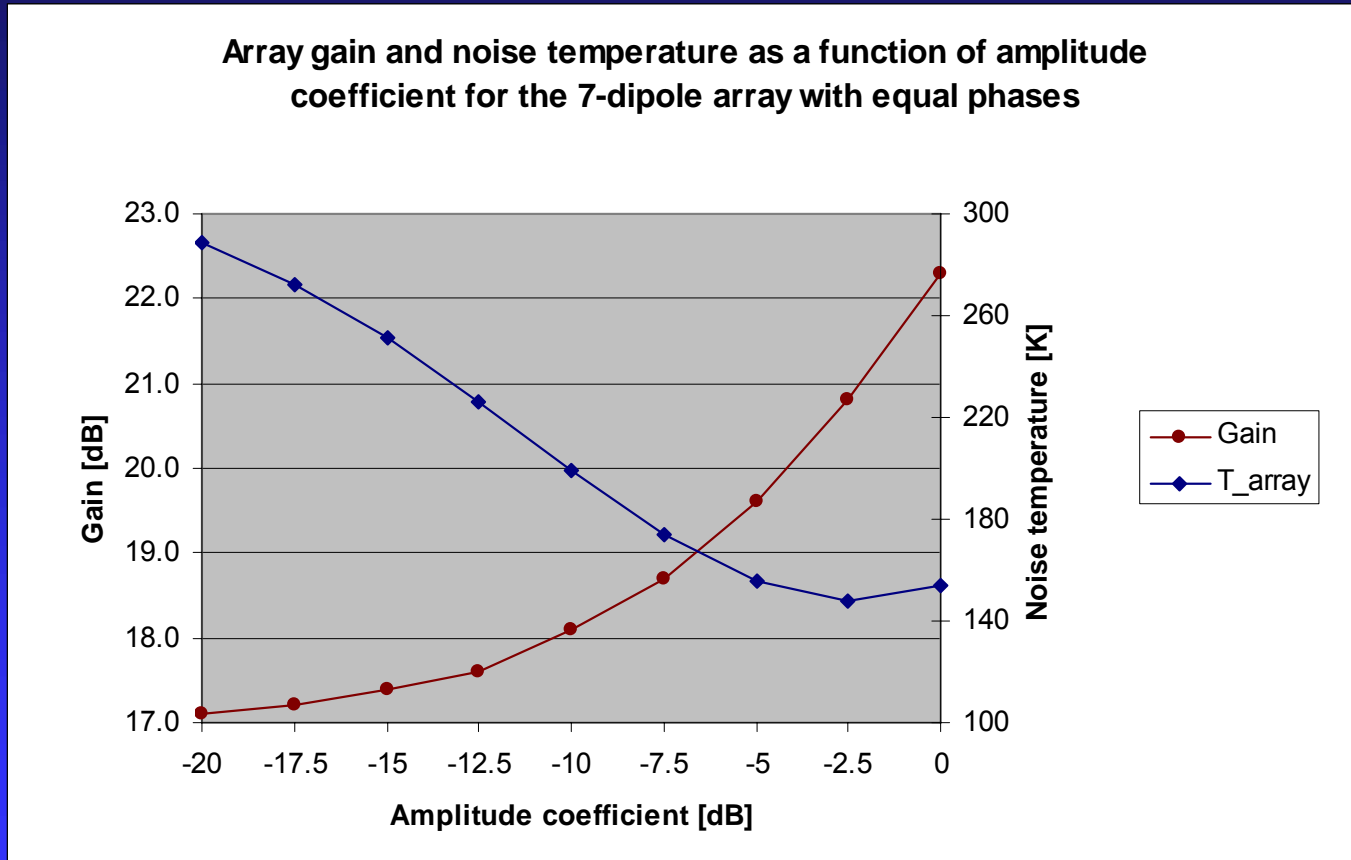
Dipole length	0.150 m
Dipole width	0.001 m
Distance between dipoles	0.152 m
Distance to ground plane	0.075 m
Frequency	1 GHz

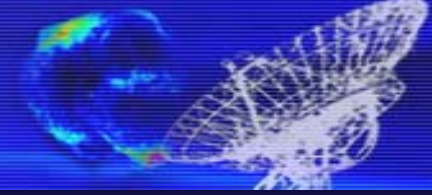


Elements 1, 2 and 3 have LNAs with $T_{\min} = 120K, \Gamma_{opt} = S_{11}, R_n = 25$
Equal weights for elements 2 and 3
Elements 4, 5, 6 and 7 have loads at 300 K

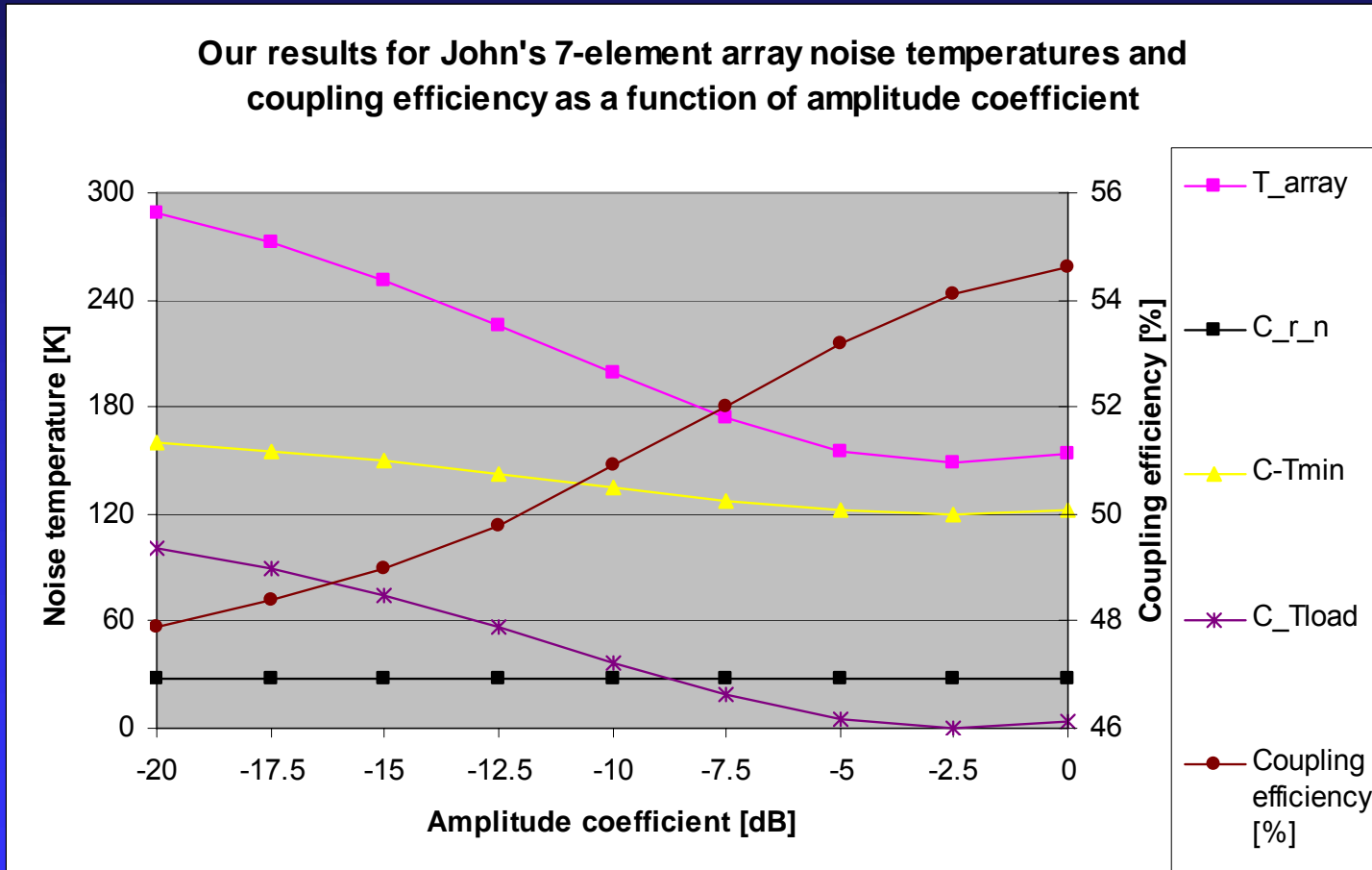


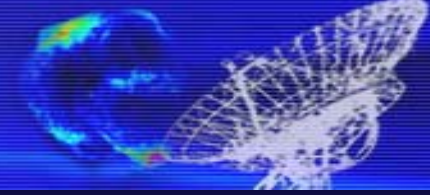
Results with Methods 1, 2 and 3 for the 7-element array, compared to noise temperature calculations with Method 4 (blue dots)





Detailed comparison of noise contributions for ASTRON and CSIRO calculations





Conclusions

- Arrays may be described in terms of a single antenna and receiver chain. (Uncorrelated) gain and noise of the array receiver are found by replacing the array receiver with a single receiver chain with weighted properties of the array receiver elements. The antenna gain accounts for the phasor summation of signals from the individual antenna elements
- Receiver element noise and available gain should include contributions from array coupling, which may be described by the active reflection coefficient
- Calculation of noise properties of complex FPAs may be done with a simulation tool (CAESAR), but also easy to use two-port noise formulas
- Examples of verification with 4 methods for a number of relatively simple FPAs, with excellent agreement